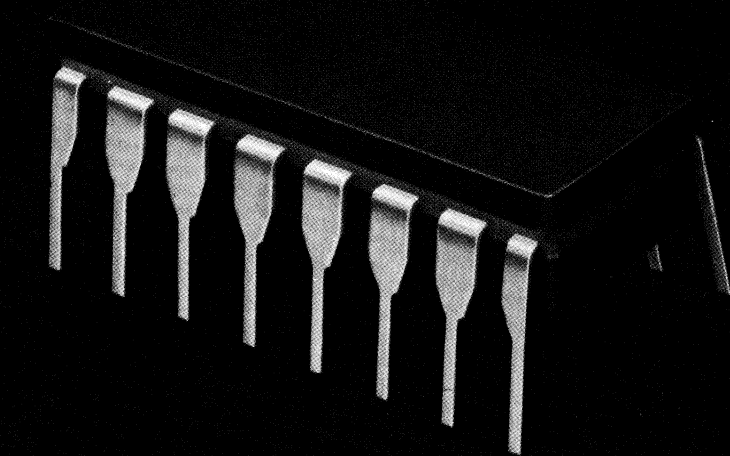


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1982 SEMICONDUCTOR CATALOG

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PART I - SWITCHING COMPONENTS

SECTION I - SWITCHING REGULATORS

LAMBDA LAS 3800

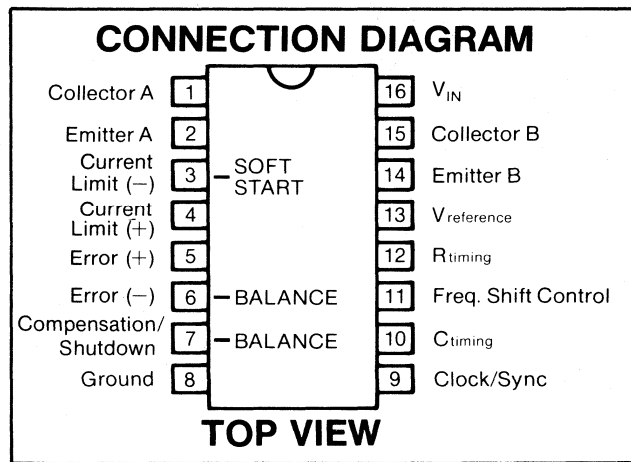
FEATURES

- 12-40 volt operation
- 1.65 volt reference
- DC to 500 KHz oscillator
- Programmable dead time
- Programmable frequency shift
- Multiple device synchronization
- Linear pulse width modulator
- Double pulse suppression
- Uncommitted error amplifier
- High speed current limit comparator
- Current limit frequency shift
- Dual .5 ampere output transistors
- Output over-current protection
- Overvoltage shutdown
- Thermal shutdown
- Soft start capability

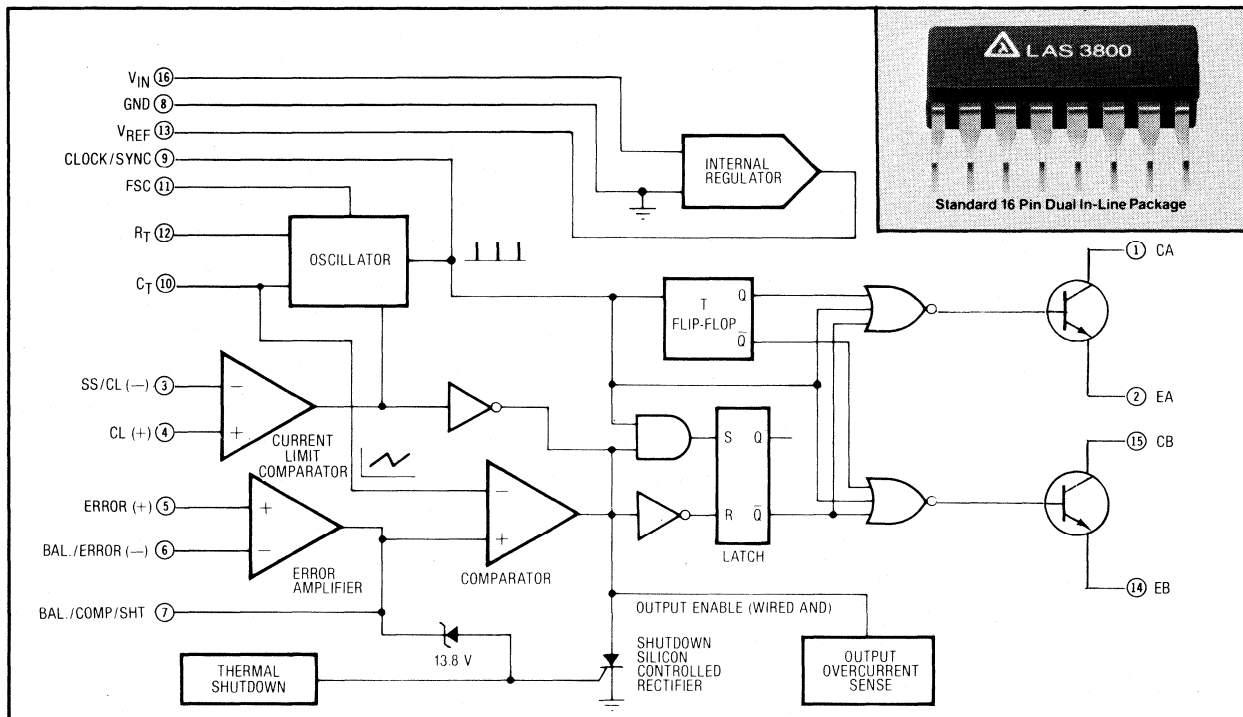
DESCRIPTION

The LAS 3800 is a high performance monolithic integrated circuit switching regulator designed for use in fixed frequency power control applications such as switching power supplies and motor controls. Included in the hermetically sealed Dual in Line package are a temperature compensated voltage reference, sawtooth oscillator with overcurrent frequency shift, linear 'trailing edge' pulse width modulator with double pulse suppression logic, error amplifier, high speed current limit comparator, two symmetrical NPN output devices with overcurrent protection, shutdown logic and provision for dynamic volt-time symmetry correction in double ended systems. This circuit can be used to implement switching regulators of either polarity, transformer coupled DC to DC converters, transformerless voltage doublers, polarity converters, and DC/AC motor controls. The LAS 3800 has been characterized for operation over the -55°C to 125°C temperature range.

PIN CONFIGURATION



BLOCK DIAGRAM



ABSOLUTE MAXIMUM RATINGS

PARAMETER	RATING	UNIT
Input Voltage Range	12 — 40	Volts
Collector Supply Voltage	40	Volts
Operating Temperature Range	-55 to 125	°C
Thermal Resistance; junction/ambient	100	°C/Watt
Minimum Timing Resistor	5000	Ohms

ELECTRICAL CHARACTERISTICS

Unless otherwise stated, these specifications apply for $T_j = 0^\circ\text{C}$ to 125°C , $V_{IN} = 15$ volts, and $f = 20$ KHz (see Test Circuit)

PARAMETER	TEST CONDITIONS	TEST LIMITS		
		MIN	MAX	UNITS
REFERENCE SECTION				
Reference Voltage	$T_j = 25^\circ\text{C}$	1.56	1.74	V
Line Regulation	$V_{IN} = 12$ to 40 volts; $T_j = 25^\circ\text{C}$	—	.01	%/V
Temperature Coefficient		—	.01	%/°C
OSCILLATOR SECTION				
Initial Frequency Accuracy	$T_j = 25^\circ\text{C}$	—	± 6	%
Line Regulation of Frequency	$V_{IN} = 12$ to 40 volts, $T_j = 25^\circ\text{C}$	—	.036	%/V
Frequency T.C.		—	.05	%/°C
Maximum Frequency		—	500	KHz
Dead Time Initial Accuracy	$T_j = 25^\circ\text{C}$	—	$\pm 20\%$	%
Dead Time T.C.		—	+ .33	%/°C
Minimum Dead Time		1.5	—	μS
ERROR AMPLIFIER SECTION				
Input Offset Voltage	$T_j = 25^\circ\text{C}$	—	± 25	mV
DC Voltage Gain		60	—	dB
Slew Rate	$T_j = 25^\circ\text{C}$.4	.7	v/ μs
Output Source Current		—	10	mA
Output Sink Current		—	.8	mA
Input Common Mode Range		.6	11	V
Output Common Mode Range		.4	8	V
Supply Rejection		70	—	dB
CURRENT LIMIT COMPARATOR				
Threshold Voltage	$T_j = 25^\circ\text{C}$	2.28	2.52	V
Threshold Voltage T.C.		-.3	-.4	%/°C
Common Mode Range		0	$V_{IN} - 5$	V
Input Bias Current		—	-.6	mA
Time Delay to Outputs Off		—	.9	μs
OUTPUT SECTION				
Collector-Emitter Voltage			40	V
Output Current Limit		.5A	1.4	A
On Voltage	$I_{OUT} = .5\text{A}$, $T_j = 25^\circ\text{C}$	—	2	V
Leakage Current	$V_{CE} = 40\text{V}$	—	.5	mA
Rise Time	$I_{OUT} = .5\text{A}$.2	μs
Fall Time	$I_{OUT} = .5\text{A}$.1	μs
SHUTDOWN LOGIC				
Shutdown Temperature		150	200	°C
Over-Voltage Threshold	$T_j = 25^\circ\text{C}$	13	14.5	V
POWER CONSUMPTION				
Standby Current (I_Q)	Test Circuit, $T_j = 25^\circ\text{C}$	7	13	mA

ESSENTIAL DESIGN FEATURES THAT SEPARATE LAMBDA FROM THE REST

DYNAMIC VOLT-TIME BALANCING

Dynamic volt-time balancing is essential in double-ended switching power supply systems to prevent transformer saturation. Saturation can occur whenever a net DC voltage exists across the switching transformer. This voltage reduces primary inductance causing switching transistor current to increase rapidly. Excessive peak transistor current reduces maximum system power and can result in switching transistor failure when available drive current is not sufficient to cause transistor saturation. Transformer saturation is eliminated by forcing alternate half-cycle volt-time products to be equal. (Volt-time = Transformer voltage X pulse width "ON" time). Equal volt-time products eliminate transformer unbalance and results in maximum power transmitted by the system. Volt-time balancing is accomplished during steady-state and transient load conditions thereby assuring circuit and component reliability for all operating conditions.

CURRENT LIMIT/FREQUENCY SHIFT

A high-speed current limit circuit is required in switching converters to protect the device from destructive peak currents. The LAS 3800's instantaneous peak current limiting guarantees current limiting in all overload conditions, including transients, by using trailing edge modulation. In some switching converter systems excessive power dissipation occurs in the switching devices (transistors and rectifiers) during current limit. The LAS 3800 reduces current limit power consumption by shifting the device oscillator down in frequency. Detection of an overcurrent condition terminates the device output pulse and reduces the slope of the LAS 3800 ramp for the duration of the timing interval, thus decreasing the timing frequency. Duty cycle (power consumption) is reduced and the user is alerted by an audible switching frequency. Frequency shift may be programmed with one resistor. The current limit frequency shift feature offered in the LAS 3800 enables the design engineer to operate switching transistors at maximum power with excellent reliability.

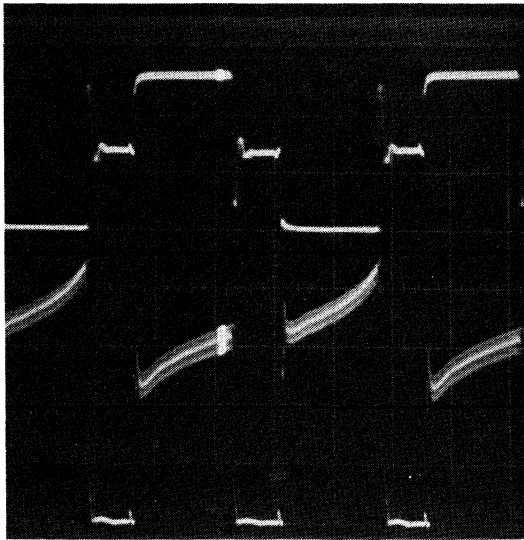
OUTPUT CURRENT CAPABILITY

Output drive capability of 0.5 Amps will directly drive most high-voltage switching transistors. An internal current limit circuit protects the device from accidental short circuit or overload conditions by terminating the output pulse.

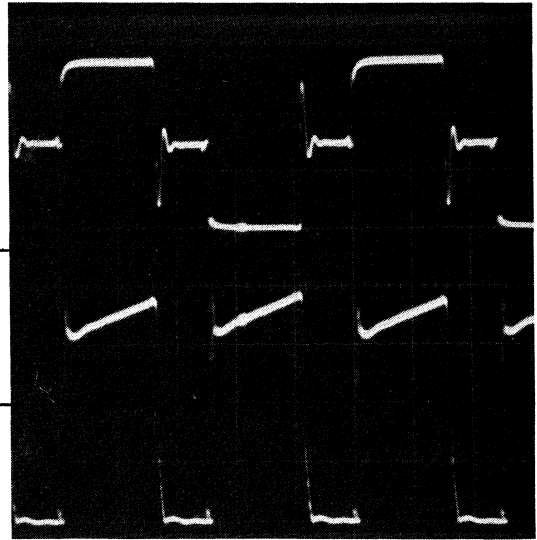
OVERVOLTAGE PROTECTION

Destructive output overvoltage conditions may be sensed with a LAMBDA overvoltage protector which crowbars the output of the power supply. With auxiliary circuitry the LAS 3800 can also be shut down simultaneously which protects the switching system and user load from destructive overvoltage conditions.

INVERTER WITH BALANCE DISCONNECTED

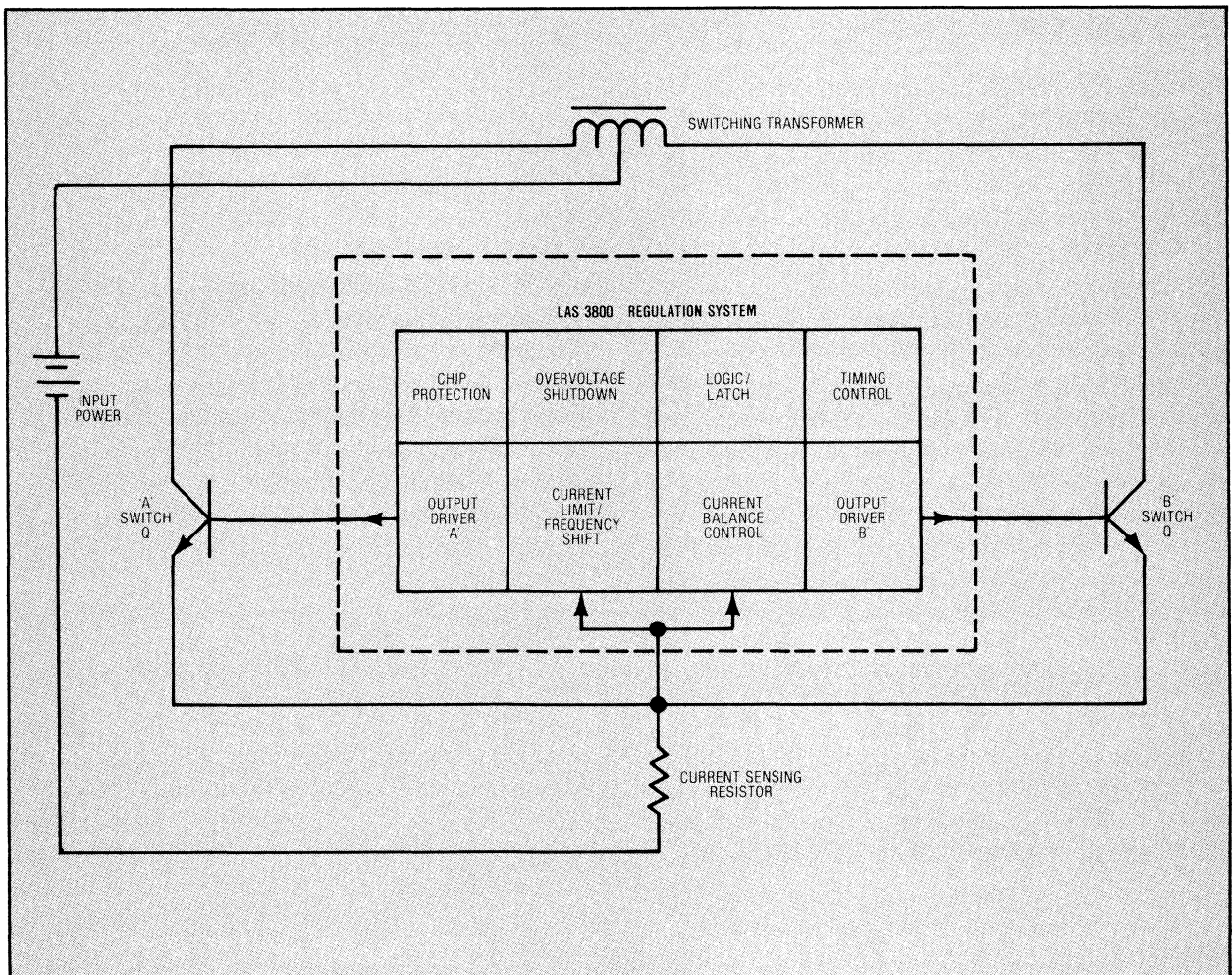


INVERTER WITH LAS 3800 BALANCE CIRCUIT CONNECTED

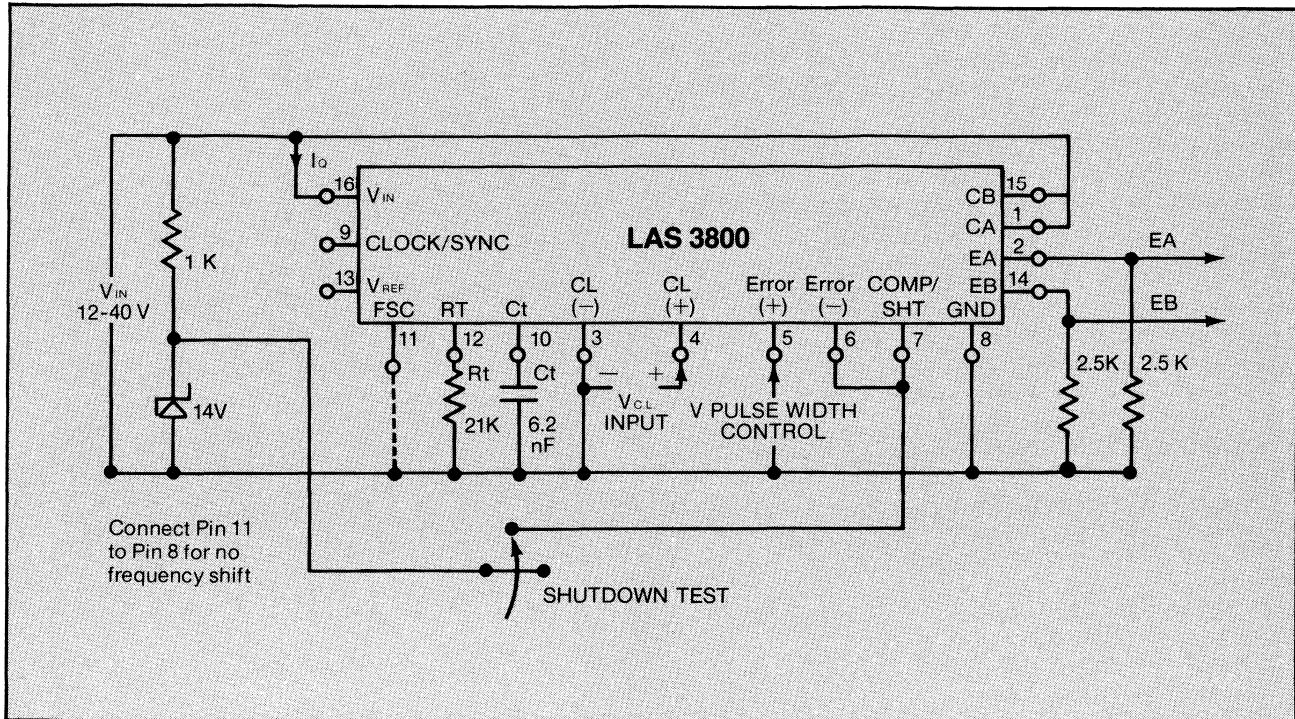


The same inverter operated with/without current balancing. Both systems are delivering maximum power (4 amperes of transistor current). The LAS 3800 system using Dynamic current balance produces 33% more peak output power.

SWITCHING INVERTER SYSTEM WITH LAMBDA LAS 3800



TEST CIRCUIT



APPLICATIONS

OSCILLATOR

The LAS 3800 uses a sawtooth oscillator to generate a linear ramp for the pulse width comparator and system timing. Capacitor charging current is controlled by the timing resistor. Current limit inhibits device outputs and reduces the positive slope of the oscillator ramp. Thus, frequency shifting reduces system power by reducing maximum duty cycle and is a function of the severity of the over current fault condition (maximum occurs at short circuit). Dead time is set by discharging the timing capacitor with a constant current. This timing pulse is used to synchronize all logic and provides output blanking to insure both chip outputs are not on at the same time during the dead time interval. A minimum dead time pulse width of $1.5\mu\text{s}$ should be used to guarantee proper device operation over line and temperature extremes. After selection of (a) system frequency, (b) dead time, and (c) short circuit frequency shift, select the proper external device components as follows: 1) using graph #1 select a timing capacitor for proper dead time, 2) using graph #2 select the proper timing resistor, 3) using graph #3 determine the proper frequency shift control resistor.

VOLTAGE REFERENCE

The voltage reference is a precision temperature compensated zener circuit. A 7K ohm output impedance allows the user to bypass system switching noise; a $.1\mu\text{F}$ capacitor is usually sufficient.

CURRENT LIMIT

The current limit circuit is a high speed 2.4 volt differential comparator designed to minimize output turnoff delay and is operated on a cycle-by-cycle basis via trailing edge modulation. Turnoff delay is less than $.9\mu\text{s}$ under all conditions to minimize any current overshoot. Common-mode range extends from ground to $V_{in} - 5$ volts. Automatic system thermal derating is accomplished by a $-8\text{mv}/^\circ\text{C}$ threshold temperature coefficient. Frequency shift in current limit mode is energized under the following condition:

$$V_{\text{SECONDARY}} \frac{T_{\text{ON(MIN)}}}{T_{\text{ON(MIN)}} + T_{\text{OFF}}} > I_{\text{MAX}} (R_{\text{LOAD}} + R_{\text{LOSSES}})$$

where, $T_{\text{ON(MIN)}} = (T_{\text{DELAY}})_{3800} + (T_{\text{MIN}})_{\text{S.W. TR.}}$

For a practical system $T_{\text{ON(MIN)}}$ will be approximately $2\mu\text{s}$. As R_{LOAD} is decreased, T_{OFF} is increased keeping the output current at some preset limit (I_{MAX}). In the 3800 chip, $T_{\text{ON(MIN)}}$ is never increased during frequency shift so that the volt-time requirements on the inverter transformer are never greater than normal constant voltage operation.

ERROR AMPLIFIER

The error amplifier is a slightly modified operational amplifier with low input offset and high slew rate necessary in dynamic volt-time balancing.

Modifications include a constant-current sink to insure smooth operation of the overvoltage shutdown circuit and limiting current sourcing capability so that the pulse width modulator can be inhibited by shorting chip pin #7 to ground (soft-start) with no damage to the error amplifier. The gain of the current balance circuit is selectable by means of external resistors.

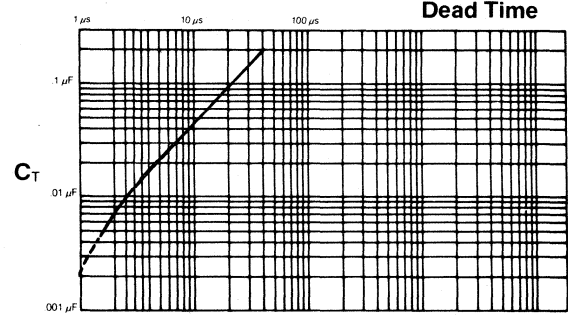
OUTPUT TRANSISTORS

LAS 3800 outputs are two NPN power transistors with collector and emitter uncommitted. Output current limiting of 1.4 amperes is provided and coupled directly to the pulse width logic. The uncommitted collectors are intended to enable the user to reduce device power dissipation if output currents greater than .5 ampere are required.

SHUTDOWN LOGIC

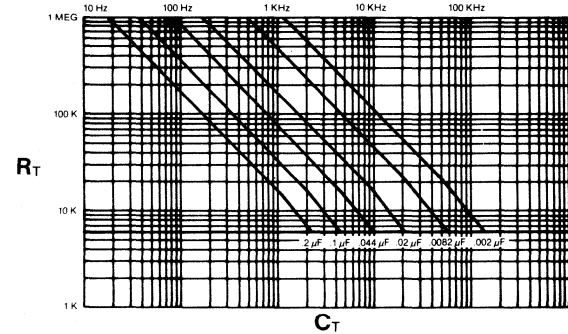
Shutdown logic causes the chip to disable outputs until V_{in} is reduced below 3 volts. Thus, conditions in the system considered dangerous by the user may be detected and used to shutdown the power conversion system. Forcing internal junction temperature above 165° C causes shutdown. External shutdown is accomplished by biasing device pin #7 above 13.8 volts for 10 μ s. Normal output range of the error amplifier is .2—8 volts and cannot initiate shutdown.

graph #1



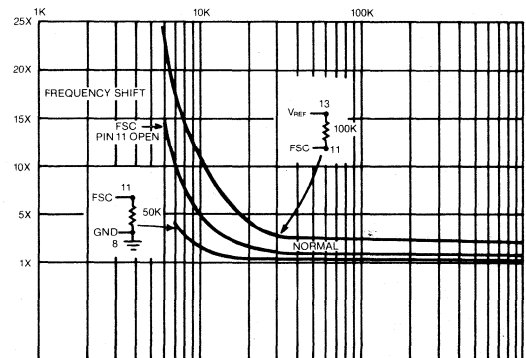
DEAD TIME VS
TIMING CAPACITOR

graph #2
Frequency
Oscillator



FREQUENCY VS
TIMING RESISTOR-CAPACITOR

graph #3



FREQUENCY SHIFT VS
TIMING RESISTOR

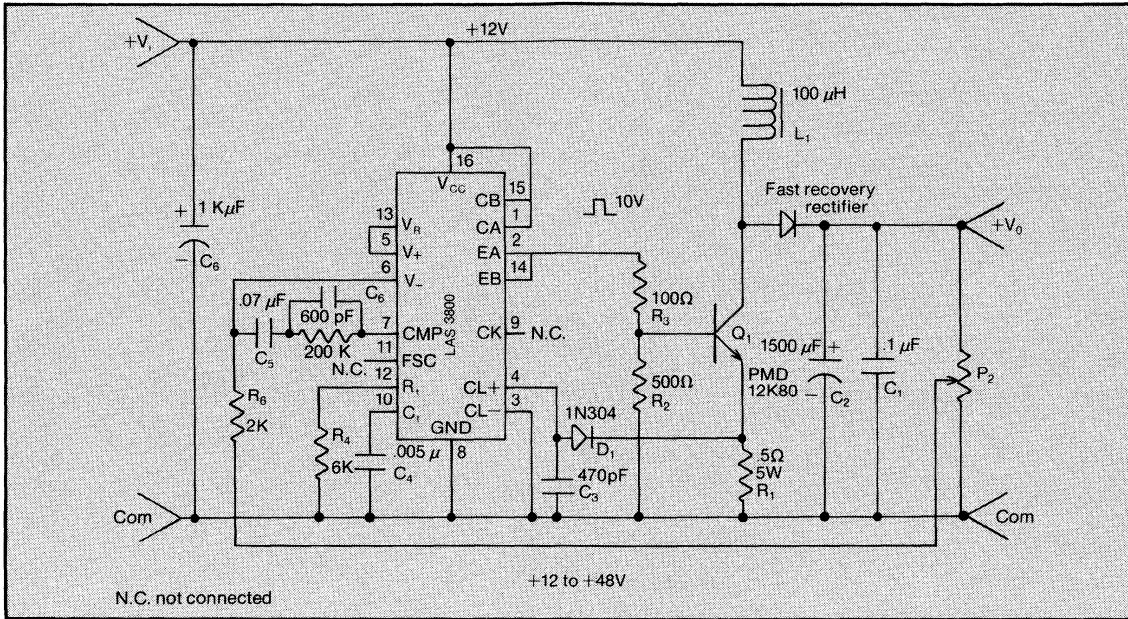
ORDERING INFORMATION

DEVICE TYPE	PRICE & QUANTITY							
	1-24	25-49	50-99	100-249	250-499	500-999	1000-2499	2500-4999
LAS 3800	\$9.95	\$7.95	\$7.95	\$6.95	\$6.95	\$6.95	\$6.20	\$5.95

Contact your area sales representative for higher quantity prices. Device configurations, specifications, and prices subject to change without notice.

TYPICAL APPLICATIONS

LAS 3800 STEP UP DC-DC CONVERTER

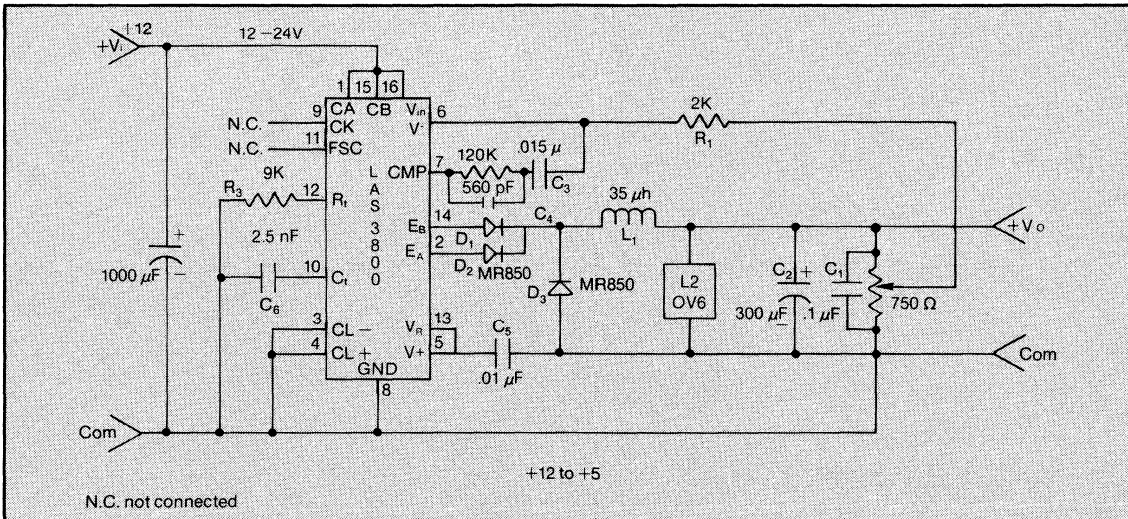


The above circuit is a simple 12 to 48 volt DC/DC using a PMD12K80 as the power switch.

- Frequency—58 KHz
- Load regulation—13 mv
- Line regulation—10 mv (10-14 V)
- Full load current—.25 A
- Max. Switch Current—3 A
- Short circuit frequency shift—14X

- The following LAS 3800 features are used:
1. Error amplifier is voltage control
 2. Internal reference
 3. Current limit to protect 12K80; C₃, D₁ eliminate inductive spike across R₁
 4. Frequency shift
 5. Timing Control

LAS 3800 STEP DOWN DC-DC CONVERTER



The above circuit is a 2.5 watt stand alone 12-5 volt DC/DC converter.

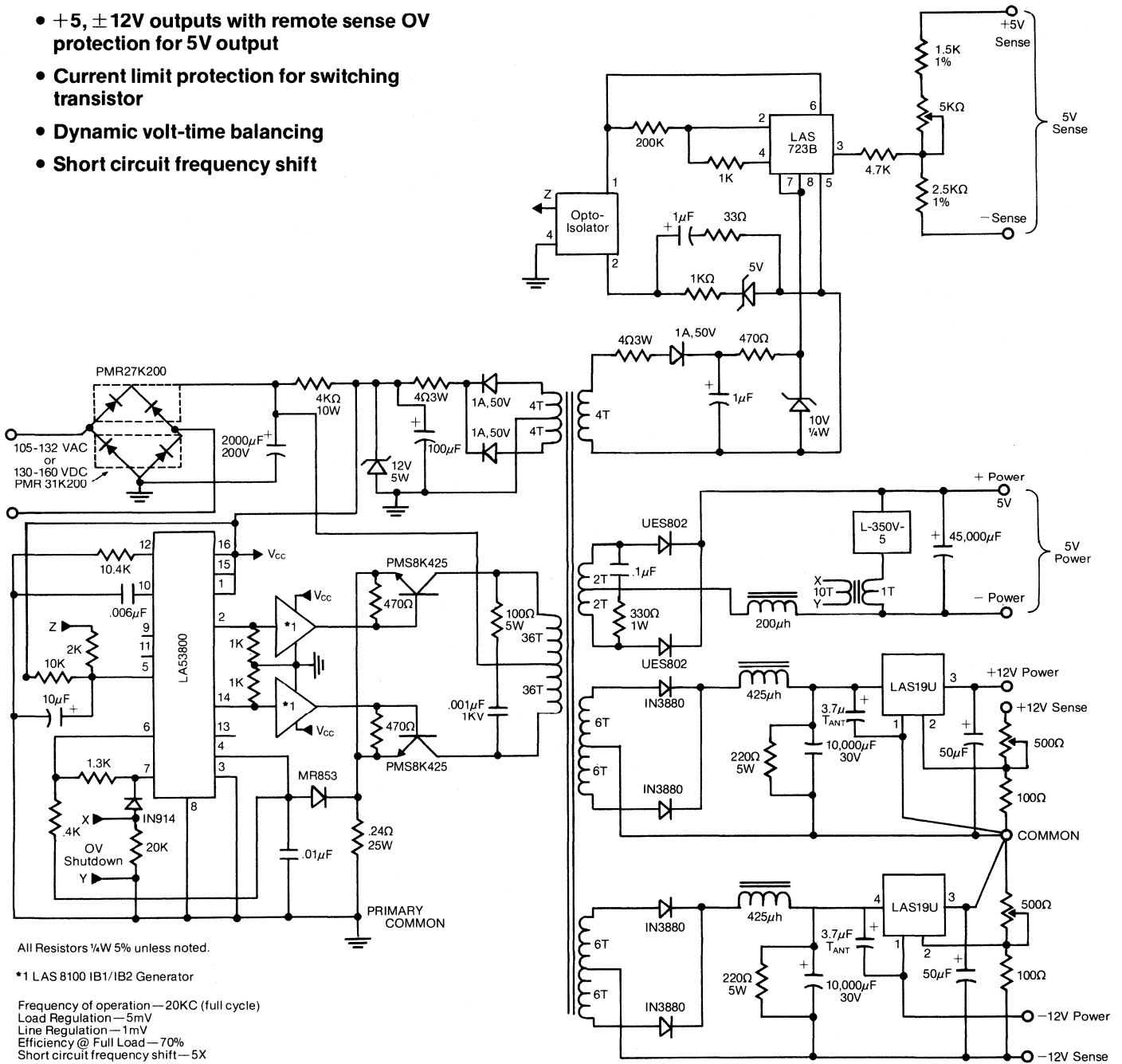
- Frequency—90 KHz
- Load regulation—5 mv (V_{cc}=15 V)
- Line regulation—13 mv (FL, 12-24 V)
- Full load output current—.5 A
- Short circuit current—1.4 A
- Short circuit frequency shift—6X

- The following LAS 3800 features are used:
1. Error amplifier is voltage control
 2. Internal reference
 3. Internal current limit protection
 4. Frequency shift
 5. Timing Control

5 VOLT AT 40A ±12 VOLT AT 5A SWITCHING POWER SUPPLY

FEATURES

- +5, ±12V outputs with remote sense OV protection for 5V output
- Current limit protection for switching transistor
- Dynamic volt-time balancing
- Short circuit frequency shift



LAMBDA LAS 6300

PRELIMINARY

MONOLITHIC SWITCHING REGULATOR

FEATURES

- 5-35 volt operation
- 2.00 volt reference
- DC to 200KHz oscillator
- Linear pulse width modulator
- Double pulse suppression
- Shutdown latch
- Uncommitted error amplifier
- Uncommitted output transistor
- 5 ampere output current
- High speed internal current limit
- Thermal shutdown
- Inherit Soft Start

DESCRIPTION

The LAS 6300 is a high performance monolithic integrated circuit switching regulator designed for use in fixed frequency power control applications such as switching power supplies and motor controls. Included in the hermetically sealed eight pin TO-3 or 10 pin TO-96 packages are a temperature compensated voltage reference, saw-tooth oscillator with over current frequency shift, linear trailing edge pulse width modulator with double pulse suppression logic, error amplifier, and a 5 ampere output transistor with internal current limit protection. The LAS 6300 is characterized for use over the -55°C to 125°C operating temperature range.

ABSOLUTE MAXIMUM RATINGS

PARAMETER	RATING	UNIT
Input Voltage Range	5-20	Volts (TO-96 pack)
Input Voltage Range	5-40	Volts (TO-3 pack)
Collector Supply Voltage	40	Volts
Operating Temperature Range	-55 to 125	$^{\circ}\text{C}$
Thermal Resistance	3.5	$^{\circ}\text{C/W TO-3}$

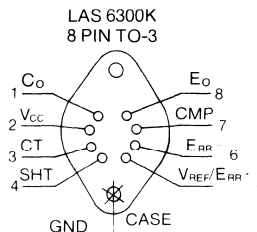
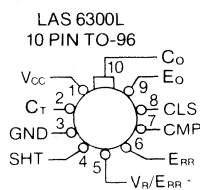
UNLESS OTHERWISE STATED THESE SPECIFICATIONS APPLY FOR $T_J=25^{\circ}\text{C}$, $V_{IN}=12$ VOLTS, $F=40\text{KHz}$, AND TO-3 PACKAGE.

ELECTRICAL CHARACTERISTICS

PARAMETER	TEST CONDITIONS	TEST LIMITS			
		MIN	TYP	MAX	UNITS
REFERENCE SECTION					
Reference Voltage		1.9	2	2.1	Volts
Line Regulation	$V_{CC}=6$ to 35 Volts		0.055		%/V
Temperature Coefficient	$T_J=0^{\circ}\text{C}$ to 125°C		0.005		%/ $^{\circ}\text{C}$
OSCILLATOR SECTION					
Initial Frequency Accuracy			± 10		%
Line Regulation of Frequency	$V_{CC}=6$ to 35 Volts		0.06		%/Volt
Frequency Temperature Coefficient	$T_J=0^{\circ}\text{C}$ to 125°C		0.015		%/ $^{\circ}\text{C}$
Maximum Frequency				200	KHz
Sawtooth Duty Cycle			85		%
ERROR AMPLIFIER SECTION					
Input Offset Voltage				10	mVolt
Transconductance			10		mA/Volt
Output Sink/Source Current			2		mA
Input Common Mode Range		1.5		3.0	Volts
OUTPUT SECTION					
Collector-Emitter Voltage				40	Volts
Output Current		5			Ampere
Saturation Voltage	Emitter grounded $I_O=5\text{A}$			1	Volts
Rise/Fall Time	$I_{OUT}=5$ amperes			1	μsec
Current Limit Sense Voltage			1.7		Volts
Shutdown Threshold			0.8		Volts
POWER CONSUMPTION					
Standby Current			20		mA

PACKAGE OUTLINES

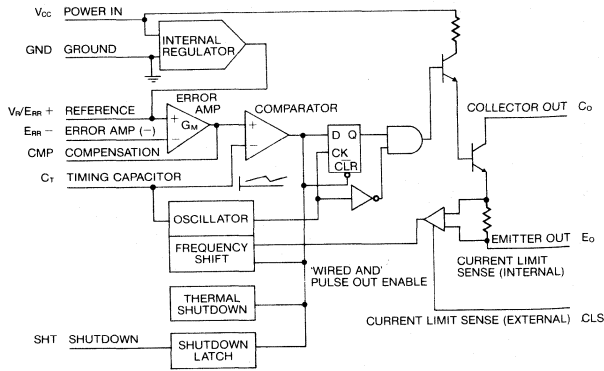
(Top View)



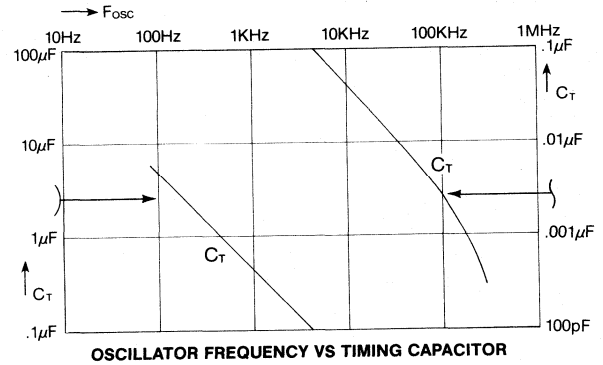
ORDERING INFORMATION

DEVICE TYPE	PRICE QTY							
	1-24	25-49	50-99	100-249	250-499	500-999	1000-2499	2500-4999
LAS 6300K	\$13.00	\$11.50	\$10.25	\$9.50	\$8.75	\$7.50	\$7.00	\$6.60
LAS 6300L	10.00	9.25	8.50	8.00	7.50	6.75	5.90	5.35

BLOCK DIAGRAM



OPERATIONAL DATA



TYPICAL APPLICATIONS

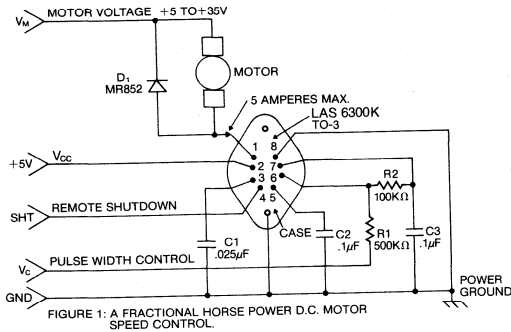


FIGURE 1: A FRACTIONAL HORSE POWER D.C. MOTOR SPEED CONTROL.

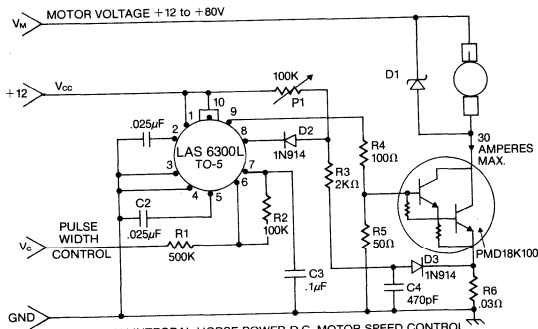


FIGURE 2: AN INTEGRAL HORSE POWER D.C. MOTOR SPEED CONTROL UTILIZING PM063 AS CONTROL AND PMD18K100 AS THE POWER SWITCH.

APPLICATIONS DESCRIPTION

Figures 1 and 2 show variable speed D.C. motor control systems using the Lambda LAS 6300 switching regulator. Speed is controlled by pulse width modulating the average applied motor voltage at a frequency sufficient such that the motor's self inductance is the energy storage element (25 KHz in these applications). Motor frequency is described by the equation:

$$F(\text{motor}) = K_e (T_{on}/T) V_m \text{ where,}$$

K_e — relates frequency to motor D.C. voltage
 T_{on}/T — pulse width modulation duty cycle (DC)
 V_m — motor voltage.

The LAS 6300 controls the duty cycle ratio according to the equation:

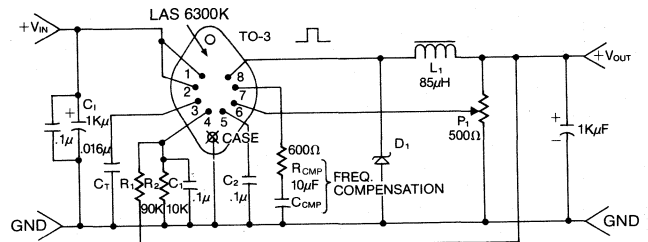
$$T_{on}/T = 2 (V_{ref} (1 + A_v) - 1.225V - V_c(A_v)) \text{ where:}$$

T_{on} — pulse width modulator 'on' time
 T — system switching period (4.0 μ s in these applications)
 A_v — error amplifier gain (R_2/R_1) of LAS 6300
 V_{ref} — LAS 6300 reference voltage
 $1.225V$ — LAS 6300 lower oscillator ramp voltage
 V_c — pulse width control voltage

Thus, control of V_c will modulate the motor frequency (C_3 is a high frequency by pass for the LAS 6300 error amplifier).

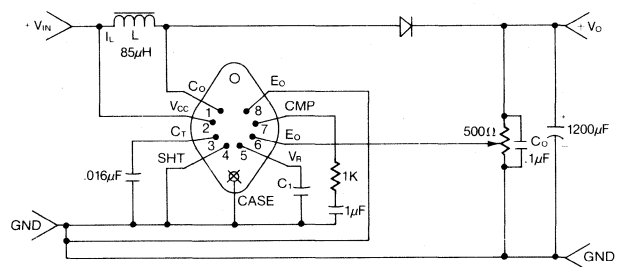
Current limit protection and remote shutdown have been implemented to protect the power switches and motors. In figure 1, current limit is internal to the LAS 6300 and no user programming is necessary. In figure 2 switch current is sensed by R_6 (D_3 and C_4 remove the effect of R_6 's inductance) and feed to the LAS 6300's current limit sense pin via P_1 . R_3 voltage divider ratio. When the voltage on pin 8 of the LAS 6300 reaches 1.7 volts the device output pulse will be terminated and frequency shift initiated (Note: D_2 is necessary to allow the LAS 6300's internal current limit to functional in parallel with external sensing). In both applications current limit frequency shift limits average power dissipation in the power switch and alerts the user to fault conditions. Current limit is also useful for limiting maximum motor torque. Remote shutdown/restart may be implemented using the LAS 6300 SHT pin. Driving this pin's voltage above .8V will actuate the internal shutdown latch that inhibits device output pulses until power is removed or the SHT pin is shorted to common. If the D.C. resistance of Pin 1 (shutdown) to common, is less than 5000 ohms, operation of shutdown will be inhibited.

6300 APPLICATIONS STEP-DOWN CONVERTER



- $F_{osc} = 30Kc$ R_1/R_2 C_1 OUTPUT OV SHUTDOWN
- PIN 1, 2 — V_{cc} INPUT 40 VOLTS MAX.
- PIN 3 — TIMING CAP. = .01 μ F FOR 30KC SWX FREQ.
- PIN 4 — SHUTDOWN INPUT. THRESHOLD = .7V. GROUNDING PIN ENABLES DEV. OUTPUT OVERVOLTAGE THRESHOLD IS 7V ($1 + R_1/R_2$) = 7V. C_1 DECOUPLES SWX NOISE
- PIN 5 — $V_{ERR} + V_{REF} - 2V$ VOLTAGE REFERENCE FOR OUTPUT VOLTAGE. C_2 DECOUPLES SWX NOISE
- PIN 6 — VOLTAGE FEEDBACK VOLTAGE = $V_{REF} IN EQ.$
- PIN 7 — VOLTAGE LOOP COMPENSATION. UNITY GAIN CROSSOVER GOES AS $\frac{1}{R_{CMP} C_{CMP}}$
- PIN 8 — EMITTER OUT (POWER OUT)
- OUTPUT FILTER POLE = 545HZ: TWO DECADES BELOW F_{SWX}

STEP-UP DC/DC CONVERTER



- PIN 1 — POWER DEVICE COLLECTOR OUT
- PIN 2 — DEVICE V_{cc}
- PIN 3 — TIMING CAP. $F_{osc} = 30KC$ for $C_1 = .016\mu F$
- PIN 4 — SHUTDOWN (NOT USED)
- PIN 5 — $V_{ERR} + V_{REF} - 2V$ VOLTAGE REFERENCE (2V). C BY PASSES SWX NOISE
- PIN 6 — $V_{ERR} - V_{REF}$ OUTPUT VOLTAGE FEEDBACK
- PIN 7 — VOLTAGE LOOP COMPENSATION
- PIN 8 — POWER DEVICE EMITTER OUT (GROUNDED)
- OUTPUT FILTER POLE = 500Hz

SECTION II – SWITCHING TRANSISTORS

▲ LAMBDA PMS SERIES

FEATURES

- Guaranteed and 100% tested for I_s/b (second breakdown current) insuring maximum performance at high energy levels.
- Guaranteed and 100% tested for E_s/b (second breakdown energy). Truly the most rugged and superior switching transistors available today.
- Guaranteed collector to emitter breakdown sustaining voltage of 425 volts on all switching transistors.
- Guaranteed DC current gain greater than or equal to 5 on all the three switching transistors.
- Excellent thermal resistance on all switching transistors providing for more usable power and lower operating temperatures.
- All switching transistors are hermetically sealed and are subject to leak testing.
- Full CVD Glass Passivation on all switching transistors.

DESCRIPTION

The PMS series of switching transistors are NPN Silicon Power Transistors particularly suited for line operated switching applications such as switching regulators, inverters, solenoid and relay drivers, motor controls and other high voltage power switching applications.

The devices are fabricated on a passivated epitaxial double diffused structure. The unique structure of the device incorporates multiple emitters which are connected to a common emitter electrode through dielectrically isolated polysilicon ballast resistors. These ballast resistors insure uniform current density in each emitter region and prevent hot spots, thus providing exceptional I_s/b and E_s/b ratings necessary for high-speed power switching applications.

Lambda's switching transistors feature greatly enhanced forward and reverse second breakdown characteristics. However, successful application of these devices will require the use of proper negative base drive (I_{B2}) commensurate with adequate device fall time and storage time characteristics. Presumptions of interchangeability of these devices with any other similar devices should be approached with caution. Device characteristics with regard to switching speed and reverse energy capability are compromises, which are met with large variation in design goals from one manufacturer's part number to another. When in doubt, consult the factory for specific information regarding either device parametric performance or application information.

ABSOLUTE MAXIMUM RATINGS

RATING	SYMBOL	PMS3K425	PMS5K425	PMS8K425	UNIT
Collector Emitter Voltage	V_{CEO}	425	425	425	V_{dc}
Collector Base Voltage	V_{CBO}	470	470	470	V_{dc}
Emitter Base Voltage	V_{EBO}	7	7	7	V_{dc}
Collector Current—Continuous	I_c	3	5	8	A_{dc}
Collector Current—Peak	I_c	3	5	8	A_{dc}
Base Current	I_B	1	2	3	A_{dc}
Total Internal Power Dissipation	$P_D @$ $T_C = 0^\circ C$	150	150	225	Watts
	Derate $T_C > 0^\circ C$	1	1	1.49	W/ $^\circ C$
Temperature Range					
Operating	T_J	-65 to +150	-65 to +150	-65 to +150	$^\circ C$
Storage	T_{STG}	-65 to +200	-65 to +200	-65 to +200	$^\circ C$
Thermal Resistance	$R_{\theta JC}$	1	1	0.67	$^\circ C/W$

ORDERING INFORMATION

DEVICE TYPE	VOLTAGE RATING Volts	CURRENT RATING Amps	PRICE & QUANTITY			
			1-99	100-999	1000-2499	2500-4999
PMS3K425	425	3	\$3.35	\$2.70	\$2.40	\$2.35
PMS5K425	425	5	4.30	3.45	3.10	3.05
PMS8K425	425	8	9.30	7.45	6.70	6.55

Contact your area sales representative for higher quantity prices. Device configurations, specifications, and prices subject to change without notice.

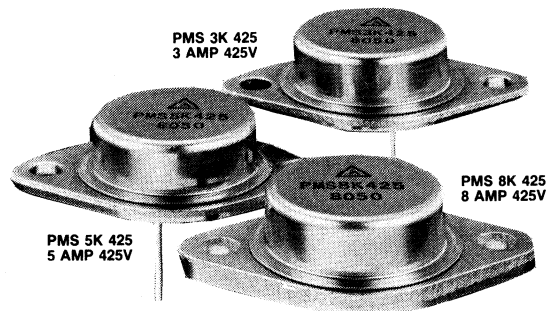
ELECTRICAL CHARACTERISTICS

All parameters are guaranteed at T_j of 0°C to 150°C, unless otherwise specified.

DEVICE TYPE	TEST	SYMBOL	TEST CONDITIONS	MIN.	MAX.	UNITS	NOTES
PMS3K425 PMS5K425 PMS8K425	Collector Emitter Saturation Voltage	$V_{CE(Sat)}$	$\frac{I_C}{I_B} = 5$ $I_C = 3A_{dc}$ $I_C = 5A_{dc}$ $I_C = 8A_{dc}$		2.5	Vdc	1
PMS3K425 PMS5K425 PMS8K425	Base Emitter Saturation Voltage	$V_{BE(Sat)}$	$\frac{I_C}{I_B} = 5$ $I_C = 3A_{dc}$ $I_C = 5A_{dc}$ $I_C = 8A_{dc}$		3.0	Vdc	1
PMS3K425 PMS5K425 PMS8K425	DC Current Gain	h_{FE}	$V_{CE} = 3V_{dc}$ $I_C = 3A_{dc}$ $I_C = 5A_{dc}$ $I_C = 8A_{dc}$	5	25		1
PMS3K425 PMS5K425 PMS8K425	Forward Bias Second Breakdown Current	$I_{s,b}$	$V_{CE} = 100V_{dc}$ 1 second nonrepetitive pulse	1		A _{dc}	3
PMS3K425 PMS5K425 PMS8K425	Second Breakdown Energy	$E_{s,b}$	$V_{CC} = 400V_{dc}$ $\frac{I_C}{I_{B1}} = 5$ $L = 1mH$ $I_{B1} = I_{B2}$ $I_C = 3A_{dc}$ $I_C = 5A_{dc}$ $I_C = 8A_{dc}$	4.5 12.5 32		mJ	3
PMS3K425 PMS5K425 PMS8K425	Collector Emitter Breakdown Voltage (Base Open)	$V_{(BR)CEO}$	$I_C = 100mA_{dc}$	425		Vdc	1, 2
PMS3K425 PMS5K425 PMS8K425	Collector Emitter Sustaining Voltage	$V_{(BR)CER}$	$I_C = 100mA_{dc}$ $R_{BE} = 100\Omega$	425		Vdc	1
PMS3K425 PMS5K425 PMS8K425	Base Emitter Leakage Current	I_{EBO}	$V_{EB} = 7V_{dc}$		1	mA _{dc}	1
PMS3K425 PMS5K425 PMS8K425	Collector Emitter Leakage Current	I_{CER}	$V_{CE} = 450V_{dc}$ $R_{BE} = 100\Omega$		3	mA _{dc}	1
PMS3K425 PMS5K425 PMS8K425	Rise Time R_L is non-inductive	t_r	$V_{CC} = 400V_{dc}$ $\frac{I_C}{I_{B1}} = 5$ $I_{B1} = I_{B2}$ $I_C = 3A_{dc}$ $I_C = 5A_{dc}$ $I_C = 8A_{dc}$		0.75	μs	4
PMS3K425 PMS5K425 PMS8K425	Storage Time R_L is non-inductive	t_s	$V_{CC} = 400V_{dc}$ $\frac{I_C}{I_{B1}} = 5$ $I_{B1} = I_{B2}$ $I_C = 3A_{dc}$ $I_C = 5A_{dc}$ $I_C = 8A_{dc}$		1.5	μs	4
PMS3K425 PMS5K425 PMS8K425	Fall Time R_L is non-inductive	t_f	$V_{CC} = 400V_{dc}$ $\frac{I_C}{I_{B1}} = 5$ $I_{B1} = I_{B2}$ $I_C = 3A_{dc}$ $I_C = 5A_{dc}$ $I_C = 8A_{dc}$		0.3	μs	4

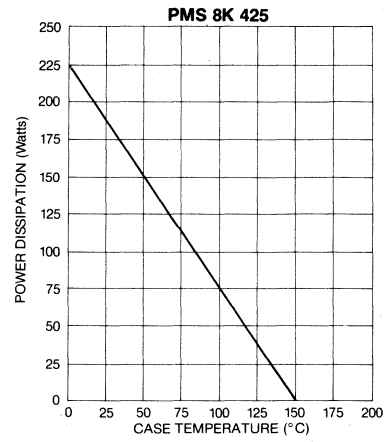
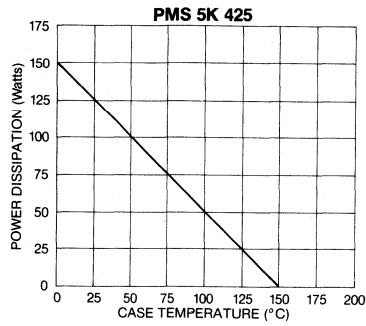
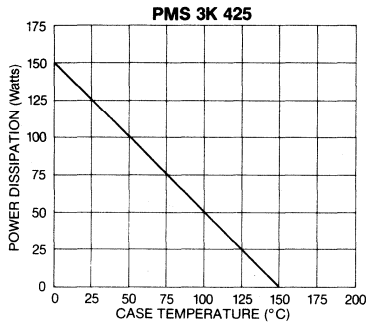
NOTES:

1. Pulse tested: pulse width is less than or equal to 300 microseconds and duty cycle is less than or equal to 2%.
2. $T_j = 25^\circ C$
3. $T_a = 25^\circ C$
4. $T_c = 25^\circ C$

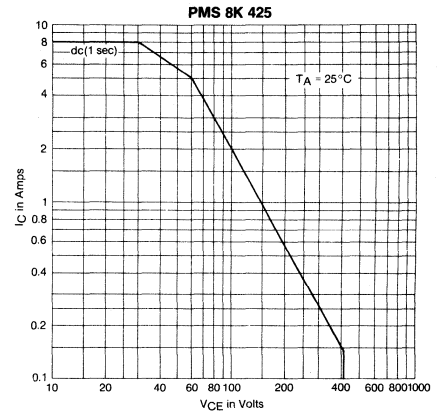
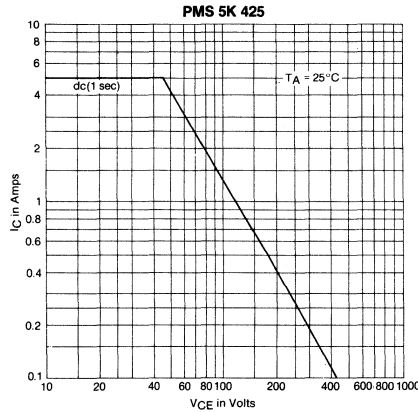
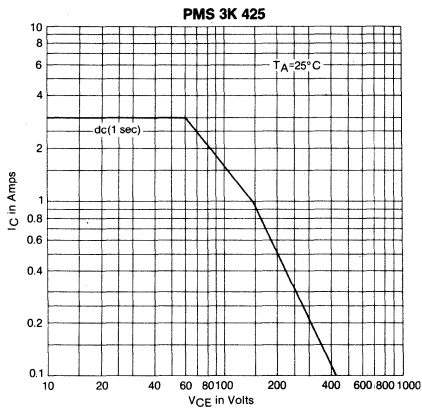


OPERATIONAL DATA

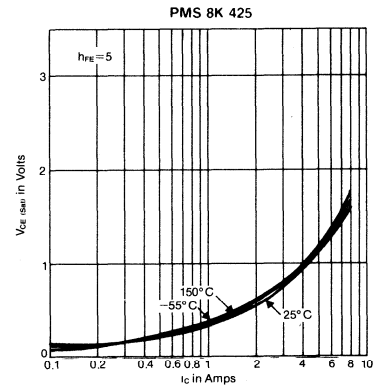
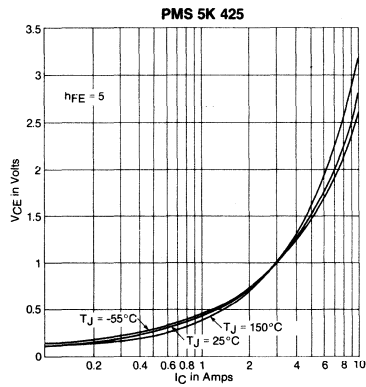
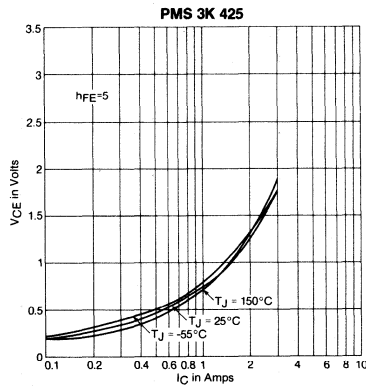
POWER DERATING



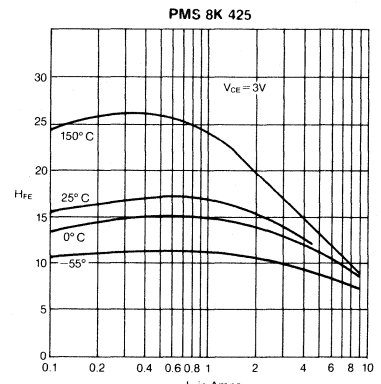
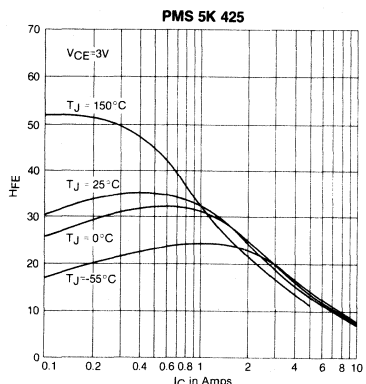
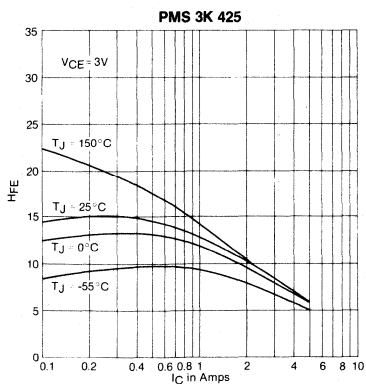
SAFE OPERATING AREA



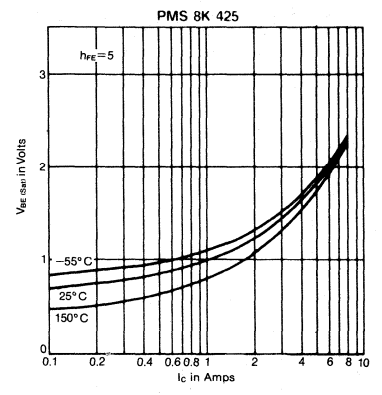
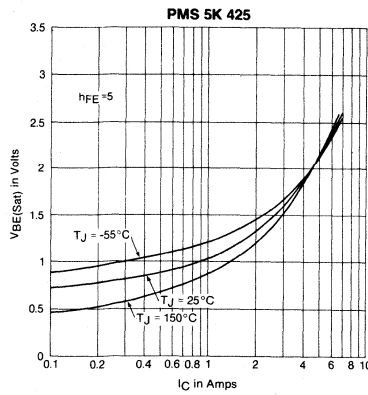
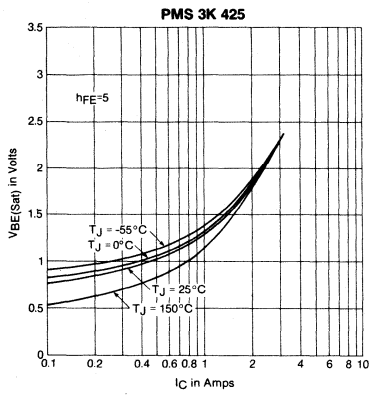
V_{CE} (Sat) VERSUS COLLECTOR CURRENT



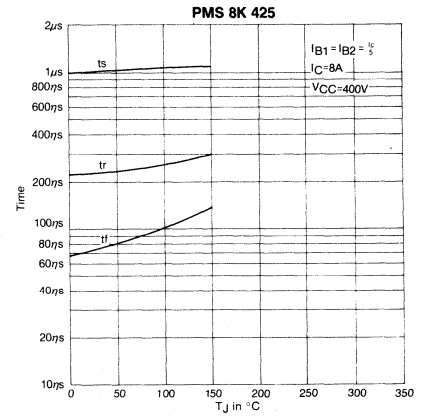
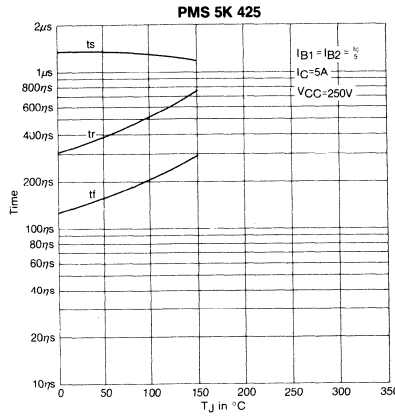
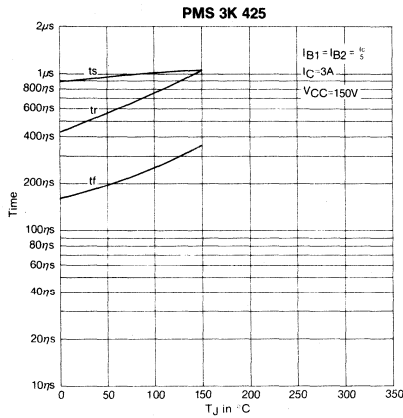
DC COLLECTOR CURRENT GAIN VERSUS COLLECTOR CURRENT



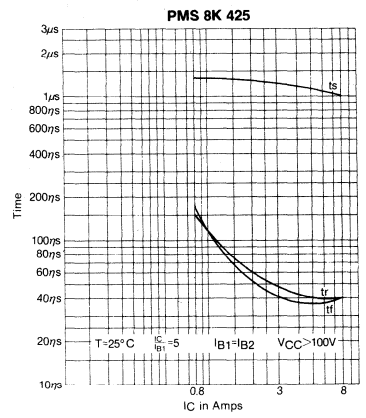
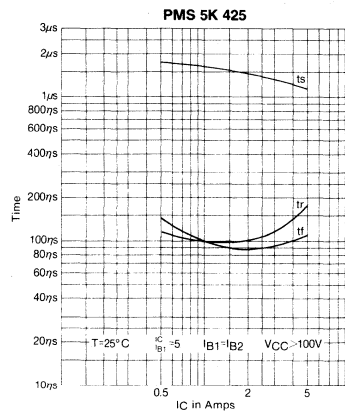
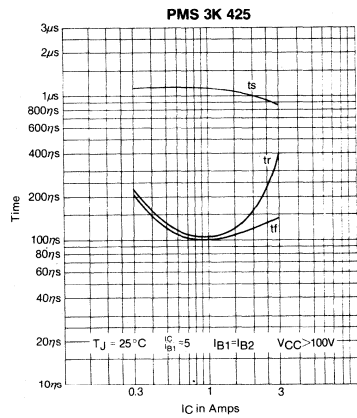
V_{BE} (Sat) VERSUS COLLECTOR CURRENT



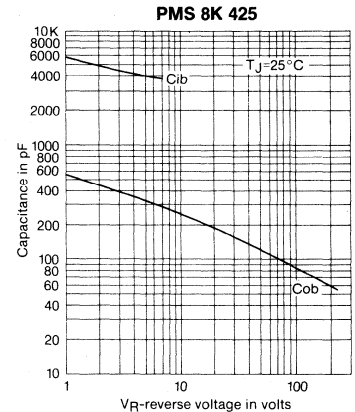
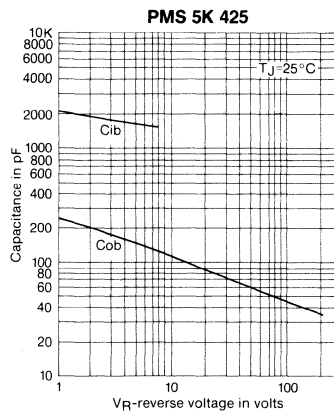
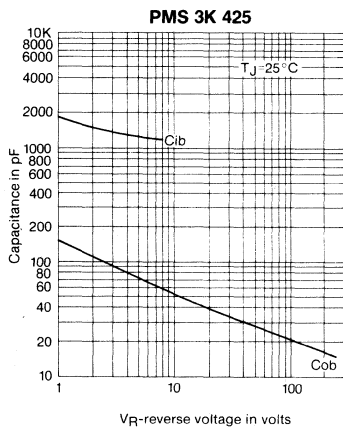
DYNAMIC CHARACTERISTICS VERSUS JUNCTION TEMPERATURE



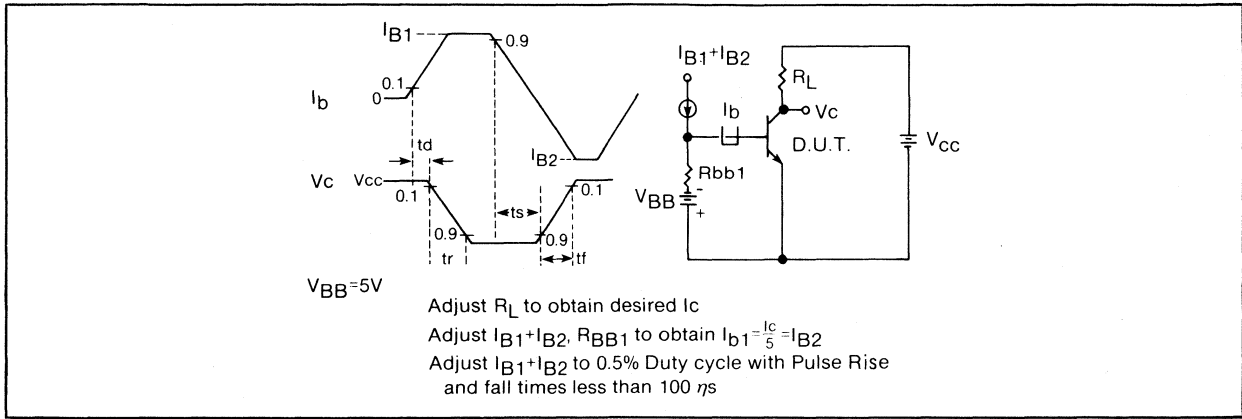
DYNAMIC CHARACTERISTICS VERSUS COLLECTOR CURRENT



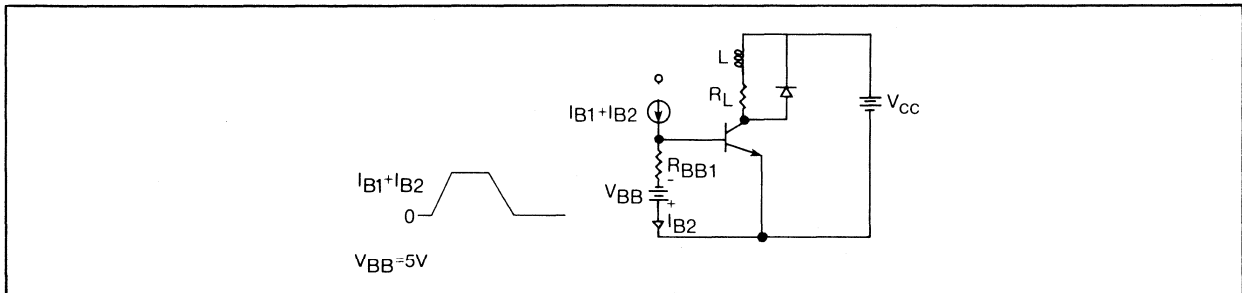
CAPACITANCE VERSUS REVERSE VOLTAGE



SWITCHING TEST CIRCUIT

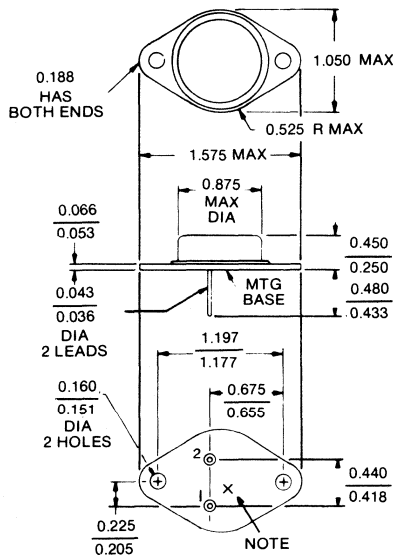


$E_{S/B}$ TEST CIRCUIT



PACKAGE OUTLINE

PMS 3K 425, PMS 5K 425

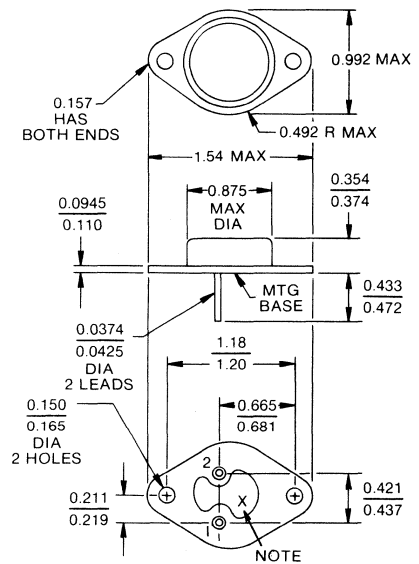


TO-3
3 TERMINAL (STEEL)

PIN	FUNCTION
1	BASE
2	EMITTER
CASE	COLLECTOR

NOTE (X): CASE TEMPERATURE MEASURED AT THIS POINT

PMS 8K 425



SECTION III – SWITCHING DARLINGTONS

△ LAMBDA PMD 20K & 25K SERIES

SWITCHING POWER DARLINGTONS

FEATURES

- Full CVD glass passivation on all Switching Darlingtons.
- Guaranteed h_{FE} greater than 300 at 10 amps.
- Guaranteed collector to emitter breakdown sustaining voltages of 200, 150, and 120 Volts for 5 and 10 amp Switching Darlingtons.
- Guaranteed and 100% tested for $I_{s/b}$ (Secondary Breakdown Current) insuring Maximum performance at high energy levels.
- Guaranteed secondary energy of 1.6 millijoules with a Baker Clamp Circuit.
- Excellent thermal resistance (θ_{JC}) on all NPN Switching Darlingtons providing for more usable power and lower operating temperatures.
- Built-in speed up diode for fast turn-off with negative base drive.
- All NPN Switching Darlingtons are hermetically sealed TO-3 packages and subject to fine and gross leak testing.

DESCRIPTION

The PMD-20K and PMD-25K series of devices are three terminal NPN silicon Switching Power Darlingtons. These monolithic devices are double-diffused epitaxial structures with built-in base-emitter shunt resistors, and they have been designed to switch at frequencies up to 30kHz. These devices are CVD glass passivated to improve reliability.

These darlingtons are designed to be used in power applications such as in Switching Regulators and Inverters, and can be driven directly by IC's with no negative base drive required for turn off. Internal diode protection (D1) of the darlington configuration is built into the structure to limit the device power dissipation during negative overshoot. Diode D2 is built-in to assist the reduction of device turn-off when negative base drive is used.

The two different series of Switching Power Darlington devices are available in sustaining voltages of 120, 150, and 200 volts at a power dissipation level of 150 watts. The devices feature high current gain (typically 1000 at $I_C = 10$ amps for the PMD-20K, and 1000 at $I_C = 5$ amps for the PMD-25K). All darlington devices are hermetically sealed in steel TO-3 packages, providing high reliability and low thermal resistance when used with appropriate heat sinks.

ABSOLUTE MAXIMUM RATINGS

RATING	SYMBOL	PMD 20K 200	PMD 25K 200	PMD 20K 150	PMD 25K 150	PMD 20K 120	PMD 25K 120	UNIT
Collector-Emitter Voltage	V_{CE0}	200		150		120		V_{dc}
Collector-Base Voltage	V_{CBO}	200		150		120		V_{dc}
Emitter-Base Voltage	V_{EBO}	2		2		2		V_{dc}
Collector Current Cont. Peak	I_C	14	9	14	9	14	9	A_{dc}
	I_C	20	12	20	12	20	12	A_{dc}
Base Current	I_B	0.5	0.2	0.5	0.2	0.5	0.2	A_{dc}
Total Internal Power Dissipation	$P_D @ T_C = 0^\circ C$	150		150		150		WATTS
	$> 0^\circ C$	Derate at 1 Watt/ $^\circ C$						
Temp. Range Oper. Storage	T_j T_{stg}	-65 to + 150 -65 to + 200						$^\circ C$ $^\circ C$
Thermal Resistance	$R_{\theta JC}$	1.0		1.0		1.0		$^\circ C/WATT$

ORDERING INFORMATION

DEVICE TYPE	VOLTAGE RATING	CURRENT RATING	QTY 1-24	QTY 25-49	QTY 50-99	QTY 100-249	QTY 250-499	QTY 500-999	QTY 1000-2499	QTY 2500-4999
PMD 20K 120	120V	14A	\$8.10	\$7.20	\$6.61	\$6.39	\$5.76	\$4.85	\$4.41	\$3.97
PMD 20K 150	150V	14A	9.00	8.00	7.35	7.10	6.40	5.39	4.90	4.41
PMD 20K 200	200V	14A	11.00	9.90	9.00	8.70	7.80	6.60	6.00	5.40
PMD 25K 120	120V	9A	5.40	4.80	4.45	4.32	3.87	3.27	2.97	2.67
PMD 25K 150	150V	9A	6.00	5.45	4.95	4.80	4.30	3.63	3.30	2.97
PMD 25K 200	200V	9A	7.20	6.60	6.00	5.80	5.20	4.40	4.00	3.60

Contact your area sales representative for higher quantity prices. Device configurations, specifications, and prices subject to change without notice.

PMD 20K, 25K SERIES

PERFORMANCE SPECIFICATIONS

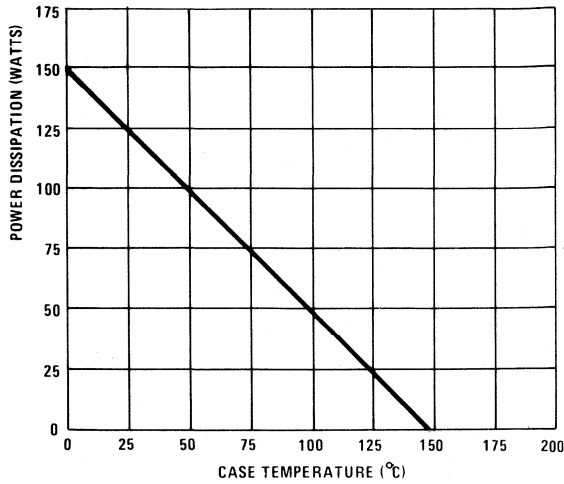
PMD-20K AND 25K SERIES
ALL PARAMETERS ARE GUARANTEED AT A T_J OF 0° C TO 150° C, UNLESS OTHERWISE NOTED

DEVICE TYPE	TEST	SYMBOL	TEST CONDITIONS		MIN.	MAX.	UNITS	NOTE
PMD-20K-SERIES PMD-25K-SERIES	Collector Emitter Saturation Voltage	V _{CE(Sat)}	I _C = 10Adc; I _B = 50MAdc I _C = 5Adc; I _B = 25MAdc			1.7	Vdc	1,2
PMD-20K-SERIES PMD-25K-SERIES	Base Emitter Turn-On Voltage	V _{BE(On)}	I _C = 10Adc; I _B = 50MAdc I _C = 5Adc; I _B = 25MAdc			1.8	Vdc	1
PMD-20K-SERIES PMD-25K-SERIES	Base Emitter Saturation Voltage	V _{BE(Sat)}	I _C = 10Adc; I _B = 50MAdc I _C = 5Adc; I _B = 25MAdc			2.6	Vdc	1,2
PMD-20K-SERIES PMD-25K-SERIES	DC Current Gain	h _{FE}	I _C = 10Adc; V _{CE} = 3Vdc I _C = 5Adc; V _{CE} = 3Vdc		300			1
PMD-20K-SERIES PMD-25K-SERIES	Forward Bias Second Breakdown Collector Current	I _{s,b}	V _{CE} = 26V 1 Second, non-repetitive pulse T _A = 25° C		5.8		Adc	1
PMD-20K-SERIES PMD-25K-SERIES	Second Breakdown Energy	E _{s,b} T _A = 25° C	I _C = 8.43Adc I _C = 6.7Adc	L = 45uH	1.6		Milli-Joule	
PMD-20K-120 PMD-25K-120	Collector Emitter Breakdown Voltage (Base Open)	V _{BRICEO}	I _C = 100MAdc I _B = 0		120		Vdc	1,2
PMD-20K-150 PMD-25K-150	Collector Emitter Sustaining Voltage	V _{BRICER(Sus)}	I _C = 100MAdc R _{BE} = 2.2K ohms		150		Vdc	1
PMD-20K-200 PMD-25K-200					200			
PMD-20K-120 PMD-25K-120					120			
PMD-20K-150 PMD-25K-150	Collector Emitter Leakage Current (Base Open)	I _{CEO}	V _{CE} = 80Vdc V _{CE} = 100Vdc V _{CE} = 150Vdc			0.5	MAdc	2
PMD-20K-200 PMD-25K-200								
PMD-20K-120 PMD-25K-120								
PMD-20K-150 PMD-25K-150	Collector Emitter Leakage Current	I _{CER}	V _{CE} = 80Vdc V _{CE} = 100Vdc V _{CE} = 150Vdc			5	MAdc	1
PMD-20K-200 PMD-25K-200								
PMD-20K-120 PMD-25K-120								
PMD-20K-SERIES PMD-25K-SERIES	Emitter Base Leakage Current	I _{EBO}	V _{EB} = 0.9Vdc; I _C = 0			70	MAdc	1
PMD-20K-SERIES PMD-25K-SERIES	Non-Saturated Switching Times Rise Time	t _r	I _C = 10Adc, V _{CC} = 30V I _C = 5Adc, V _{CC} = 30V			0.3	uS	2,3
PMD-20K-SERIES PMD-25K-SERIES	Turn-On Time	t _{on}	I _C = 10Adc, V _{CC} = 30V I _C = 5Adc, V _{CC} = 30V			0.5	uS	2,3
PMD-20K-SERIES PMD-25K-SERIES	Fall Time	t _{fall}	I _C = 10Adc, V _{CC} = 30V I _C = 5Adc, V _{CC} = 30V			0.8	uS	2,3
PMD-20K-SERIES PMD-25K-SERIES	Turn-Off Time	t _{off}	I _C = 10Adc, V _{CC} = 30V I _C = 5Adc, V _{CC} = 30V			1.0	uS	2,3

NOTES: (1) Pulse Tested: Pulse Width is less than or equal to 300 uSec. and Duty Cycle is less than or equal to 2.0%.
(2) T_J = 25° C
(3) See A.C. test circuit.

PMD 20K, 25K SERIES

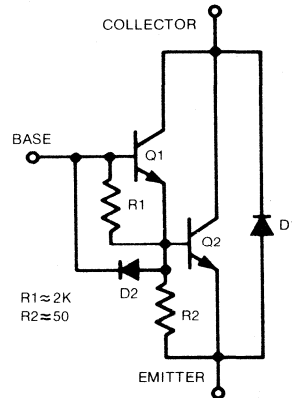
SWITCHING POWER DARLINGTONS



POWER DERATING

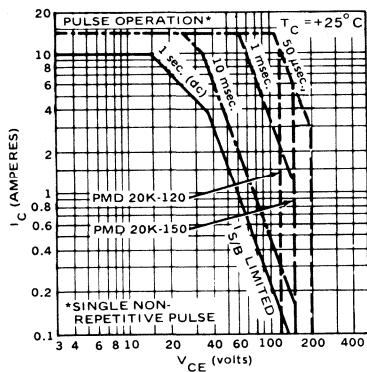
SCHEMATIC DIAGRAM

MONOLITHIC CONSTRUCTION

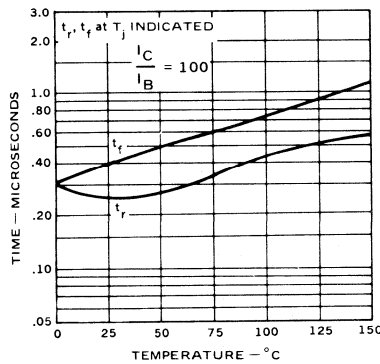


For higher efficiencies in DC-DC switching converters, an external fast recovery diode should be connected in parallel to D1.

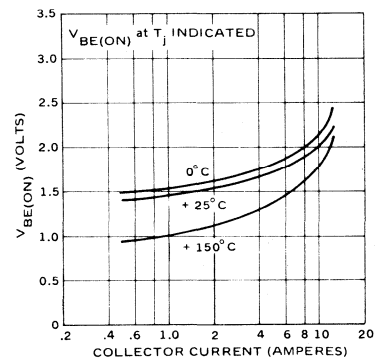
PMD 20K SERIES



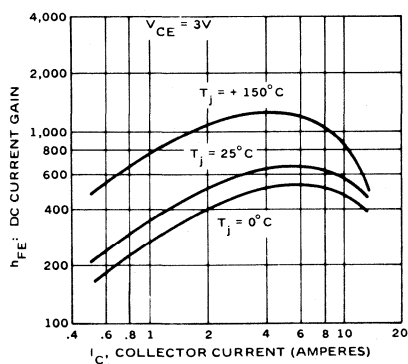
SAFE OPERATING AREA



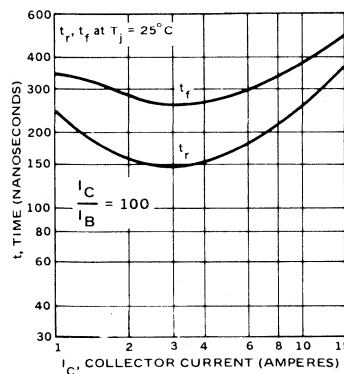
DYNAMIC CHARACTERISTICS VERSUS JUNCTION TEMPERATURE



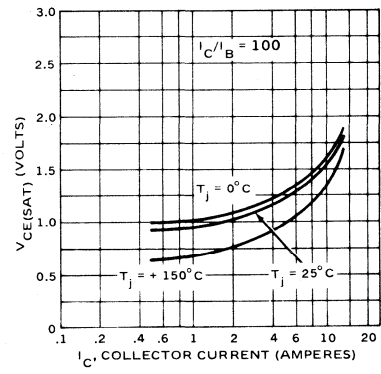
$V_{BE(ON)}$ VERSUS COLLECTOR CURRENT AND JUNCTION TEMP



DC CURRENT GAIN VERSUS COLLECTOR CURRENT



DYNAMIC CHARACTERISTICS VERSUS COLLECTOR CURRENT

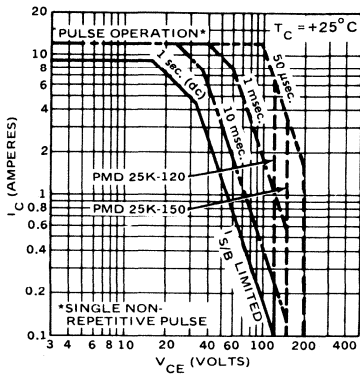


$V_{CE(SAT)}$ VERSUS COLLECTOR CURRENT AND JUNCTION TEMP

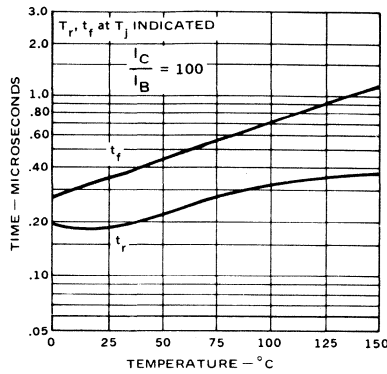
PMD 20K, 25K SERIES

SWITCHING POWER DARLINGTONS

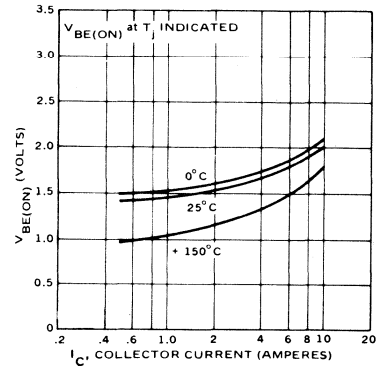
PMD 25K SERIES



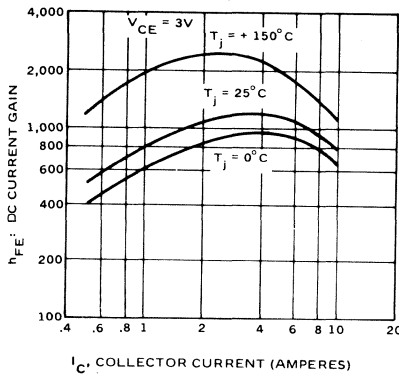
SAFE OPERATING AREA



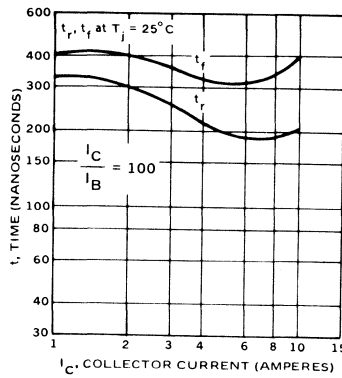
DYNAMIC CHARACTERISTICS VERSUS JUNCTION TEMPERATURE



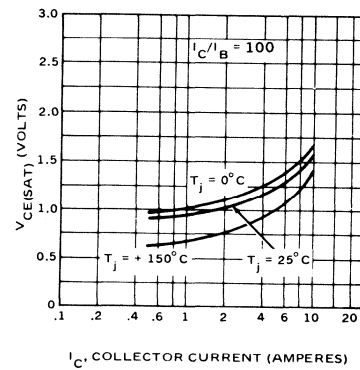
V_{BE(ON)} VERSUS COLLECTOR CURRENT AND JUNCTION TEMP



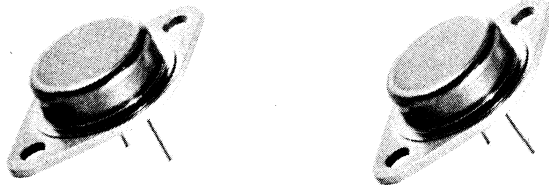
V_{CE(SAT)} VERSUS COLLECTOR CURRENT



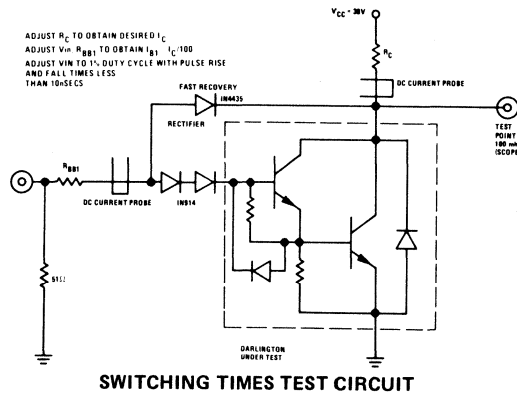
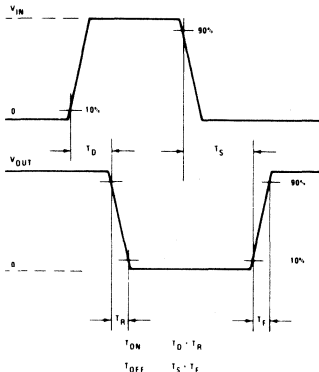
DYNAMIC CHARACTERISTICS VERSUS COLLECTOR CURRENT



DC CURRENT GAIN VERSUS COLLECTOR CURRENT



A.C. TEST CIRCUIT

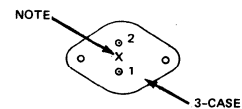
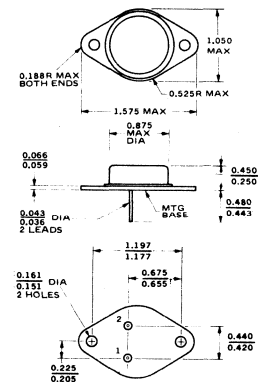


SWITCHING TIMES TEST CIRCUIT

Switching times are specified only under the condition that the device under test is not allowed to enter classical saturation. This condition is insured by the use of a Baker Clamp as shown above. A Baker Clamp insures that the base-collector junction of the D.U.T. is never forward biased, under worst case temperature and drive conditions.

OUTLINE DRAWING

PMD 20K, 25K SERIES TO-3 3-TERMINAL (STEEL)



PIN	FUNCTION
1	BASE
2	EMITTER
CASE	COLLECTOR

NOTE (X): Case temperature measured at this point.

PART II – LINEAR COMPONENTS

SECTION I – VOLTAGE REGULATORS

LAS 723, 723B SERIES

FEATURES

- The only low-noise, temperature-compensated diode reference voltage in the industry $2.5V \pm 5\%$.
- Guaranteed line regulation:
LAS-723: $0.020\% V_O/V$
LAS-723B: $0.008\% V_O/V$
- Guaranteed load regulation:
LAS-723: $0.05\% V_O @ 25mA$
LAS-723B: $0.05\% V_O @ 25mA$
- Guaranteed input-output differential: $+2.0$ volts
- Guaranteed temperature coefficient:
 $0^\circ C$ to $+125^\circ C$ — $0.015\% V_O/^\circ C$
- Guaranteed ripple attenuation:
LAS-723: 60dB
LAS-723B: 69dB
- Guaranteed thermal resistance, junction to ambient: $150^\circ C/W$

DESCRIPTION

The LAS-723 series voltage regulators are monolithic integrated circuits designed for use in applications requiring a well regulated positive output voltage. Outstanding features include full power usage up to 150 milliamperes of load variation, internal current limiting, and thermal shutdown on the chip under most operating conditions. Hermetically sealed TO-96 packages are employed for high reliability and low thermal resistance. A low-noise temperature stable diode reference is the key design factor insuring excellent temperature regulation of the LAS-723 series. This, coupled to a very low output impedance, insures superior performance and load regulation. A very low reference voltage of $+2.5$ volts $\pm 5\%$ compared to $+7.0$ volts $\pm 5\%$, which is standard for similar devices, allows the LAS-723 series a much greater output voltage range without the use of external components.

The LAS-723 series of ten-terminal regulators is available in an input voltage range from $+5$ to $+40$ volts, and an output voltage of $+2.63$ to $+38$ volts. The high voltage version of this device, the LAS-723B, is available in input voltage ranges from $+5$ to $+50$ volts, and output voltage ranges from $+2.63$ to $+48$ volts. Both devices offer low standby current drain, and high ripple rejection. When additional current capability is required, series NPN or PNP transistors may be added.

ABSOLUTE MAXIMUM RATINGS

PARAMETER	SYMBOL	MINIMUM	MAXIMUM	UNITS
Input Voltage LAS-723 LAS-723B	V_{IN}	0	$40^{(1)}$ $50^{(1)}$	Volts
Input/Output Differential LAS-723 LAS-723B	$V_{IN}-V_{OUT}$	0	$38^{(1)}$ $48^{(1)}$	Volts
Power Dissipation @ $T_c \leq 25^\circ C$	P_D		$0.8^{(1) (2)}$	Watts
Thermal Resistance Junction to Ambient	θ_{JA}		150	$^\circ C/Watt$
Operating Junction Temperature Range	T_J	0	150	$^\circ C$
Storage Temperature Range	T_{STG}	.65	150	$^\circ C$
Lead Temperature (Soldering, 60 Seconds Time Limit)	T_{LEAD}		300	$^\circ C$

⁽¹⁾ The maximum input voltage of the LAS-723 Series is limited by the maximum input-output differential, maximum power dissipation, and the maximum current limit-safe operating area, whichever is less.

⁽²⁾ For operation above $25^\circ C$ T_{CASE} derate @ 6.7 mW/ $^\circ C$.

REGULATOR PERFORMANCE SPECIFICATIONS – LAS 723, 723B

$V_{IN} = +V = 10$ VOLTS; $-V = 0$ VOLTS; $V_O = +5$ VOLTS; $I_O = 1$ MA; $R_{SC} = 0$; UNLESS OTHERWISE SPECIFIED.

PARAMETER	SYMBOL	TEST CONDITIONS			LAS 723		LAS 723B		UNITS
		V_{IN}	I_O	T_J	MIN.	MAX.	MIN.	MAX.	
Output Voltage	V_O	$V_{IN} = V_O + 2.0V$	25mA	$0-125^\circ C$	2.63 ⁽¹⁾	38	2.63 ⁽¹⁾	48	volts
Input-Output Differential	$V_{IN}-V_O$	$V_{IN} = V_O + 2.0V$	150mA	$0-125^\circ C$	2.0		2.0		volts
Output Current	I_O			$25^\circ C$		150		150	mA
Line Regulation ⁽²⁾	$REG_{(LINE)}$	$V_O = +3.0V$ $V_{IN} = +5$ to $+40V$		$0-125^\circ C$		0.020		0.008 ⁽²⁾	% $V_O/V^{(3)}$
Load Regulation ⁽²⁾	$REG_{(LOAD)}$		1 to 25mA 1 to 100mA	$0-125^\circ C$ $0-125^\circ C$		0.05 0.15		0.05 0.15	% V_O
Quiescent Current	I_Q	$V_{IN} = 40V$		$25^\circ C$		4.2		4.2 ⁽⁹⁾	mA
Reference Voltage	V_R			$25^\circ C$	2.375	2.625	2.375	2.625	volts
Reference Output Current	I_R	$V_O = 10V$		$25^\circ C$		3		3	mA
Temperature Coefficient	T_C			$0-125^\circ C$		0.015		0.015	% $V_O/^\circ C^{(4)}$
Ripple Attenuation	R_A	$V_{IN} = +10V + 1V_{rms}$		$0-125^\circ C$	60		69		dB
Output Noise Voltage ⁽⁵⁾	V_N			$0-125^\circ C$		50		50	μV_{rms}
Current Limit Sense Voltage ⁽⁶⁾	V_S			$25^\circ C$	0.62	0.74	0.62	0.73	volts
Error Amplifier Voltage Gain	A_V			$25^\circ C$	4000		4000		V/V

⁽¹⁾ $V_O = V_R (1 + R_1/R_2)$

R1 = Resistance from output to inverting input.
R2 = Resistance from inverting input to common.

⁽²⁾ Instantaneous measurement, average chip temperature changes must be accounted for separately.

⁽³⁾ Percentage times V_O ; per volt change in V_{IN} .

⁽⁴⁾ Percentage times V_O ; per $^\circ C$ change in temperature.

⁽⁵⁾ BW = 10Hz – 100KHz.

⁽⁶⁾ $R_{SC} = 1K\Omega$

⁽⁷⁾ $V_{IN} = +5$ to 50 volts

⁽⁸⁾ $V_{IN} = +10$ to 20 volts; $I_{OUT} = 25mA$

⁽⁹⁾ $V_{IN} = 50$ volts

ORDERING INFORMATION

MODEL	V_O VOLTS	I_O	PRICE QTY				
			1-99	100-249	250-999	1000-2499	2500-4999
LAS 723	2.63-38	150mA	\$1.15	\$1.04	\$.98	\$.92	\$.86
LAS 723B	2.63-48	150mA	1.19	1.08	1.02	.96	.90
LAS 1000	2.63-38	150mA	1.39	1.24	1.17	1.10	1.03
LAS 1100	2.63-48	150mA	1.64	1.43	1.40	1.33	1.23

LAS 1000, 1100 SERIES

150 mA POSITIVE REGULATORS FEATURES

- The only low-noise, temperature-compensated diode reference voltage in the industry $2.5V \pm 5\%$.
- Guaranteed line regulation:
LAS-1000: $0.020\% V_O/V$
LAS-1100: $0.008\% V_O/V$
- Guaranteed load regulation:
LAS-1000: $0.05\% V_O @ 25mA$
LAS-1100: $0.05\% V_O @ 25mA$
- Guaranteed input-output differential: +2.0 volts
- Guaranteed temperature coefficient:
 $0^\circ C$ to $+125^\circ C$ — $0.015\% V_O/^\circ C$
- Guaranteed ripple attenuation:
LAS-1000: 60dB
LAS-1100: 69dB
- Guaranteed thermal resistance, junction to ambient: $150^\circ C/W$

DESCRIPTION

The LAS-1000 series voltage regulators are monolithic integrated circuits designed for use in applications requiring a well regulated positive output voltage. Outstanding features include full power usage up to 150 milliamperes of load variation, internal current limiting, and thermal shutdown on the chip under most operating conditions. Hermetically sealed TO-96 packages are employed for high reliability and low thermal resistance. A low-noise temperature stable diode reference is the key design factor insuring excellent temperature regulation of the LAS-1000 series. This, coupled to a very low output impedance, insures superior performance and load regulation. A very low reference voltage of $+2.5$ volts $\pm 5\%$ compared to $+7.0$ volts $\pm 5\%$, which is standard for similar devices allows the LAS-1000 series a much greater output voltage range without the use of external components. The addition of remote electronic/shutdown on this device type allows the designer a great deal of flexibility in design and applications.

The LAS-1000 series of ten-terminal regulators is available in an input voltage range from $+5$ to $+40$ volts, and an output voltage of $+2.63$ to $+38$ volts. The high voltage version of this device, the LAS-1100, is available in input voltage ranges from $+5$ to $+50$ volts, and output voltage ranges from $+2.63$ to $+48$ volts. Both devices offer low standby current drain, and high ripple rejection. When additional current capability is required, series NPN or PNP transistors may be added.

ABSOLUTE MAXIMUM RATINGS

PARAMETER	SYMBOL	MINIMUM	MAXIMUM	UNITS
Input Voltage LAS-1000 LAS-1100	V_{IN}	0 0	$40^{(1)}$ $50^{(1)}$	Volts
Input/Output Differential LAS-1000 LAS-1100	$V_{IN}-V_{OUT}$	0 0	$38^{(1)}$ $48^{(1)}$	Volts
Power Dissipation @ $T_c \leq 25^\circ C$	P_D		$0.8^{(1)(2)}$	Watts
Thermal Resistance Junction to Ambient	θ_{JA}		150	$^\circ C/Watt$
Operating Junction Temperature Range	T_J	0	150	$^\circ C$
Storage Temperature Range	T_{STG}	-65	150	$^\circ C$
Lead Temperature (Soldering, 60 Seconds Time Limit)	T_{LEAD}		300	$^\circ C$

⁽¹⁾ The maximum input voltage of the LAS-1000 Series is limited by the maximum input-output differential, maximum power dissipation, and the maximum current limit-safe operating area, whichever is less.

⁽²⁾ For operation above $25^\circ C$ T_{CASE} , derate @ 6.7 mW/ $^\circ C$.

REGULATOR PERFORMANCE SPECIFICATIONS – LAS 1000, 1100

$V_{IN} = +V = 10$ VOLTS; $-V = 0$ VOLTS; $V_O = +5$ VOLTS; $I_O = 1mA$; $R_{SC} = 0$; UNLESS OTHERWISE SPECIFIED.

LAS 1000, 1100 PARAMETER	SYMBOL	TEST CONDITIONS			LAS-1000 TEST LIMITS		LAS-1100 TEST LIMITS		UNITS
		V_{IN}	I_O	T_J	MIN.	MAX.	MIN.	MAX.	
Output Voltage	V_O	$V_O + 2.0$ volts	25mA	$0-125^\circ C$	2.63 ⁽¹⁾	38	2.63 ⁽¹⁾	48	volts
Input-Output Differential	$V_{IN}-V_O$	$V_O + 2.0$ volts	150mA	$0-125^\circ C$	2.0		2.0		volts
Output Current	I_O			$25^\circ C$		150		150	mA
Line Regulation ⁽²⁾	$REG_{(LINE)}$	$V_O + 3.0$ volts		$0-125^\circ C$		0.020		0.008 ⁽⁷⁾	$\% V_O/V^{(3)}$
Load Regulation ⁽²⁾	$REG_{(LOAD)}$	$V_{IN} = +5$ to $+40V$	1 to 25mA	$0-125^\circ C$		0.05		0.02 ⁽⁸⁾	$\% V_O/V^{(3)}$
			1 to 100mA			0.15		0.15	$\% V_O$
Quiescent Current	I_Q	40 volts		$25^\circ C$		4.2		4.2 ⁽⁹⁾	mA
Reference Voltage	V_R			$25^\circ C$	2.375	2.625	2.375	2.625	volts
Reference Output Current	I_R	$V_O = 10$ volts		$25^\circ C$		3		3	mA
Temperature Coefficient	T_C			$0-125^\circ C$		0.015		0.015	$\% V_O/^\circ C^{(4)}$
Ripple Attenuation	R_A	10 volts + 1Vrms		$0-125^\circ C$	60		69		dB
Output Noise Voltage ⁽⁵⁾	V_N			$0-125^\circ C$		50		50	$\mu Vrms$
Current Limit Sense Voltage ⁽⁶⁾	V_S			$25^\circ C$	0.060	0.110	0.060	0.110	volts
Error Amplifier Voltage Gain	A_V			$25^\circ C$	4000		4000		V/V
Shutdown Voltage Threshold	V			$25^\circ C$	0.65	0.75	0.65	0.75	volts
Shutdown Source Current	I			$25^\circ C$	140	260	140	260	μA
Current Limit Source Current	I			$25^\circ C$	70	130	70	130	μA

⁽¹⁾ $V_O = V_R (1 + R_1/R_2)$

R1 = Resistance from output to inverting input.

R2 = Resistance from inverting input to common.

⁽²⁾ Instantaneous measurement, average chip temperature changes must be accounted for separately.

⁽³⁾ Percentage times V_O ; per volt change in V_{IN} .

⁽⁴⁾ Percentage times V_O ; per $^\circ C$ change in temperature.

⁽⁵⁾ BW = 10Hz – 100KHz.

⁽⁶⁾ $R_{SC} = 100\Omega$

⁽⁷⁾ $V_{IN} = +5$ to $+50$ volts

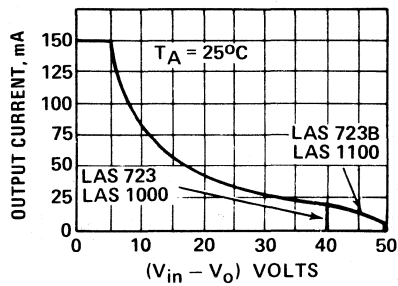
⁽⁸⁾ $V_{IN} = +10$ to $+20$ volts; $I_{OUT} = 25mA$

⁽⁹⁾ $V_{IN} = +50$ volts

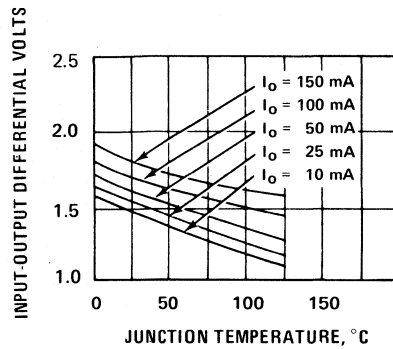
LAS 723, 723B, 1000, 1100 SERIES

150 mA POSITIVE REGULATORS

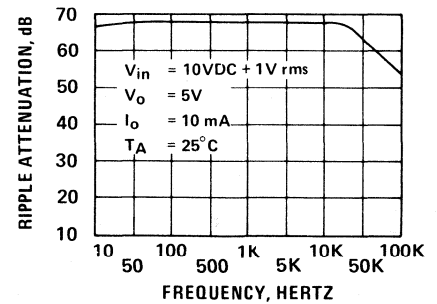
OPERATIONAL DATA



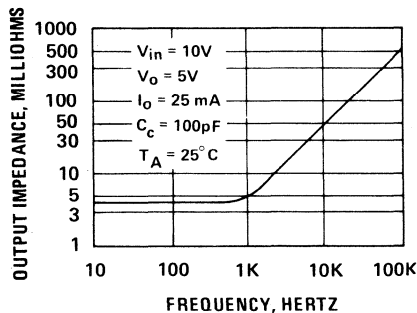
OUTPUT CURRENT VERSUS INPUT-OUTPUT VOLTAGE DIFFERENTIAL



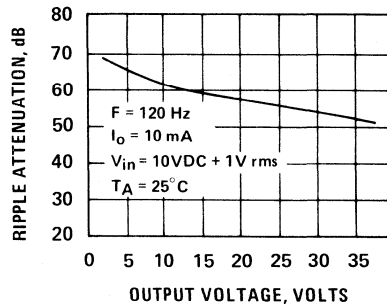
INPUT-OUTPUT DIFFERENTIAL VERSUS JUNCTION TEMPERATURES



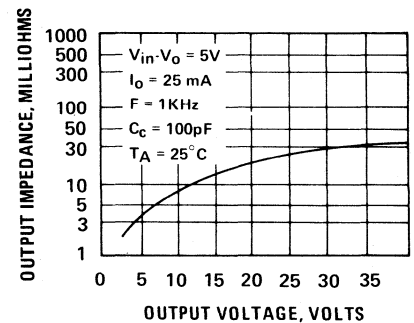
TYPICAL RIPPLE ATTENUATION VS FREQUENCY



OUTPUT IMPEDANCE VERSUS FREQUENCY

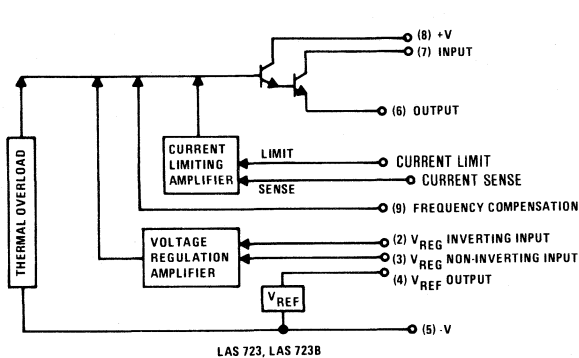


RIPLLE ATTENUATION VERSUS OUTPUT VOLTAGE

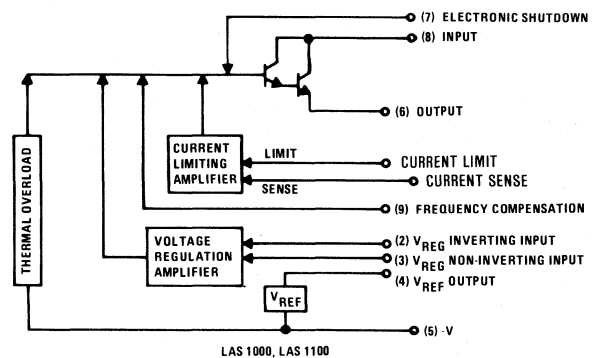


OUTPUT IMPEDANCE VERSUS OUTPUT VOLTAGE

FUNCTIONAL BLOCK DIAGRAMS

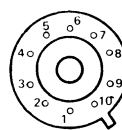
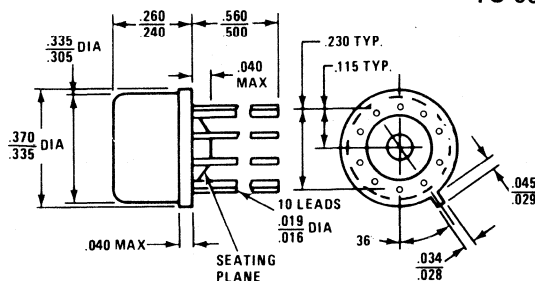


LAS 723, LAS 723B



LAS 1000, LAS 1100

TO-96 10-TERMINAL



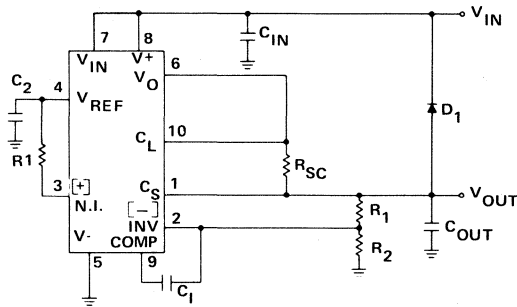
Bottom View

Pin	Function
1	Current Sense
2	Inverting Input
3	Non-Inverting Input
4	VREF
5	-V
6	VOUT
7	VIN (LAS 723), Electronic Shutdown (LAS 1000)
8	+V (LAS 723), Input (LAS 1000)
9	Frequency Compensation
10	Current Limit

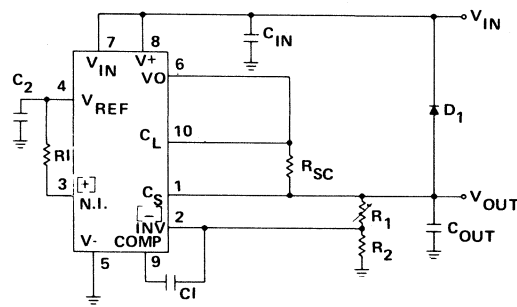
LAS 723, 723B, 1000, 1100 SERIES

TYPICAL APPLICATIONS

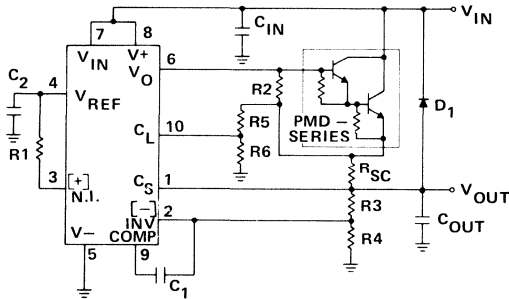
LAS 723, 723B



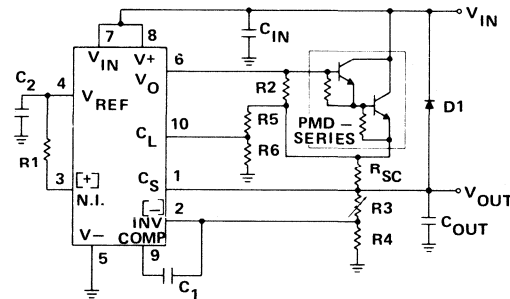
BASIC FIXED POSITIVE REGULATOR, LAS 723, LAS 723B



BASIC ADJUSTABLE POSITIVE REGULATOR, LAS 723, LAS 723B

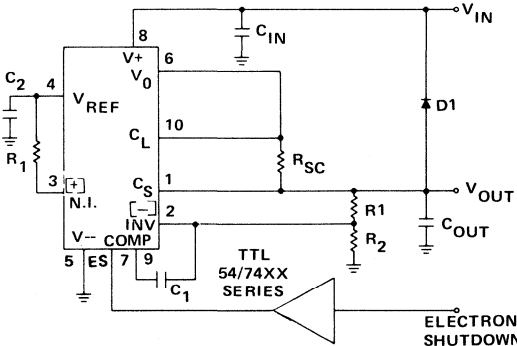


BASIC FIXED POSITIVE REGULATOR WITH AN EXTERNAL NPN PASS DARLINGTON FOR HIGH-CURRENT APPLICATIONS, LAS 723, LAS 723B

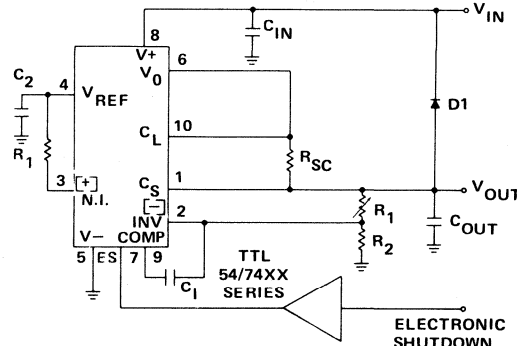


BASIC ADJUSTABLE POSITIVE REGULATOR WITH AN EXTERNAL NPN PASS DARLINGTON FOR HIGH-CURRENT APPLICATIONS, LAS 723, LAS 723B

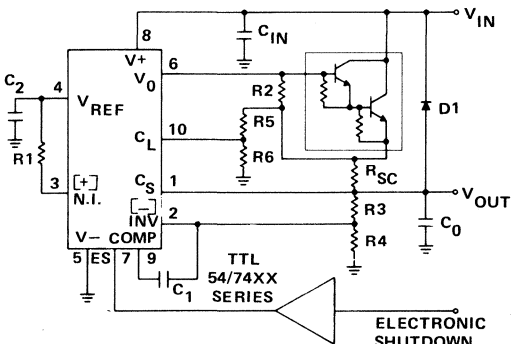
LAS 1000, 1100



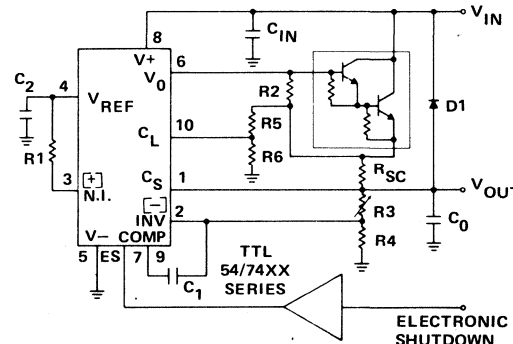
BASIC FIXED POSITIVE REGULATOR WITH ELECTRONIC SHUTDOWN LAS 1000, LAS 1100



BASIC ADJUSTABLE POSITIVE REGULATOR WITH ELECTRONIC SHUTDOWN LAS 1000, LAS 1100



BASIC FIXED POSITIVE REGULATOR WITH AN EXTERNAL NPN PASS DARLINGTON FOR HIGH-CURRENT APPLICATIONS WITH ELECTRONIC SHUTDOWN LAS 1000, LAS 1100



BASIC ADJUSTABLE POSITIVE REGULATOR WITH AN EXTERNAL NPN PASS DARLINGTON FOR HIGH-CURRENT APPLICATIONS WITH ELECTRONIC SHUTDOWN LAS 1000, LAS 1100

LAS 1500, 15U, 15A00 SERIES

1.5 AMP POSITIVE REGULATORS

FEATURES

- **Guaranteed input-output differential:**
+2.4 volts (LAS 1500 series)
- **Guaranteed line regulation:**
2.0% V_{OUT} @ $\Delta V_{IN} = 10$ volts
(LAS 1500 series)
- **Guaranteed temperature coefficient:**
0 to 125°C — 0.03% $V_{OUT}/^{\circ}C$
- **Guaranteed load regulation:**
0.6% V_{OUT} @ 10mA to 1.5 amps
(LAS 1500 series)
- **Guaranteed ripple attenuation:**
58dB (LAS 1500 series)
- **Guaranteed thermal resistance, junction to case:** 3°C/W

DESCRIPTION

The LAS-1500 series voltage regulators are monolithic integrated circuits designed for use in applications requiring a well regulated positive output voltage. Outstanding features include full power usage up to 1.5 amperes of load current, internal current limiting, thermal shutdown, and safe area protection on the chip, providing protection of the series pass darlington, under most operating conditions. Hermetically sealed steel TO-3 packages are utilized for high reliability and low thermal resistance when used with an appropriate heat sink. A low-noise temperature stable diode reference is the key design factor insuring excellent temperature regulation of the LAS-1500 series. This coupled to a very low output impedance insures superior performance and load regulation.

The LAS-1500 series of three-terminal regulators are available in fixed output voltage tolerances of $\pm 5\%$ with nominal output voltages ranging from +5 to +28 volts. The LAS-15A00 three-terminal regulators are available in fixed output voltage tolerances of $\pm 2\%$ with nominal output voltages available of +5, +12, and +15 volts. The LAS-15U, a four-terminal adjustable regulator, is available with an output range from +4 to +30 volts, adjustable with a single potentiometer.

ABSOLUTE MAXIMUM RATINGS

PARAMETER	SYMBOL	MINIMUM	MAXIMUM	UNITS
Input Voltage	V_{IN}	0	35 (40) ⁽¹⁾	Volts
Input/Output Differential	$V_{IN}-V_{OUT}$	0	30 ⁽¹⁾	Volts
Power Dissipation @ $T_c \leq 105^{\circ}C$	P_D		15 ^(1), 2)	Watts
Thermal Resistance Junction to Case	θ_{JC}		3	$^{\circ}C/Watt$
Operating Junction Temperature Range	T_J	-55	150	$^{\circ}C$
Storage Temperature Range	T_{STG}	-65	150	$^{\circ}C$
Lead Temperature (Soldering, 60 Seconds Time Limit)	T_{LEAD}		300	$^{\circ}C$

⁽¹⁾ The maximum input voltage of the LAS-1500 Series is limited by the maximum input-output differential, maximum power dissipation, or the maximum current limit-safe operating area, whichever is less. Values of +35V apply to V_O of +4V to +12V. Values of +40V apply to V_O of 13.8V to +28V.

⁽²⁾ For operation above 105°C T_{CASE} , derate @ 333 mW/ $^{\circ}C$.

REGULATOR PERFORMANCE SPECIFICATIONS – LAS 1500, 15U

INPUT VOLTAGE TEST CONDITIONS ARE AS FOLLOWS: $V_1 = V_O + 5$ VOLTS, $V_2 = V_O + 15$ VOLTS, $V_3 = V_O + 20$ VOLTS, OR THE MAXIMUM INPUT WHICHEVER IS LESS.

PARAMETER	SYMBOL	TEST CONDITIONS			LAS-1500 TEST LIMITS		LAS-15U TEST LIMITS		UNITS
		V_{IN}	I_O	T_J	MIN.	MAX.	MIN.	MAX.	
Output Voltage	V_O	V_1 to V_2	10mA to 1.0A	25°C	0.95 V_O ⁽¹⁾	1.05/ V_O	4.0 ⁽⁶⁾	30	volts
Input-Output Differential	$V_{IN}-V_O$		1.0A	0-125°C	2.4		2.4		volts
Output Current	I_O	V_1		25°C	10mA	1.5	10mA	1.5	amps
Line Regulation ⁽²⁾	$REG_{(LINE)}$	V_1 to V_2	1.0A	25°C		2.0		2.0	% V_O
Load Regulation ⁽²⁾	$REG_{(LOAD)}$	V_1	10mA to 1.5A	25°C		0.6		0.6	% V_O
Quiescent Current	I_Q	V_1	10mA	25°C		10		10	mA
Quiescent Current Line	$I_{Q(LINE)}$	V_1 to V_3 ⁽⁵⁾	10mA	25°C		3.0		3.0	mA
Quiescent Current Load	$I_{Q(LOAD)}$	V_1	10mA to 1.5A	25°C		3.0		3.0	mA
Current Limit	I_{LIM}	V_1		25°C		3.5		3.5	amps
Short Circuit Current	I_S	V_1		25°C		3.5		3.5	amps
Temperature Coefficient	T_C	V_1	0.1A	0-125°C		0.03		0.03	% $V_O/^{\circ}C$
Output Noise Voltage	V_N	V_1	0.1A	0-125°C		10 ⁽³⁾		10 ⁽³⁾	$\mu V_{rms}/V$
Ripple Attenuation	R_A	V_1	1.0A	0-125°C		58 ⁽⁴⁾		58 ⁽⁴⁾	dB
Control Voltage	V_C	V_1 to V_2	10mA	25°C			3.5	4.0	volts

⁽¹⁾ Nominal output voltages are specified under ordering information.

⁽²⁾ Instantaneous regulation, average chip temperature changes must be accounted for separately.

⁽³⁾ BW = 10Hz – 100KHz.

⁽⁴⁾ Ripple attenuation is specified for a 1Vrms, 120 Hz input ripple. Ripple attenuation is a minimum of 58dB at a 5 volt output, and is 1dB less for each volt increase in the output voltage.

⁽⁵⁾ The maximum input voltage of the LAS-1500 series is limited by maximum input-output differential voltage, maximum power dissipation, or the current limit-SOA, whichever is less.

⁽⁶⁾ $V_O = V_C (1 + R1/R2)$

R1 = Resistance from output to control.

R2 = Resistance from control to common.

LAS 1500, 15U, 15A00 SERIES

1.5 AMP POSITIVE REGULATORS

REGULATOR PERFORMANCE SPECIFICATIONS – LAS 15A00

INPUT VOLTAGE TEST CONDITIONS ARE AS FOLLOWS: $V_1 = V_0 + 5$ VOLTS, $V_2 = V_0 + 15$ VOLTS, $V_3 = V_0 + 20$ VOLTS, OR THE MAXIMUM INPUT WHICHEVER IS LESS.

PARAMETER	SYMBOL	TEST CONDITIONS			LAS-15A00 TEST LIMITS		UNITS
		V_{IN}	I_o	T_J	MIN.	MAX.	
Output Voltage	V_o	V_1 to V_2	10mA to 1.0A	25° C	$0.98 V_o ^{(1)}$	$1.02 V_o $	volts
Input-Output Differential	$V_{IN} - V_o$		1.0A	0-125° C	2.4		volts
Output Current	I_o	V_1		25° C	10	1.5	mA
Line Regulation ⁽²⁾	$REG_{(LINE)}$	V_1 to V_2	1.0A	25° C		2.0	% V_o
Load Regulation ⁽²⁾	$REG_{(LOAD)}$	V_1	10mA to 1.5A	25° C		0.6	% V_o
Quiescent Current	I_Q	V_1	10mA	25° C		10	mA
Quiescent Current Line	$I_Q (LINE)$	V_1 to V_3	10mA	25° C		3.0	mA
Quiescent Current Load	$I_Q (LOAD)$	V_1	10mA to 1.5A	25° C		3.0	mA
Current Limit	I_{LIM}	V_1		25° C		3.5	amps
Short Circuit Current	I_S	V_1		25° C		3.5	amps
Temperature Coefficient	T_C	V_1	0.1A	0-125° C		0.03	% $V_o/°C$
Output Noise Voltage	V_N	V_1	0.1A	0-125° C		$10^{(3)}$	$\mu V_{rms}/V$
Ripple Attenuation	R_A	V_1	1.0A	0-125° C		$58^{(4)}$	dB

⁽¹⁾ Nominal output voltages are specified under ordering information.

⁽²⁾ Instantaneous regulation, average chip temperature changes must be accounted for separately.

⁽³⁾ BW = 10Hz – 100KHz.

⁽⁴⁾ Ripple attenuation is specified for a 1Vrms, 120 Hz input ripple.

Ripple attenuation is a minimum of 58dB at a 5 volt output, and is 1dB less for each volt increase in the output voltage.

⁽⁵⁾ The maximum input voltage of the LAS-1500 series is limited by maximum input-output differential voltage, maximum power dissipation, or the current limiting-SOA, whichever is less.

ORDERING INFORMATION

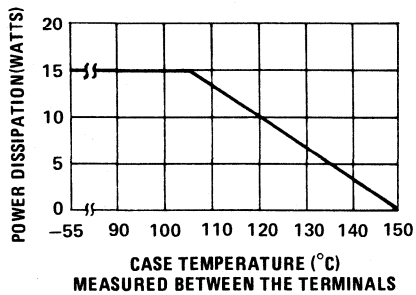
LAS 1500 SERIES 1.5 AMPS, 15 WATTS

MODEL	NOMINAL V_o VOLTS	PRICE QTY							
		1-24	25-49	50-99	100-249	250-499	500-999	1000-2499	2500-4999
LAS 1505	5	\$2.05	\$1.88	\$1.73	\$1.61	\$1.50	\$1.43	\$1.35	\$1.28
LAS 15A05	5	2.75	2.50	2.30	2.15	2.00	1.90	1.80	1.70
LAS 1506	6	2.05	1.88	1.73	1.61	1.50	1.43	1.35	1.28
LAS 1508	8	2.05	1.88	1.73	1.61	1.50	1.43	1.35	1.28
LAS 1510	10	2.05	1.88	1.73	1.61	1.50	1.43	1.35	1.28
LAS 1512	12	2.05	1.88	1.73	1.61	1.50	1.43	1.35	1.28
LAS 15A12	12	2.75	2.50	2.30	2.15	2.00	1.90	1.80	1.70
LAS 15CB	13.8	2.05	1.88	1.73	1.61	1.50	1.43	1.35	1.28
LAS 1515	15	2.05	1.88	1.73	1.61	1.50	1.43	1.35	1.28
LAS 15A15	15	2.75	2.50	2.30	2.15	2.00	1.90	1.80	1.70
LAS 1518	18	2.05	1.88	1.73	1.61	1.50	1.43	1.35	1.28
LAS 1520	20	2.05	1.88	1.73	1.61	1.50	1.43	1.35	1.28
LAS 1524	24	2.05	1.88	1.73	1.61	1.50	1.43	1.35	1.28
LAS 1528	28	2.05	1.88	1.73	1.61	1.50	1.43	1.35	1.28
LAS 15U	4 to 30	4.25	3.86	3.55	3.33	3.09	2.94	2.78	2.63

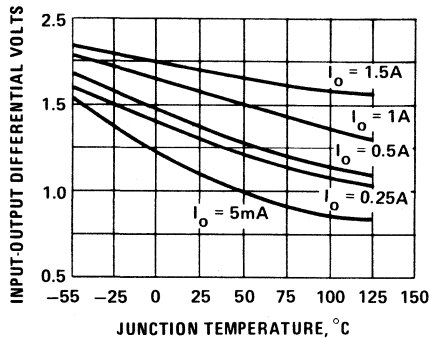
LAS 1500, 15U, 15A00 SERIES

1.5 AMP POSITIVE REGULATORS

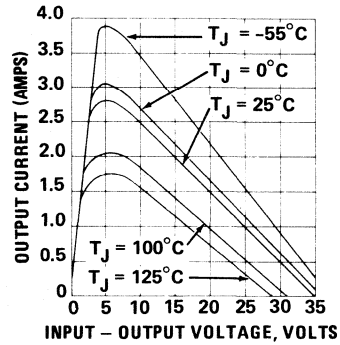
OPERATIONAL DATA



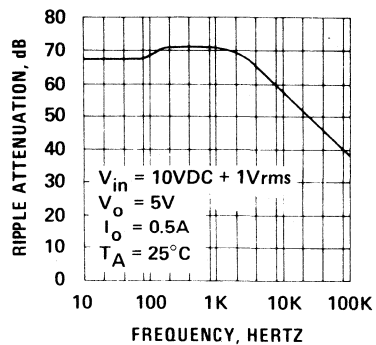
POWER DERATING



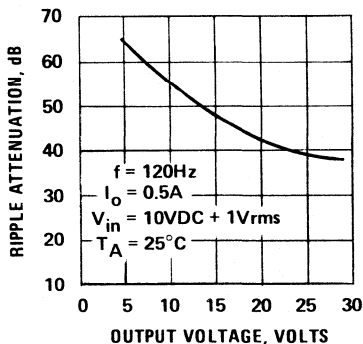
TYPICAL MINIMUM INPUT-OUTPUT DIFFERENTIAL VOLTAGE VS JUNCTION TEMPERATURE



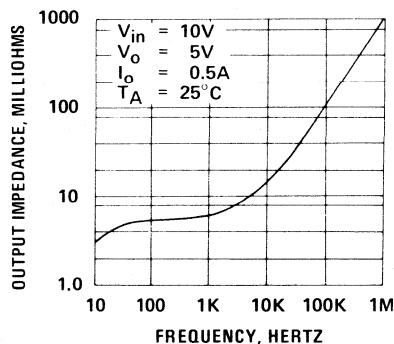
TYPICAL CURRENT LIMIT VS INPUT-OUTPUT VOLT. DIFF.



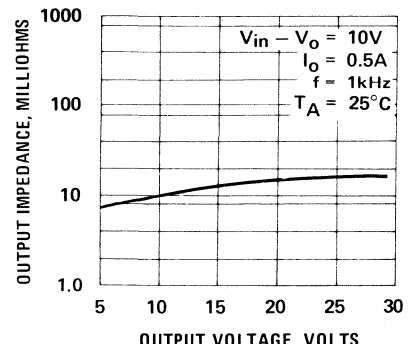
TYPICAL RIPPLE ATTENUATION VS FREQUENCY



TYPICAL RIPPLE ATTENUATION VS OUTPUT VOLTAGE

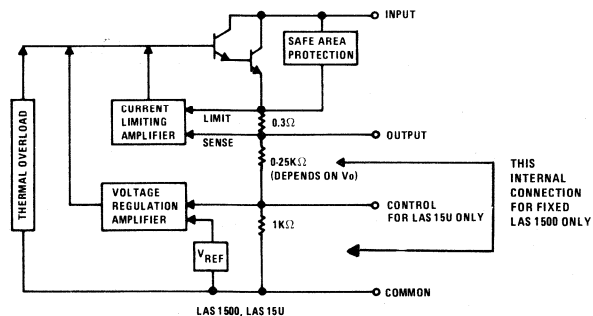


TYPICAL OUTPUT IMPEDANCE VS FREQUENCY



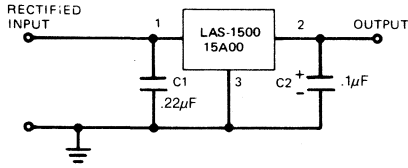
TYPICAL OUTPUT IMPEDANCE VS OUTPUT VOLTAGE

FUNCTIONAL BLOCK DIAGRAM

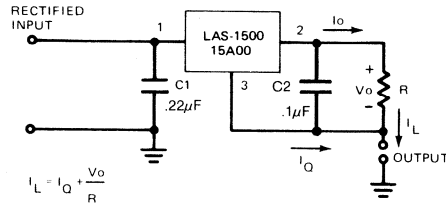


LAS 1500, 15U, 15A00 SERIES

TYPICAL APPLICATIONS



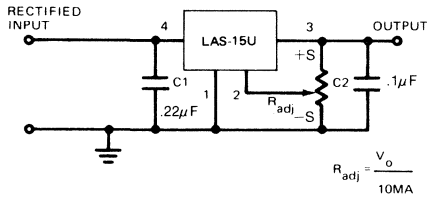
C1 AND C2 TO BE PLACED AS CLOSE TO THE DEVICE AS POSSIBLE.
 FILTER CAPACITOR = 2000μF/AMP



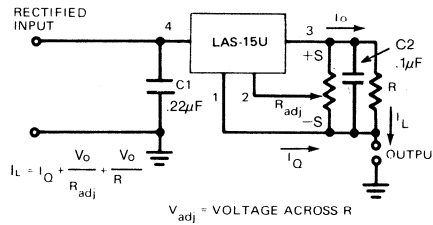
C1 AND C2 TO BE PLACED AS CLOSE TO THE DEVICE AS POSSIBLE.
 FILTER CAPACITOR = 2000μF/AMP.

1.5 AMP POSITIVE REGULATOR CIRCUIT

1.5 AMP FIXED CURRENT REGULATOR



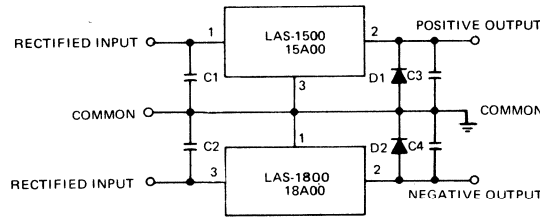
PIN 1 IS CONNECTED TO CASE.
 C1 AND C2 TO BE PLACED AS CLOSE TO THE DEVICE AS POSSIBLE.
 FILTER CAPACITOR = 2000μF/AMP.



PIN 1 IS CONNECTED TO CASE.
 C1 TO BE PLACED AS CLOSE TO THE DEVICE AS POSSIBLE.
 FILTER CAPACITOR = 2000μF/AMP.

1.5 AMP POSITIVE ADJUSTABLE REGULATOR

1.5 AMP POSITIVE ADJUSTABLE CURRENT REGULATOR

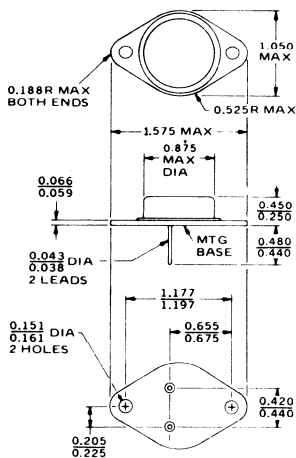


DIODES D1 & D2 SHOULD BE LOW V_F , HIGH CURRENT DIODES FOR OPTIMUM PROTECTION.
 C1, C2, C3 AND C4 TO BE PLACED AS CLOSE TO DEVICE AS POSSIBLE.
 FILTER CAPACITOR = 2000 μ F/AMP

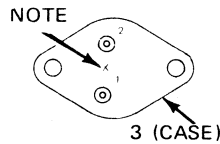
1.5 AMP FIXED DUAL REGULATOR

OUTLINE DRAWING

TO-3 3-TERMINAL (STEEL)



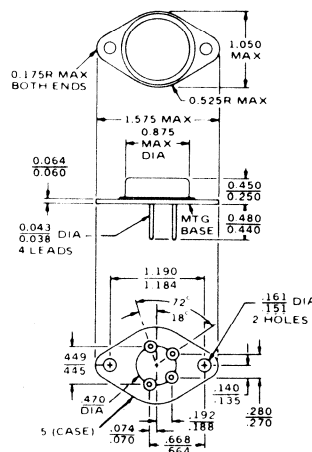
LAS-1500
 LAS-15A00



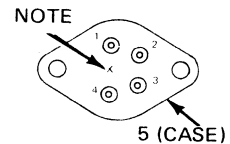
PIN	FUNCTION
1	INPUT
2	OUTPUT
3	COMMON

NOTE (X) = CASE TEMPERATURE MEASURED AT THIS POINT.

TO-3 4-TERMINAL (STEEL)



LAS-15U



PIN	FUNCTION
1	COMMON (ELEC)
2	CONTROL
3	OUTPUT
4	INPUT
5	COMMON

NOTE (X) = CASE TEMPERATURE MEASURED AT THIS POINT.

LAS 1800, 18U, 18A00 SERIES

1.5 AMP NEGATIVE REGULATORS

FEATURES

- **Guaranteed input-output differential:**
+2.1 volts (LAS 1800 series)
- **Guaranteed line regulation:**
2.0% V_{OUT} @ $\Delta V_{IN} = -15$ volts
(LAS 1800 series)
- **Guaranteed temperature coefficient:**
0 to 125°C — 0.03% $V_{OUT}/^{\circ}C$
- **Guaranteed load regulation:**
0.6% V_{OUT} @ 10mA to 1.5 amps
(LAS 1800 series)
- **Guaranteed ripple attenuation:**
59dB (LAS 1800 series)
- **Guaranteed thermal resistance, junction to case:** 3°C/W

DESCRIPTION

The LAS-1800 series voltage regulators are monolithic integrated circuits designed for use in applications requiring a well regulated negative output voltage. Outstanding features include full power usage up to 1.5 amperes of load current, internal current limiting, thermal shutdown, and safe area protection on the chip, providing protection of the series pass darlington, under most operating conditions. Hermetically sealed steel TO-3 packages are utilized for high reliability and low thermal resistance when used with an appropriate heat sink. A low-noise temperature stable diode reference is the key design factor insuring excellent temperature regulation of the LAS-1800 series. This coupled to a very low output impedance insures superior performance and load regulation.

The LAS-1800 series of three-terminal regulators are available in fixed output voltage tolerances of $\pm 5\%$ with nominal output voltages ranging from -2 to -28 volts. The LAS-18A00 three-terminal regulators are available in fixed output voltage tolerances of $\pm 2\%$ with nominal output voltages available of -5, -12, and -15 volts. The LAS-18U, a four-terminal adjustable regulator, is available with an output range from -2.6 to -30 volts, adjustable with a single potentiometer.

ABSOLUTE MAXIMUM RATINGS

PARAMETER	SYMBOL	MINIMUM	MAXIMUM	UNITS
Input Voltage	V_{IN}	-35 ⁽¹⁾ -40	0	Volts
Input/Output Differential	$V_{IN}-V_{OUT}$	0	30 ⁽¹⁾	Volts
Power Dissipation @ $T_C \leq 105^{\circ}C$	P_D		15 ^(1), 2)	Watts
Thermal Resistance Junction to Case	θ_{JC}		3	$^{\circ}C/Watt$
Operating Junction Temperature Range	T_J	-55	150	$^{\circ}C$
Storage Temperature Range	T_{STG}	-65	150	$^{\circ}C$
Lead Temperature (Soldering, 60 Seconds Time Limit)	T_{LEAD}		300	$^{\circ}C$

⁽¹⁾ The maximum input voltage of the LAS-1800 Series is limited by the maximum input-output differential, maximum power dissipation, and the maximum current limit-safe operating area, whichever is less. Values of -35V apply to LAS-1802—LAS-1812. Values of -40V apply to LAS-1815—LAS-1828, and LAS-18U.

⁽²⁾ For operation above 105°C T_{CASE} , derate @ 333 mW/ $^{\circ}C$.

REGULATOR PERFORMANCE SPECIFICATIONS — LAS 1800, 18U

INPUT VOLTAGE TEST CONDITIONS ARE AS FOLLOWS: $V_1 = V_0 + (-5)$ VOLTS, $V_2 = V_0 + (-15)$ VOLTS, $V_3 = V_0 + (-20)$ VOLTS,⁽⁵⁾ OR THE MINIMUM INPUT WHICHEVER IS LESS.

PARAMETER	SYMBOL	TEST CONDITIONS			LAS-1800 TEST LIMITS		LAS-18U TEST LIMITS		UNITS
		V_{IN}	I_O	T_J	MIN.	MAX.	MIN.	MAX.	
Output Voltage	V_O	V_1 to V_2	10mA to 1.0A	25°C	0.95 V_O ⁽¹⁾	1.05 V_O	-30 ⁽⁶⁾	-2.6	volts
Input-Output Differential	$V_{IN}-V_O$		1.0A	0-125°C	-2.1		-2.1		volts
Output Current	I_O	V_1		25°C	10mA	1.5	10mA	1.5	amps
Line Regulation ⁽²⁾	$REG_{(LINE)}$	V_1 to V_2	1.0A	25°C		2.0		2.0	% V_O
Load Regulation ⁽²⁾	$REG_{(LOAD)}$	V_1	10mA to 1.5A	25°C		0.6		0.6	% V_O
Quiescent Current	I_Q	V_1	10mA	25°C		10		10	mA
Quiescent Current Line	$I_Q(LINE)$	V_1 to V_3 ⁽⁵⁾	10mA	25°C		1.3		1.3	mA
Quiescent Current Load	$I_Q(LOAD)$	V_1	10mA to 1.5A	25°C		0.75		0.75	mA
Current Limit	I_{LIM}	V_1		25°C		3.5		3.5	amps
Short Circuit Current	I_S	V_1		25°C		3.5		3.5	amps
Temperature Coefficient	T_C	V_1	0.1A	0-125°C		0.03		0.03	% $V_O/^{\circ}C$
Output Noise Voltage	V_N	V_1	0.1A	0-125°C		10 ⁽³⁾		10 ⁽³⁾	$\mu V_{rms}/V$
Ripple Attenuation	R_A	V_1	1.0A	0-125°C		59 ⁽⁴⁾		59 ⁽⁴⁾	dB
Control Voltage	V_C	V_1 to V_2	10mA	0-125°C			-2.6	-2.25	volts

⁽¹⁾ Nominal output voltages are specified under ordering information.

⁽²⁾ Instantaneous regulation, average chip temperature changes must be accounted for separately.

⁽³⁾ BW = 10Hz—100KHz.

⁽⁴⁾ Ripple attenuation is specified for a 1Vrms, 120 Hz input ripple. Ripple attenuation is a minimum of 59dB at a 5 volt output, and is 1dB less for each volt increase in the output voltage.

⁽⁵⁾ The value of V_3 is constrained to $V_0 + (-20)$ volts or applicable absolute maximum input voltage rating, whichever is smaller in magnitude.

⁽⁶⁾ $V_O = V_C (1 + R_1/R_2)$

R_1 = Resistance from output to control.
 R_2 = Resistance from control to common.

LAS 1800, 18U, 18A00 SERIES

1.5 AMP NEGATIVE REGULATORS

REGULATOR PERFORMANCE SPECIFICATIONS, LAS 18A00

INPUT VOLTAGE TEST CONDITIONS ARE AS FOLLOWS: $V_1 = V_0 + (-5)$ VOLTS, $V_2 = V_0 + (-15)$ VOLTS, $V_3 = V_0 + (-20)$ VOLTS,⁽⁵⁾ OR THE MINIMUM INPUT WHICHEVER IS LESS.

PARAMETER	SYMBOL	TEST CONDITIONS			LAS-18A00 TEST LIMITS		
		V_{IN}	I_O	T_J	MIN.	MAX.	UNITS
Output Voltage	V_O	V_1 to V_2	10mA to 1.0A	25°C	0.98 V_O ⁽¹⁾	1.02 V_O	volts
Input-Output Differential	$V_{IN}-V_O$		1.0A	0-125°C	2.1		volts
Output Current	I_O	V_1		25°C	10mA	1.5	amps
Line Regulation ⁽²⁾	REG(LINE)	V_1 to V_2	1.0A	25°C		2.0	% V_O
Load Regulation ⁽²⁾	REG(LOAD)	V_1	10mA to 1.5A	25°C		0.6	% V_O
Quiescent Current	I_Q	V_1	10mA	25°C		10	mA
Quiescent Current Line	I_Q (LINE)	V_1 to V_3 ⁽⁵⁾	10mA	25°C		1.3	mA
Quiescent Current Load	I_Q (LOAD)	V_1	10mA to 1.5A	25°C		0.75	mA
Current Limit	I_{LIM}	V_1		25°C		3.5	amps
Short Circuit Current	I_S	V_1		25°C		3.5	amps
Temperature Coefficient	T_C	V_1	0.1A	0-125°C		0.02	% V_O /°C
Output Noise Voltage	V_N	V_1	0.1A	0-125°C		10 ⁽³⁾	μ Vrms/V
Ripple Attenuation	R_A	V_1	1.0A	0-125°C		59 ⁽⁴⁾	dB

⁽¹⁾ Nominal output voltages are specified under ordering information.

⁽²⁾ Instantaneous regulation, average chip temperature changes must be accounted for separately.

⁽³⁾ BW = 10Hz--100KHz.

⁽⁴⁾ Ripple attenuation is specified for a 1Vrms, 120Hz input ripple. Ripple attenuation is a minimum of 59dB at a 5 volt output, and is 1dB less for each volt increase in the output voltage.

⁽⁵⁾ The value of V_3 is constrained to $V_0 + (-20)$ volts or applicable absolute maximum input voltage rating, whichever is smaller in magnitude.

ORDERING INFORMATION

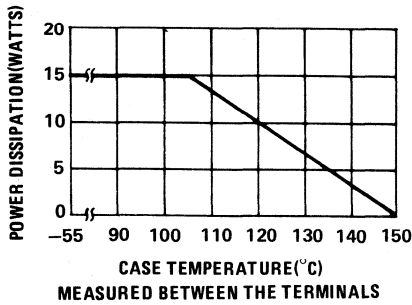
LAS 1800 SERIES 1.5 AMPS, 15 WATTS

MODEL	NOMINAL V_O VOLTS	PRICE QTY							
		1-24	25-49	50-99	100-249	250-499	500-999	1000-2499	2500-4999
LAS 1802	-2	\$2.95	\$2.70	\$2.47	\$2.30	\$2.10	\$1.95	\$1.80	\$1.60
LAS 1805	-5	2.95	2.70	2.47	2.30	2.10	1.95	1.80	1.60
LAS 18A05	-5	3.60	3.35	3.15	3.00	2.85	2.70	2.55	2.40
LAS 18052	-5.2	2.95	2.70	2.47	2.30	2.10	1.95	1.80	1.60
LAS 1806	-6	2.95	2.70	2.47	2.30	2.10	1.95	1.80	1.60
LAS 1808	-8	2.95	2.70	2.47	2.30	2.10	1.95	1.80	1.60
LAS 1810	-10	2.95	2.70	2.47	2.30	2.10	1.95	1.80	1.60
LAS 1812	-12	2.95	2.70	2.47	2.30	2.10	1.95	1.80	1.60
LAS 18A12	-12	3.60	3.35	3.15	3.00	2.85	2.70	2.55	2.40
LAS 1815	-15	2.95	2.70	2.47	2.30	2.10	1.95	1.80	1.60
LAS 18A15	-15	3.60	3.35	3.15	3.00	2.85	2.70	2.55	2.40
LAS 1818	-18	2.95	2.70	2.47	2.30	2.10	1.95	1.80	1.60
LAS 1820	-20	2.95	2.70	2.47	2.30	2.10	1.95	1.80	1.60
LAS 1824	-24	2.95	2.70	2.47	2.30	2.10	1.95	1.80	1.60
LAS 1828	-28	2.95	2.70	2.47	2.30	2.10	1.95	1.80	1.60
LAS 18U	-30 to -2.6	5.40	5.03	4.73	4.50	4.28	4.05	3.82	3.60

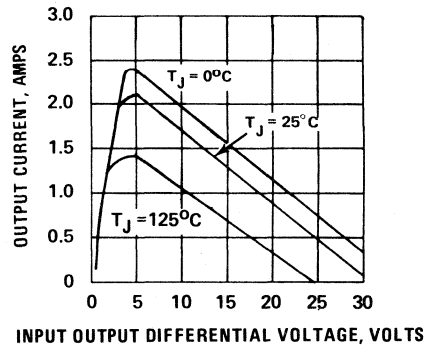
LAS 1800, 18U, 18A00 SERIES

1.5 AMP NEGATIVE REGULATORS

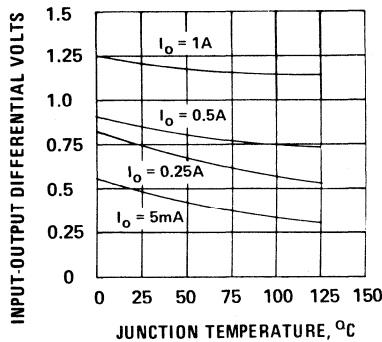
OPERATIONAL DATA



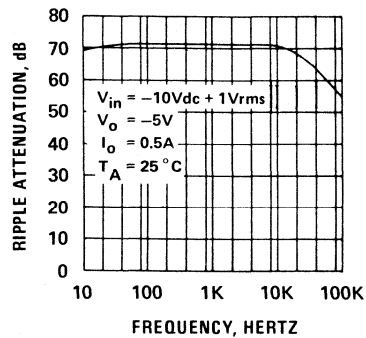
POWER DERATING



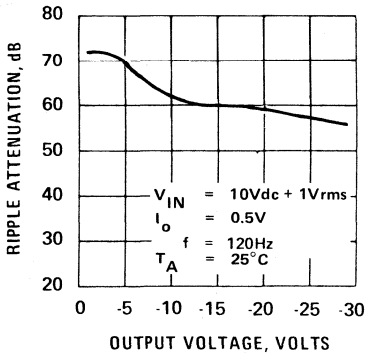
TYPICAL CURRENT LIMIT VS INPUT-OUTPUT VOLTAGE DIFFERENTIAL



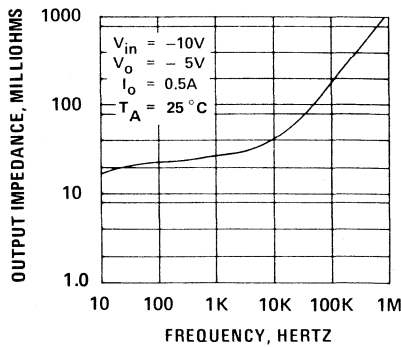
TYPICAL INPUT-OUTPUT DIFFERENTIAL VOLTAGE VS JUNCTION TEMPERATURE



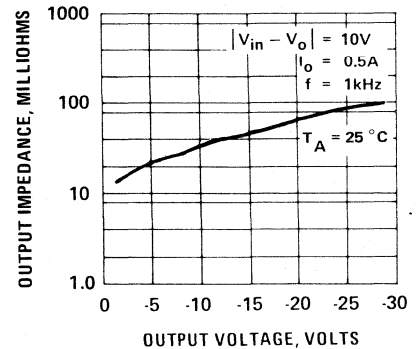
TYPICAL RIPPLE ATTENUATION VS FREQUENCY



TYPICAL RIPPLE ATTENUATION VS OUTPUT VOLTAGE

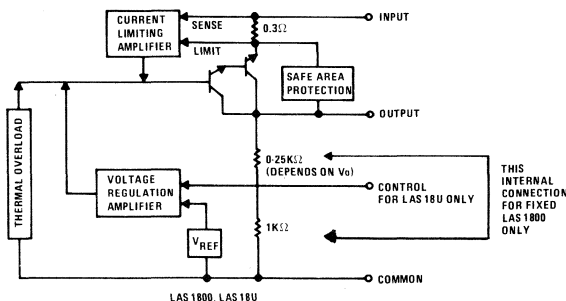


TYPICAL OUTPUT IMPEDANCE VS FREQUENCY



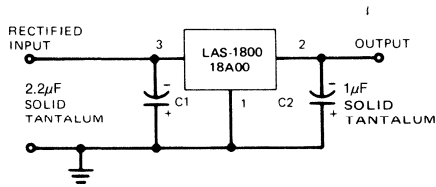
TYPICAL OUTPUT IMPEDANCE VS OUTPUT VOLTAGE

FUNCTIONAL BLOCK DIAGRAM



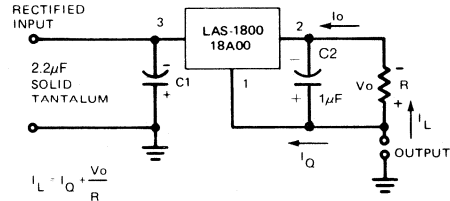
LAS 1800, 18U, 18A00 SERIES

TYPICAL APPLICATIONS



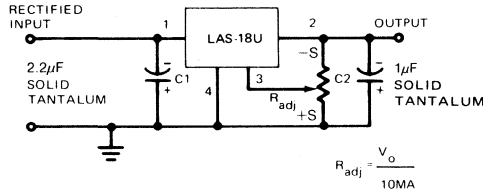
C1 AND C2 TO BE AS CLOSE TO THE DEVICE AS POSSIBLE.
 FILTER CAPACITOR = 2000µF/AMP

1.5 AMP NEGATIVE REGULATOR CIRCUIT



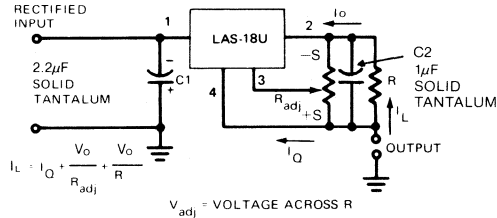
C1 TO BE AS CLOSE TO THE DEVICE AS POSSIBLE.
 FILTER CAPACITOR = 2000µF/AMP.

1.5 AMP FIXED CURRENT REGULATOR



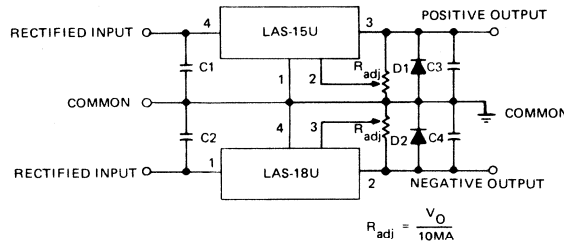
PIN 1 IS CONNECTED TO CASE.
 C1 AND C2 TO BE AS CLOSE TO THE DEVICE AS POSSIBLE.
 FILTER CAPACITOR = 2000µF/AMP.

1.5 AMP NEGATIVE ADJUSTABLE REGULATOR



PIN 1 IS CONNECTED TO CASE.
 C1 TO BE AS CLOSE TO THE DEVICE AS POSSIBLE.
 FILTER CAPACITOR = 2000µF/AMP.

1.5 AMP NEGATIVE ADJUSTABLE CURRENT REGULATOR



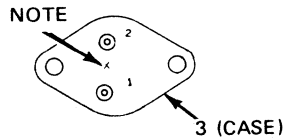
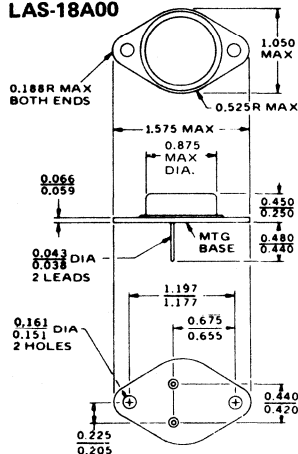
DIODES D1 & D2 SHOULD BE LOW V_F , HIGH CURRENT DIODES FOR OPTIMUM PROTECTION.
 C1, C2, C3 AND C4 TO BE PLACED AS CLOSE TO DEVICE AS POSSIBLE.
 FILTER CAPACITOR = 2000µF/AMP.

1.5 AMP DUAL ADJUSTABLE REGULATOR

OUTLINE DRAWING

TO-3 3-TERMINAL (STEEL)

LAS-1800
 LAS-18A00

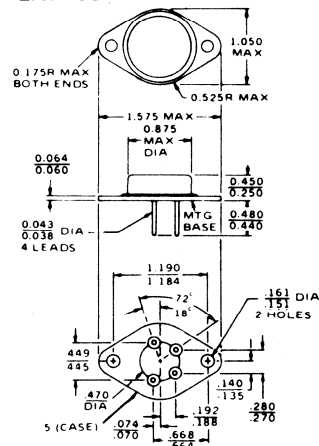


PIN	FUNCTION
1	COMMON
2	OUTPUT
3	INPUT

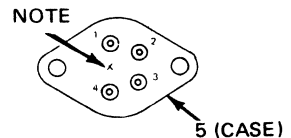
NOTE (X) = CASE TEMPERATURE MEASURED AT THIS POINT.

TO-3 4-TERMINAL (STEEL)

LAS-18U



FOR METRIC DIMENSIONS SEE PAGE 114-115



PIN	FUNCTION
1	INPUT (ELEC)
2	OUTPUT
3	CONTROL
4	COMMON
5	INPUT

NOTE (X) = CASE TEMPERATURE MEASURED AT THIS POINT

LAMBDA LLM 317

1.5 AMP 3-TERMINAL ADJUSTABLE REGULATOR

FEATURES

- Adjustable output down to 1.2V
- Guaranteed 1.5A output current
- Line regulation typically 0.01%/V
- Load regulation typically 0.1%
- Current limit constant with temperature
- Eliminates the need to stock many voltages
- Standard 3-lead transistor package
- 70 dB ripple rejection

DESCRIPTION

The LLM317 is an adjustable 3-terminal positive voltage regulator capable of supplying in excess of 1.5A over a 1.2V to 37V output range. This voltage regulator is easy to use and requires only two external resistors to get the output voltage. Outstanding features include internal current limiting, thermal shutdown and safe area protection on the chip. Hermetically sealed steel TO-3 packages are utilized for high reliability and low thermal resistance when used with an appropriate heat sink.

The LLM317 serves a wide variety of applications including local, on card regulation. This device also makes a simple adjustable switching regulator, a programmable output regulator, or by connecting a fixed resistor between the adjustment and output, the LLM317 can be used as a precision current regulator.

ABSOLUTE MAXIMUM RATINGS

Power Dissipation	Internally limited
Input—Output Voltage Differential	40V
Operating Junction Temperature Range	
LLM317	0°C to +125°C
Storage Temperature	−65°C to +150°C
Lead Temperature (Soldering, 10 seconds)	300°C

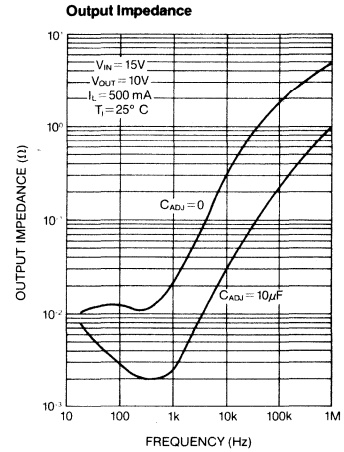
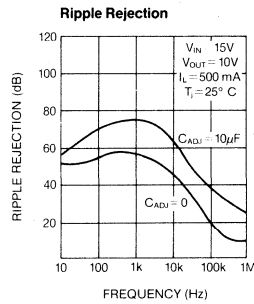
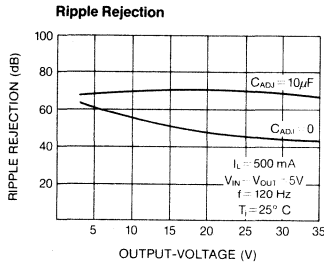
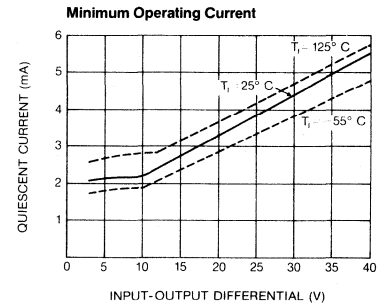
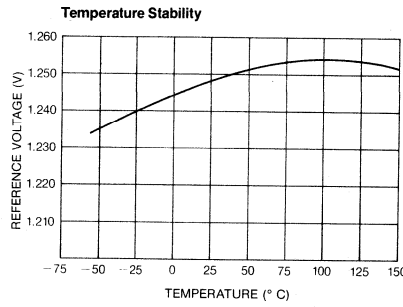
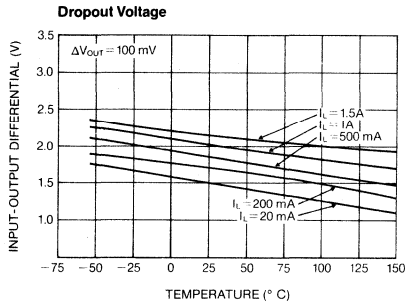
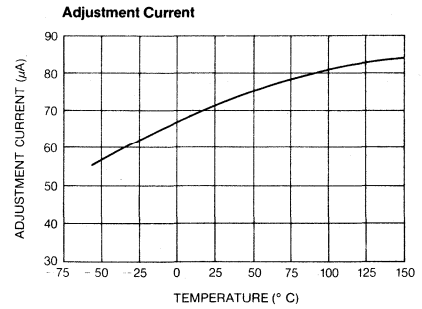
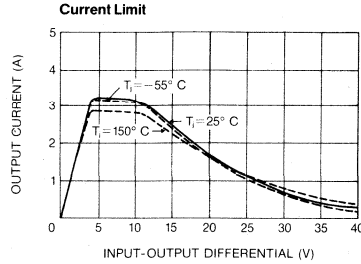
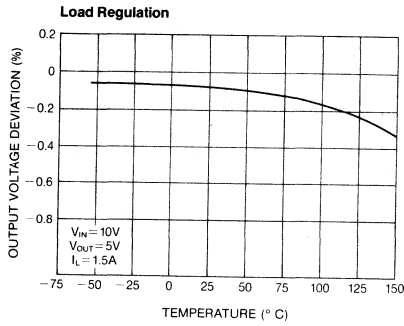
ELECTRICAL CHARACTERISTICS (NOTE 1)

PARAMETER	CONDITIONS	LLM317			UNITS
		MIN	TYP	MAX	
Line Regulation	$T_A = 25^\circ\text{C}$, $3\text{V} \leq V_{\text{IN}} - V_{\text{OUT}} \leq 40\text{V}$ (Note 2)		0.01	0.04	%/V
Load Regulation	$T_A = 25^\circ\text{C}$, $10\text{ mA} \leq I_{\text{OUT}} \leq I_{\text{MAX}}$ $V_{\text{OUT}} \leq 5\text{V}$, (Note 2) $V_{\text{OUT}} \geq 5\text{V}$, (Note 2)		5	25	mV
			0.1	0.5	%
Thermal Regulation	$T_A = 25^\circ\text{C}$, 20 ms Pulse		0.04	0.07	%/W
Adjustment Pin Current			50	100	μA
Adjustment Pin Current Change	$10\text{ mA} \leq I_L \leq I_{\text{MAX}}$ $2.5\text{V} \leq (V_{\text{IN}} - V_{\text{OUT}}) \leq 40\text{V}$		0.2	5	μA
Reference Voltage	$3 \leq (V_{\text{IN}} - V_{\text{OUT}}) \leq 40\text{V}$, $10\text{ mA} \leq I_{\text{OUT}} \leq I_{\text{MAX}}$, $P \leq P_{\text{MAX}}$	1.20	1.25	1.30	V
Line Regulation	$3\text{V} \leq V_{\text{IN}} - V_{\text{OUT}} \leq 40\text{V}$, (Note 2)		0.02	0.07	%/V
Load Regulation	$10\text{ mA} \leq I_{\text{OUT}} \leq I_{\text{MAX}}$, (Note 2) $V_{\text{OUT}} \leq 5\text{V}$ $V_{\text{OUT}} \geq 5\text{V}$		20	70	mV
			0.3	1.5	%
Temperature Stability	$T_{\text{MIN}} \leq T_J \leq T_{\text{MAX}}$		1		%
Minimum Load Current	$V_{\text{IN}} - V_{\text{OUT}} = 40\text{V}$		3.5	10	mA
Current Limit	$V_{\text{IN}} - V_{\text{OUT}} \leq 15\text{V}$	1.5	2.2		A
RMS Output Noise, % of V_{OUT}	$T_A = 25^\circ\text{C}$, $10\text{ Hz} \leq f \leq 10\text{ kHz}$		0.003		%
Ripple Rejection Ratio	$V_{\text{OUT}} = 10\text{V}$, $f = 120\text{ Hz}$ $C_{\text{ADJ}} = 10\mu\text{F}$		65		dB
		66	80		dB
Thermal Resistance, Junction to Case			2.3	3	$^\circ\text{C/W}$

Note 1: Unless otherwise specified, these specifications apply: $0^\circ\text{C} \leq T_J \leq +125^\circ\text{C}$ for the LLM317; $V_{\text{IN}} - V_{\text{OUT}} = 5\text{V}$ and $I_{\text{OUT}} = 0.5\text{A}$. Although power dissipation is internally limited, these specifications are applicable for power dissipations of 20W. I_{MAX} is 1.5A.

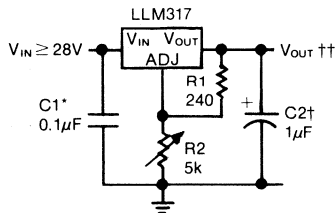
Note 2: Regulation is measured at constant junction temperature, using pulse testing with a low duty cycle. Changes in output voltage due to heating effects are covered under the specification for thermal regulation.

TYPICAL PERFORMANCE CHARACTERISTICS



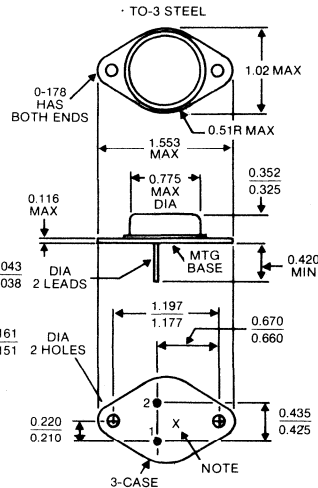
TYPICAL APPLICATIONS

1.2V—25V Adjustable Regulator



†Optional—improves transient response
*Needed if device is far from filter capacitors
 $\ddagger V_{OUT} = 1.25V \left(1 + \frac{R2}{R1} \right)$

PACKAGE OUTLINE



PIN	FUNCTION
1	ADJUSTMENT
2	V_{IN}
CASE	OUTPUT

NOTE (X): CASE TEMPERATURE MEASURED AT THIS POINT

ORDERING INFORMATION

MODEL	V_O VOLTS	I_O	PRICE QTY							
			1-24	25-49	50-99	100-249	250-499	500-999	1000-2499	2500-4999
LLM317	1.2-37.0	1.5A	\$3.75	\$3.10	\$3.10	\$2.65	\$2.65	\$2.65	\$2.00	\$1.90

LAS 1600, 16U SERIES

2.0 AMP POSITIVE REGULATORS

FEATURES

- **Guaranteed input-output differential:**
+2.60 volts
- **Guaranteed line regulation:**
2.0% V_{OUT} @ $\Delta V_{IN} = 7$ volts
- **Guaranteed load regulation:**
0.6% V_{OUT} @ 10mA to 2.0 amps
- **Guaranteed temperature coefficient:**
0 to 125°C — 0.02% $V_{OUT}/^{\circ}C$
- **Guaranteed ripple attenuation: 60dB**
- **Guaranteed thermal resistance, junction to case:**
LAS-1600/2.5°C/W
LAS-16U/2.5°C/W

DESCRIPTION

The LAS-1600 series voltage regulators are monolithic integrated circuits designed for use in applications requiring a well regulated positive output voltage. Outstanding features include full power usage up to 2.0 amperes of load current, internal current limiting, thermal shutdown, and safe area protection on the chip, providing protection of the series pass darlington, under most operating conditions. Hermetically sealed steel TO-3 packages are utilized for high reliability and low thermal resistance when used with an appropriate heat sink. A low-noise temperature stable diode reference is the key design factor insuring excellent temperature regulation of the LAS-1600 series. This coupled to a very low output impedance insures superior performance and load regulation.

The LAS-1600 series of three-terminal regulators are available in fixed output voltage tolerances of $\pm 5\%$ with nominal output voltages ranging from +5 to +15 volts. The LAS-16U, a four-terminal adjustable regulator, is available with an output range from +4 to +30 volts, adjustable with a single potentiometer.

ABSOLUTE MAXIMUM RATINGS

PARAMETER	SYMBOL	MINIMUM	MAXIMUM	UNITS
Input Voltage	V_{IN}	0	30 (35) ⁽¹⁾	Volts
Input/Output Differential	$V_{IN}-V_{OUT}$	0	25 ⁽¹⁾	Volts
Power Dissipation @ $T_c \leq 100^{\circ}C$	P_D		20 ^{(1), (2)}	Watts
Thermal Resistance Junction to Case	θ_{JC}		2.5	$^{\circ}C/Watt$
Operating Junction Temperature Range	T_J	-55	150	$^{\circ}C$
Storage Temperature Range	T_{STG}	-65	150	$^{\circ}C$
Lead Temperature (Soldering, 60 Seconds Time Limit)	T_{LEAD}		300	$^{\circ}C$

⁽¹⁾ The maximum input voltage of the LAS-1600 Series is limited by the maximum input-output differential, maximum power dissipation, or the maximum current limit-safe operating area, whichever is less. Value of +30V applies to V_O of +5 to +12V. Value of +35 volts applies to V_O of +13.8V and +15V.

⁽²⁾ For operation above 100°C T_{CASE} , derate @ 400 mW/ $^{\circ}C$.

REGULATOR PERFORMANCE SPECIFICATIONS

INPUT VOLTAGE TEST CONDITIONS ARE AS FOLLOWS: $V_1 = V_O + 3$ VOLTS, $V_2 = V_O + 10$ VOLTS, $V_3 = V_O + 12$ VOLTS, OR THE MAXIMUM INPUT WHICHEVER IS LESS.

PARAMETER	SYMBOL	TEST CONDITIONS			LAS-1600 TEST LIMITS		LAS-16U TEST LIMITS		UNITS
		V_{IN}	I_O	T_J	MIN.	MAX.	MIN.	MAX.	
Output Voltage	V_O	V_1 to V_2	10mA to 2.0A	25°C	0.95 V_O ⁽¹⁾	1.05 V_d	4.0 ⁽⁵⁾	30	volts
Input-Output Differential	$V_{IN}-V_O$		2.0A	0-125°C	2.6		2.6		volts
Output Current	I_O	V_1		25°C	10mA	2.0	10mA	2.0	amps
Line Regulation ⁽²⁾	$REG_{(LINE)}$	V_1 to V_2	1.0A	25°C		2.0		2.0	% V_O
Load Regulation ⁽²⁾	$REG_{(LOAD)}$	$V_0 + 5V$	10mA to 2.0A	25°C		0.6		0.6	% V_O
Quiescent Current	I_Q	V_1	10mA	25°C		15.0		15.0	mA
Quiescent Current Line	$I_Q (LINE)$	V_1 to V_2	10mA	25°C		5.0		5.0	mA
Quiescent Current Load	$I_Q (LOAD)$	V_1	10mA to 2.0A	25°C		5.0		5.0	mA
Current Limit	I_{LIM}	$V_0 + 5V$		25°C		4.5		4.5	amps
Short Circuit Current	I_S	$V_0 + 5V$		25°C		4.5		4.5	amps
Temperature Coefficient	T_C	V_1	0.1A	0-125°C		0.02		0.02	% $V_O/^{\circ}C$
Output Noise Voltage	V_N	V_1	0.1A	0-125°C		10 ⁽³⁾		10 ⁽³⁾	$\mu V_{rms}/V$
Ripple Attenuation	R_A	V_1	1.0A	0-125°C		60 ⁽⁴⁾		60 ⁽⁴⁾	dB
Control Voltage	V_C	V_1 to V_2	10mA	25°C			3.6	4.0	volts

⁽¹⁾ Nominal output voltages are specified under ordering information.

⁽²⁾ Instantaneous regulation, average chip temperature changes must be accounted for separately.

⁽³⁾ BW = 10Hz - 100KHz.

⁽⁴⁾ Ripple attenuation is specified for a 1Vrms, 120 Hz input ripple. Ripple attenuation is a maximum of 60dB (60dB for LAS-16U) at a 5 volt output, and is 1dB less for each volt increase in the output voltage. $V_N = V_O + 5$.

⁽⁵⁾ $V_O = V_C (1 + R1/R2)$

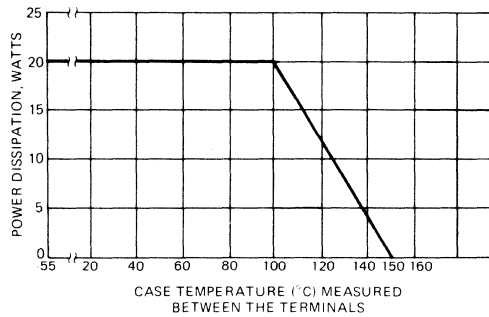
R1 = Resistance from output to control.

R2 = Resistance from control to common.

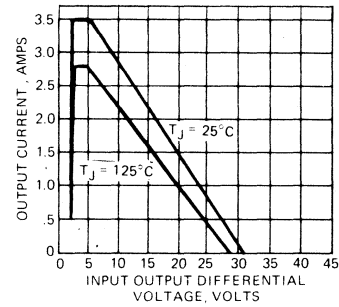
LAS 1600, 16U SERIES

2.0 AMP POSITIVE REGULATORS

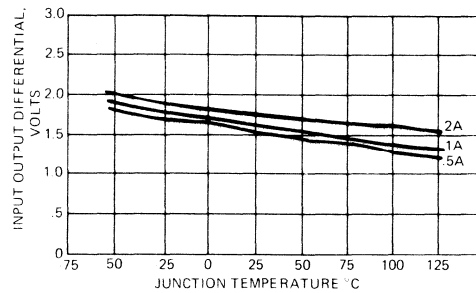
OPERATIONAL DATA



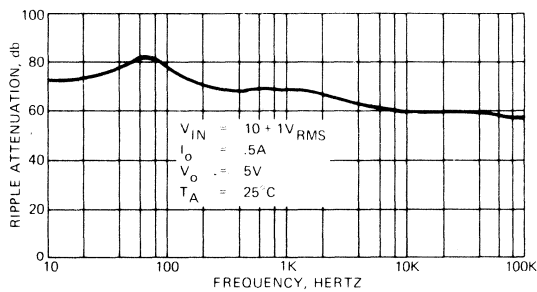
POWER DERATING



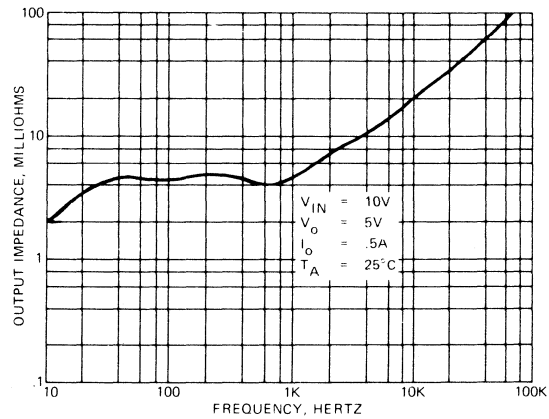
TYPICAL CURRENT LIMIT VS INPUT-OUTPUT VOLTAGE DIFFERENTIAL



TYPICAL INPUT-OUTPUT DIFFERENTIAL VOLTAGE VS JUNCTION TEMPERATURE



RIPPLE ATTENUATION VS FREQUENCY



TYPICAL OUTPUT IMPEDANCE VS FREQUENCY

ORDERING INFORMATION

LAS 1600 SERIES 2.0 AMPS, 20 WATTS

MODEL	NOMINAL V_o VOLTS	PRICE QTY							
		1-24	25-49	50-99	100-249	250-499	500-999	1000-2499	2500-4999
LAS 1605	5	\$3.00	\$2.80	\$2.60	\$2.40	\$2.24	\$2.15	\$2.05	\$1.90
LAS 1608	8	3.00	2.80	2.60	2.40	2.24	2.15	2.05	1.90
LAS 1612	12	3.00	2.80	2.60	2.40	2.24	2.15	2.05	1.90
LAS 16CB	13.8	3.00	2.80	2.60	2.40	2.24	2.15	2.05	1.90
LAS 1615	15	3.00	2.80	2.60	2.40	2.24	2.15	2.05	1.90
LAS 16U	4 to 30	4.50	4.20	3.90	3.60	3.40	3.25	3.08	2.85

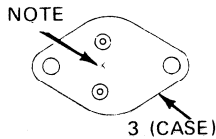
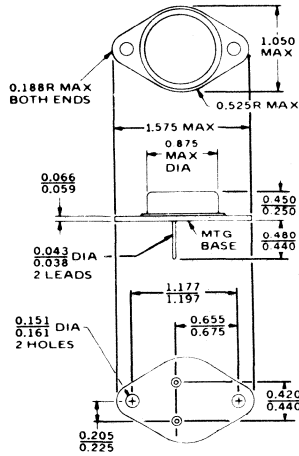
LAS 1600, 16U SERIES

2.0 AMP POSITIVE REGULATORS

OUTLINE DRAWING

TO-3 3-TERMINAL (STEEL)

LAS-1600 SERIES



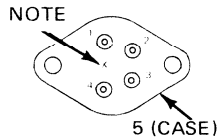
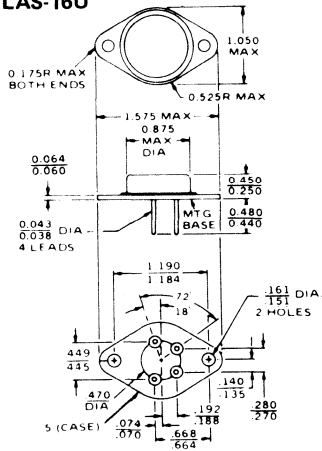
PIN	FUNCTION
1	INPUT
2	OUTPUT
3	COMMON

NOTE (X) = CASE TEMPERATURE MEASURED AT THIS POINT.

TO-3 4-TERMINAL (STEEL)

LAS-16U

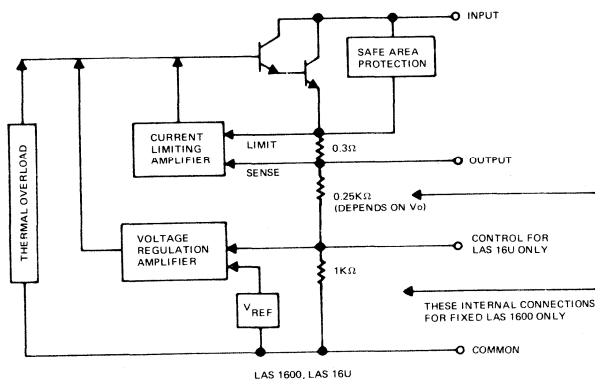
FOR METRIC DIMENSIONS SEE PAGE 114-115



PIN	FUNCTION
1	COMMON (ELEC)
2	CONTROL
3	OUTPUT
4	INPUT
5	COMMON

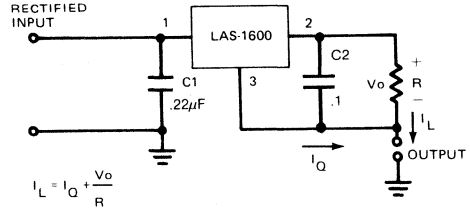
NOTE (X) = CASE TEMPERATURE MEASURED AT THIS POINT.

FUNCTIONAL BLOCK DIAGRAM



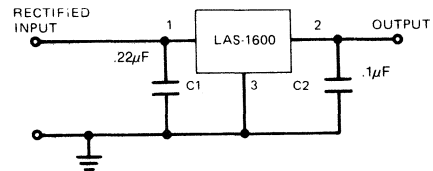
LAS 1600, LAS 16U

TYPICAL APPLICATIONS



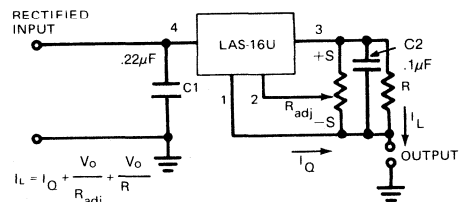
C1 TO BE PLACED AS CLOSE TO THE DEVICE AS POSSIBLE.
FILTER CAPACITOR = 2000μF/AMP.

2.0 AMP FIXED CURRENT REGULATOR



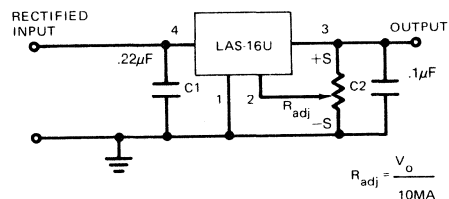
C1 AND C2 TO BE PLACED AS CLOSE TO THE DEVICE AS POSSIBLE.
FILTER CAPACITOR = 2000μF/AMP

2.0 AMP POSITIVE REGULATOR CIRCUIT



PIN 1 IS CONNECTED TO CASE.
C1 TO BE PLACED AS CLOSE TO THE DEVICE AS POSSIBLE.
FILTER CAPACITOR = 2000μF/AMP.

2.0 AMP POSITIVE ADJUSTABLE CURRENT REGULATOR



PIN 1 IS CONNECTED TO CASE.
C1 AND C2 TO BE PLACED AS CLOSE TO THE DEVICE AS POSSIBLE.
FILTER CAPACITOR = 2000μF/AMP.

2.0 AMP POSITIVE ADJUSTABLE REGULATOR

LAMBDA LAS 1400, 14AU SERIES

3 AMP POSITIVE VOLTAGE REGULATOR

FEATURES

- **Guaranteed input-output differential:**
LAS01400/+2.50 volts
LAS014AU/+2.30 volts
- **Guaranteed line regulation:**
LAS-1400/1.0% V_{OUT} @ $\Delta V_{IN} = 10$ volts
LAS-14AU/1.0% V_{OUT} @ $\Delta V_{IN} = 9$ volts
- **Guaranteed load regulation:**
0.6% V_{OUT} @ 10mA to 3.0 Amps output current
- **Guaranteed temperature coefficient:**
LAS-1400/0° C to +125° C: 0.02% $V_O/^\circ$ C
LAS-14AU/0° C to +125° C: 0.02% $V_O/^\circ$ C
- **Guaranteed ripple attenuation:**
LAS-1400/60dB min
LAS-14AU/66dB min
- **Guaranteed thermal resistance junction to case:**
LAS-1400/2.25° C/W
LAS-14AU/1.5° C/W

DESCRIPTION

The LAS-1400 series and the LAS-14AU voltage regulators are monolithic integrated circuits designed for use in applications requiring a well regulated positive output voltage. Outstanding features include full power usage up to 3.0 amperes of load current, internal current limiting, thermal shutdown, and safe area protection on the chip, providing protection of the series pass darlington, under most operating conditions. Hermetically sealed steel TO-3 packages are utilized for high reliability and low thermal resistance when used with an appropriate heat sink. A low-noise temperature stable diode reference is the key design factor insuring excellent temperature regulation. This coupled to a very low output impedance insures superior performance and load regulation.

The LAS-1400 series of three-terminal regulators are available in fixed output voltage tolerances of $\pm 5\%$ with nominal output voltages ranging from +5 to +15 volts. The LAS-14AU, a four-terminal adjustable positive voltage regulator, is available with an output voltage that can be adjusted over a 4.0 to 35 volt range by the use of a single potentiometer.

ABSOLUTE MAXIMUM RATINGS

PARAMETER	SYMBOL	MINIMUM	MAXIMUM	UNITS
Input Voltage	V_{IN}	0	35 (40) ⁽¹⁾	Volts
Input/Output Differential	$V_{IN}-V_{OUT}$	0	25 ⁽¹⁾	Volts
Power Dissipation				
@ $T_C \leq 105^\circ$ C	14AU			
@ $T_C \leq 80^\circ$ C	1400			
Thermal Resistance				
Junction to Case	θ_{JC}		1.5	$^\circ$ C/Watt
			2.25	
Operating Junction Temperature Range	T_J	-55	150	$^\circ$ C
Storage Temperature Range	T_{STG}	-65	150	$^\circ$ C
Lead Temperature (Soldering, 60 Seconds Time Limit)	T_{LEAD}		300	$^\circ$ C

- (1) The maximum input voltage of the LAS-1400 Series is limited by the maximum input-output differential, maximum power dissipation, or the maximum current limit-safe operating area, whichever is less. $V_{IN}(MAX)$ for LAS-1405, 1406, 1408, 1410, and 1412 is +35V $V_{IN}(MAX)$ for LAS-14CB, 1415 and 14AU is 40V.
- (2) For operation above 80° C T_{CASE} , derate @ 444 mW/°C for LAS-1405-1415. Above 105° C T_{CASE} , derate @ 667 mW/°C for LAS-14AU.

ELECTRICAL CHARACTERISTICS

INPUT VOLTAGE TEST CONDITIONS ARE AS FOLLOWS: $V_1 = V_O + 5$ VOLTS, $V_2 = V_O + 10$ VOLTS $V_3 = V_O + 15$ VOLTS, OR THE MAXIMUM INPUT WHICHEVER IS LESS. [FOR LAS-14AU, $V_1 = V_O + 3$ VOLTS, $V_2 = V_O + 10$ VOLTS $V_3 = V_O + 12$ VOLTS]

PARAMETER	SYMBOL	TEST CONDITIONS			LAS-1400 TEST LIMITS		LAS-14AU TEST LIMITS		UNITS
		V_{IN}	I_O	T_J	MIN.	MAX.	MIN.	MAX.	
Output Voltage	V_O	V_1 to V_2	10mA to 3.0A	25° C	0.95 V_O ⁽¹⁾	1.05 V_O	4.0	35 ⁽⁵⁾	volts
Input-Output Differential	$V_{IN}-V_O$		3.0A	0-125° C	2.5		2.3		volts
Output Current	I_O	V_1		25° C	0	3.0	0	3.0	amps
Line Regulation ⁽²⁾	$REG_{(LINE)}$	V_1 to V_3	2.0A	25° C		1.0		1.0	% V_O
Load Regulation ⁽²⁾	$REG_{(LOAD)}$	V_1	10mA to 3.0A	25° C		0.6		0.6	% V_O
Quiescent Current	I_Q	V_1	10mA	25° C		18		21	mA
Quiescent Current Line	$I_{Q(LINE)}$	V_1 to V_2	5mA	25° C		1.3		5.0 ⁽⁶⁾	mA
Quiescent Current Load	$I_{Q(LOAD)}$	V_1	5mA to 3.0A	25° C		1.5		5.0 ⁽⁷⁾	mA
Current Limit	I_{LIM}	$V_O + 5V$		25° C		5.2		5.5	amps
Short Circuit Current	I_S	$V_O + 5V$		25° C		5.2		5.5	amps
Temperature Coefficient	T_C	V_1	5mA	0-125° C		0.02		0.02	% $V_O/^\circ$ C
Output Noise Voltage	V_N	V_1	0.1A	0-125° C		10 ⁽³⁾		10 ⁽³⁾	μ Vrms/V
Ripple Attenuation	R_A	V_1	2.0A	0-125° C		60 ⁽⁴⁾		60 ⁽⁴⁾	dB
Control Voltage	V_C	V_1 to V_2	10mA	25° C				3.6	volts

⁽¹⁾ Nominal output voltages are specified under ordering information.

⁽²⁾ Instantaneous regulation, average chip temperature changes must be accounted for separately.

⁽³⁾ BW = 10Hz - 100KHz.

⁽⁴⁾ Ripple attenuation is specified for a 1Vrms, 120 Hz input ripple.

Ripple attenuation is a minimum of 60dB (60dB for LAS-14AU) at a 5 volt output, and is 1dB less for each volt increase in the output voltage. $V_{IN} = V_O + 5$

⁽⁵⁾ $V_O = V_C (1 + R_1/R_2)$

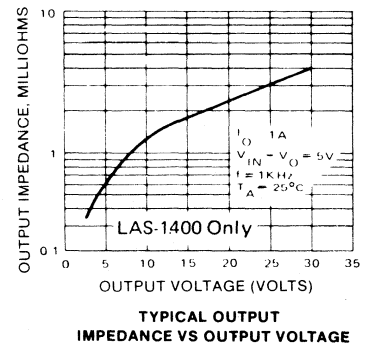
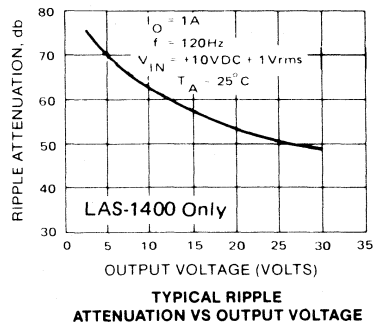
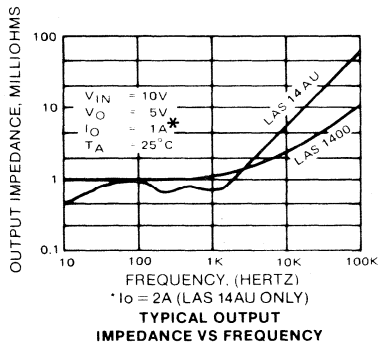
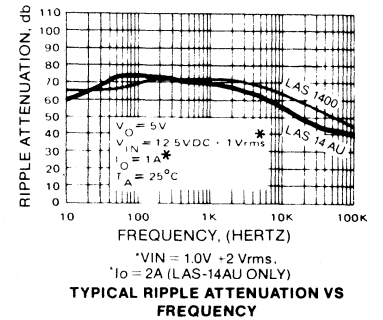
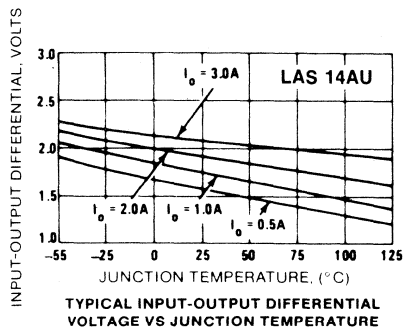
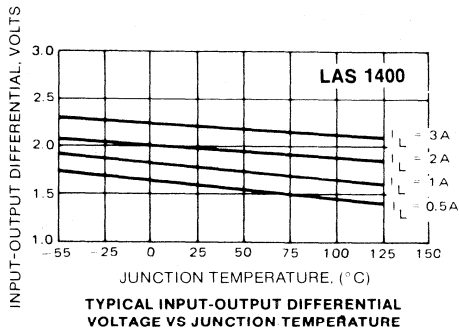
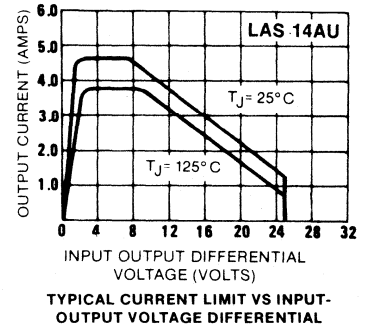
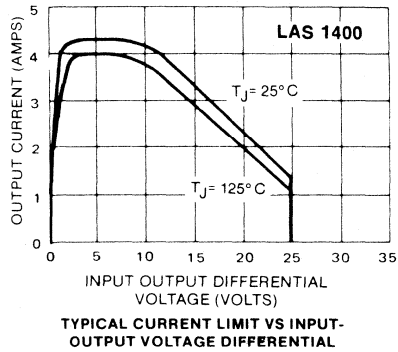
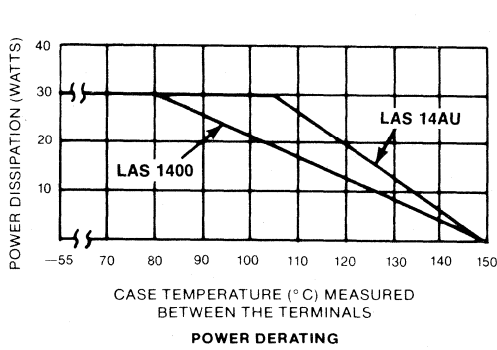
R_1 = Resistance from output to control.

R_2 = Resistance from control to common.

⁽⁶⁾ $I_O = 10$ mA for LAS-14AU

⁽⁷⁾ $I_O = 10$ mA to 3A for LAS-14AU

OPERATIONAL DATA



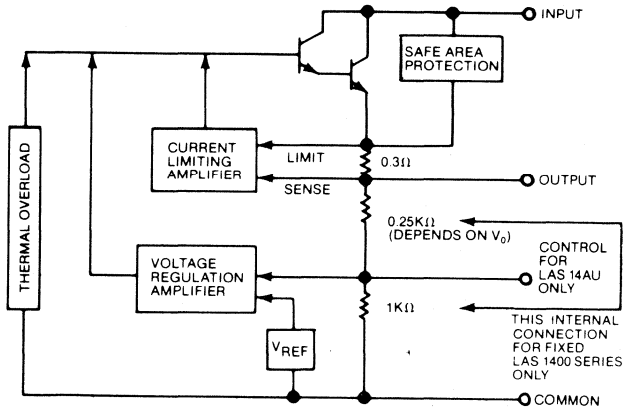
ORDERING INFORMATION

LAS 1400 SERIES 3.0 AMPS, 30 WATTS

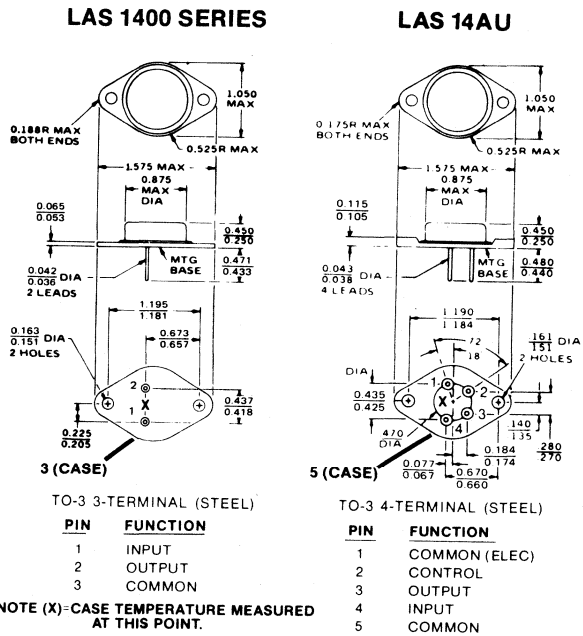
MODEL	NOMINAL V _o VOLTS	PRICE QTY							
		1-24	25-49	50-99	100-249	250-499	500-999	1000-2499	2500-4999
LAS 1405	5	\$5.00	\$4.75	\$4.50	\$4.30	\$4.10	\$3.90	\$3.70	\$3.40
LAS 1406	6	5.00	4.75	4.50	4.30	4.10	3.90	3.70	3.40
LAS 1412	12	5.00	4.75	4.50	4.30	4.10	3.90	3.70	3.40
LAS 1415	15	5.00	4.75	4.50	4.30	4.10	3.90	3.70	3.40
LAS 14AU	4 to 35	9.00	8.00	7.50	7.25	6.22	5.55	4.95	4.60

FUNCTIONAL BLOCK DIAGRAM

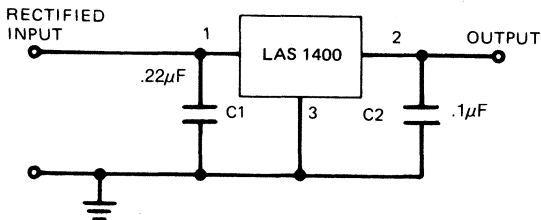
PACKAGE OUTLINE



LAS-1400 Series, LAS 14AU

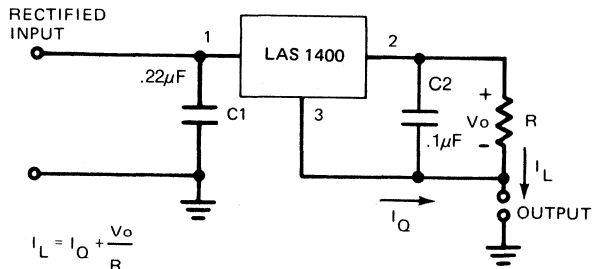


TYPICAL APPLICATIONS



C1 AND C2 TO BE PLACED AS CLOSE AS POSSIBLE TO DEVICE.
 FILTER CAPACITOR = 2000μF/AMP.

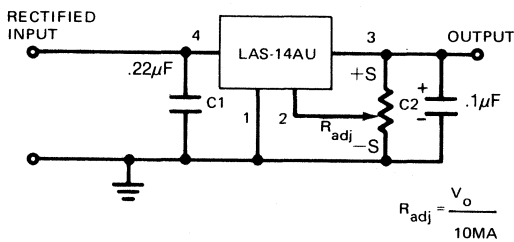
3.0 AMP POSITIVE REGULATOR CIRCUIT



$$I_L = I_Q + \frac{V_o}{R}$$

PIN 3 IS CONNECTED TO CASE.
 C1 TO BE PLACED AS CLOSE AS POSSIBLE TO DEVICE.
 FILTER CAPACITOR = 2000μF/AMP.

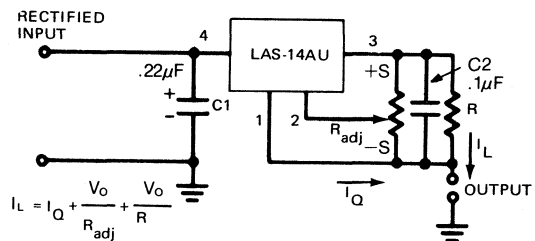
3.0 AMP FIXED CURRENT REGULATOR



$$R_{adj} = \frac{V_o}{10MA}$$

PIN 1 IS CONNECTED TO CASE.
 C1 AND C2 TO BE PLACED AS CLOSE TO THE DEVICE AS POSSIBLE.
 FILTER CAPACITOR = 2000μF/AMP.

3.0 AMP POSITIVE ADJUSTABLE CURRENT REGULATOR



$$I_L = I_Q + \frac{V_o}{R_{adj}} + \frac{V_o}{R}$$

PIN 1 IS CONNECTED TO CASE.
 C1 TO BE PLACED AS CLOSE TO THE DEVICE AS POSSIBLE.
 FILTER CAPACITOR = 2000μF/AMP.

3.0 AMP POSITIVE ADJUSTABLE REGULATOR

LAMBDA LLM 350

3 AMP 3-TERMINAL ADJUSTABLE REGULATOR

FEATURES

- Adjustable output down to 1.2V
- Guaranteed 3A output current
- Line regulation typically 0.005%/V
- Load regulation typically 0.1%
- Guaranteed thermal regulation
- Current limit constant with temperature
- Eliminates the need to stock many voltages
- Standard 3-lead transistor package
- 80 dB ripple rejection

DESCRIPTION

The LLM350 is an adjustable 3-terminal positive voltage regulator capable of supplying in excess of 3A over a 1.2V to 33V output range. This voltage regulator is easy to use and requires only two external resistors to get the output voltage. Outstanding features include internal current limiting, thermal shutdown and safe area protection on the chip. Hermetically sealed steel TO-3 packages are utilized for high reliability and low thermal resistance when used with an appropriate heat sink.

The LLM350 serves a wide variety of applications including local, on card regulation. This device also makes a simple adjustable switching regulator, a programmable output regulator, or by connecting a fixed resistor between the adjustment and output, the LLM350 can be used as a precision current regulator.

ABSOLUTE MAXIMUM RATINGS

Power Dissipation	Internally limited
Input—Output Voltage Differential	35V
Operating Junction Temperature Range	
LLM350	0° C to + 125° C
Storage Temperature	-65° C to + 150° C
Lead Temperature (Soldering, 10 second)	300° C

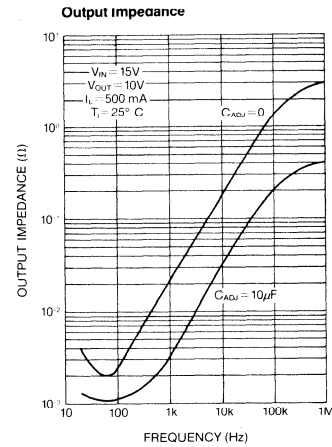
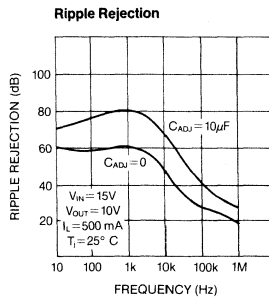
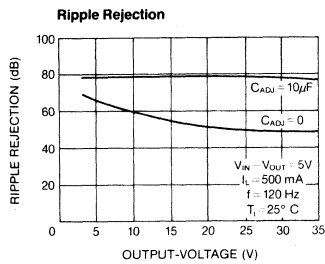
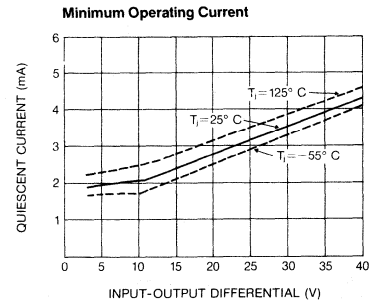
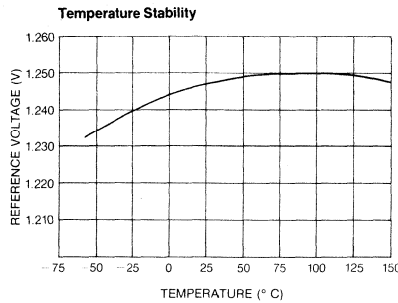
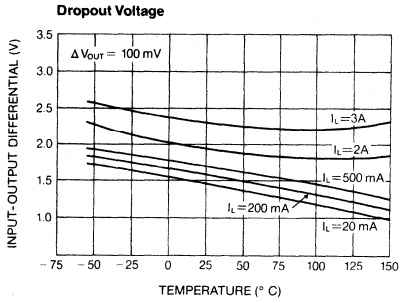
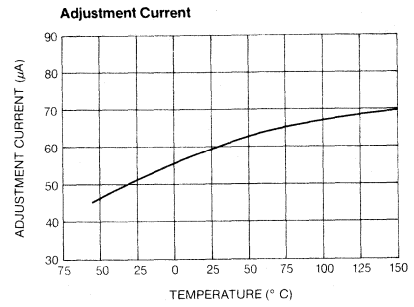
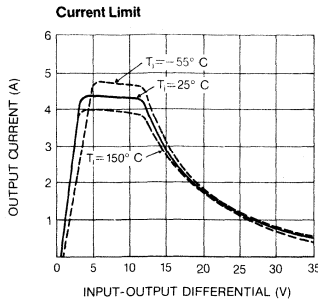
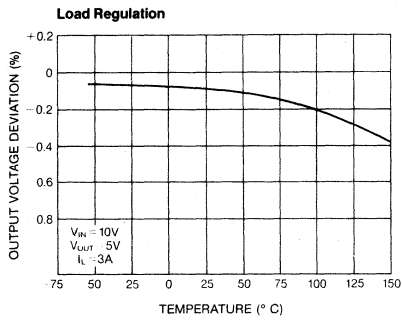
ELECTRICAL CHARACTERISTICS (NOTE 1)

PARAMETER	CONDITIONS	LLM350			UNITS
		MIN	TYP	MAX	
Line Regulation	$T_A = 25^\circ\text{C}$, $3\text{V} \leq V_{IN} - V_{OUT} \leq 35\text{V}$, (Note 2)		0.005	0.03	%/V
Load Regulation	$T_A = 25^\circ\text{C}$, $10\text{ mA} \leq I_{OUT} \leq 3\text{A}$ $V_{OUT} \leq 5\text{V}$, (Note 2) $V_{OUT} \geq 5\text{V}$, (Note 2)		5	25	mV
			0.1	0.5	%
Thermal Regulation	Pulse = 20 ms		0.002	0.03	%/W
Adjustment Pin Current			50	100	μA
Adjustment Pin Current Change	$10\text{ mA} \leq I_L \leq 3\text{A}$ $3\text{V} \leq (V_{IN} - V_{OUT}) \leq 35\text{V}$		0.2	5	μA
Reference Voltage	$3 \leq (V_{IN} - V_{OUT}) \leq 35\text{V}$, $10\text{ mA} \leq I_{OUT} \leq 3\text{A}$, $P \leq 30\text{W}$	1.20	1.25	1.30	V
Line Regulation	$3\text{V} \leq V_{IN} - V_{OUT} \leq 35\text{V}$, (Note 2)		0.02	0.07	%/V
Load Regulation	$10\text{ mA} \leq I_{OUT} \leq 3\text{A}$, (Note 2) $V_{OUT} \leq 5\text{V}$ $V_{OUT} \geq 5\text{V}$		20	70	mV
			0.3	1.5	%
Temperature Stability	$T_{MIN} \leq T_J \leq T_{MAX}$		1		%
Minimum Load Current	$V_{IN} - V_{OUT} = 35\text{V}$		3.5	10	mA
Current Limit	$V_{IN} - V_{OUT} \leq 10\text{V}$ $V_{IN} - V_{OUT} = 30\text{V}$	3.0	4.5		A
			1		A
RMS Output Noise, % of V_{OUT}	$T_A = 25^\circ\text{C}$, $10\text{ Hz} \leq f \leq 10\text{ kHz}$		0.003		%
Ripple Rejection Ratio	$V_{OUT} = 10\text{V}$, $f = 120\text{ Hz}$ $C_{ADJ} = 10\mu\text{F}$	66	65		dB
			86		dB
Thermal Resistance, Junction to Case				1.5	$^\circ\text{C}/\text{W}$

Note 1: Unless otherwise specified, these specifications apply: $0^\circ\text{C} \leq T_J \leq +125^\circ\text{C}$ for the LLM350; $V_{IN} - V_{OUT} = 5\text{V}$ and $I_{OUT} = 1.5\text{A}$. Although power dissipation is internally limited, these specifications are applicable for power dissipations up to 30W.

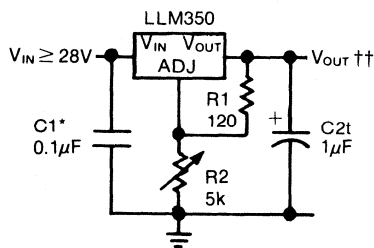
Note 2: Regulation is measured at constant junction temperature. Changes in output voltage due to heating effects must be taken into account separately. Pulse testing with low duty cycle is used.

TYPICAL PERFORMANCE CHARACTERISTICS



TYPICAL APPLICATIONS

1.2V—25V Adjustable Regulator

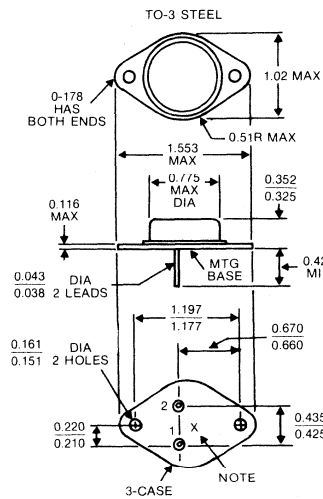


†Optional—improves transient response

*Needed if device is far from filter capacitors

$$††V_{OUT} = 1.25V \left(1 + \frac{R2}{R1} \right)$$

PACKAGE OUTLINE



PIN	FUNCTION
1	ADJUSTMENT
2	V_{IN}
CASE	OUTPUT

NOTE (X): CASE TEMPERATURE MEASURED AT THIS POINT

ORDERING INFORMATION

MODEL	V_O VOLTS	I_O	PRICE QTY							
			1-24	25-49	50-99	100-249	250-499	500-999	1000-2499	2500-4999
LLM 350	1.2-33.0	3.0A	\$5.75	\$4.80	\$4.80	\$4.00	\$4.00	\$4.00	\$3.35	\$3.20

LAMBDA LAS 1900, 19U SERIES

5 AMP POSITIVE VOLTAGE REGULATOR

FEATURES

- Guaranteed input—output differential: +2.6 Volts
- Guaranteed line regulation: 2.0% V_{OUT} @ $\Delta V_{IN} = 9$ Volts
- Guaranteed load regulation: 0.6% V_{OUT} @ 10mA to 5 Amps
- Guaranteed temperature coefficient:
 LAS-1900/0° C to +125° C: 0.03% $V_O/^\circ$ C
 LAS-19U/0° C to +125° C: 0.02% $V_O/^\circ$ C
- Guaranteed ripple attenuation:
 LAS-1900/60dB min
 LAS-19U/62dB min
- Guaranteed thermal resistance junction to case:
 LAS-1900/0.9° C/W
 LAS-19U/0.9° C/W

DESCRIPTION

The LAS-1900 series voltage regulators are monolithic integrated circuits designed for use in applications requiring a well regulated positive output voltage. Outstanding features include full power usage up to 5.0 amperes of load current, internal current limiting, thermal shutdown, and safe area protection on the chip, providing protection of the series pass darlington, under most operating conditions. Hermetically sealed copper TO-3 packages are utilized for high reliability and low thermal resistance when used with an appropriate heat sink. A low-noise temperature stable diode reference is the key design factor insuring excellent temperature regulation of the LAS-1900 series. This coupled to a very low output impedance insures superior performance and load regulation.

The LAS-1900 series of three-terminal regulators are available in fixed output voltage tolerances of $\pm 5\%$ with nominal output voltages ranging from +5 to +15 volts. The LAS-19U, a four terminal adjustable, regulator is available with an output range from +4 to +30 volts, adjustable with a single potentiometer.

ABSOLUTE MAXIMUM RATINGS

PARAMETER	SYMBOL	MINIMUM	MAXIMUM	UNITS
Input Voltage	V_{IN}	0	30 (35) ⁽¹⁾	Volts
Input/Output Differential	$V_{IN}-V_{OUT}$	0	25 ⁽¹⁾	Volts
Power Dissipation @ $T_C \leq 90^\circ$ C	P_D		50 ⁽¹⁾⁽²⁾	Watts
Thermal Resistance Junction To Case	θ_{JC}		0.9	°C/Watt
Operating Junction Temperature Range	T_J	-55	135	°C
Storage Temperature Range	T_{STG}	-65	150	°C
Lead Temperature (Soldering, 60 Seconds Time Limit)	T_{LEAD}		300	°C

(1) The maximum input voltage of the LAS-1900 Series is limited by the maximum input-output differential, maximum power dissipation, or the maximum current limit-safe operating area, whichever is less. Value of 30V applies to V_O of +5V to +12V. Value of 35V applies to V_O of 13.8V, 15V, and LAS-19U.
 (2) For operation above 90° C T_{CASE} derate @ 1.111 watt/°C

ELECTRICAL CHARACTERISTICS

INPUT VOLTAGE TEST CONDITIONS ARE AS FOLLOWS: $V_1 = V_O + 3$ VOLTS, $V_2 = V_O + 10$ VOLTS, $V_3 = V_O + 12$ VOLTS OR THE MAXIMUM INPUT WHICHEVER IS LESS.

PARAMETER	SYMBOL	TEST CONDITIONS			LAS-1900 TEST LIMITS		LAS-19U TEST LIMITS		UNITS
		V_{IN}	I_O	T_J	MIN.	MAX.	MIN.	MAX.	
Output Voltage	V_O	V_1 to V_2	10mA to 5.0A	25° C	$0.95 V_O ^{(1)}$	$1.05 V_O $	$4.0^{(5)}$.30	volts
Input-Output Differential	$V_{IN}-V_O$		5.0A	0-125° C	2.6		2.6		volts
Output Current	I_O	V_1		25° C	10mA	5.0	10mA	.50	amps
Line Regulation ⁽²⁾	$REG_{(LINE)}$	V_1 to V_3	3.0A	25° C		2.0		1.0	% V_O
Load Regulation ⁽²⁾	$REG_{(LOAD)}$	V_1	10mA to 5.0A	25° C		0.6		0.6	% V_O
Quiescent Current	I_Q	V_1	10mA	25° C		20.0		25.0	mA
Quiescent Current Line	$I_{Q(LINE)}$	V_1 to V_2	10mA	25° C		5.0		5.0	mA
Quiescent Current Load	$I_{Q(LOAD)}$	V_1	10mA to 5.0A	25° C		5.0		5.0	mA
Current Limit	I_{LIM}	$V_O + 5V$		25° C		12.0		12.0	amps
Short Circuit Current	I_S	$V_O + 5V$		25° C		12.0		12.0	amps
Temperature Coefficient	T_C	V_1	0.1A	0-125° C		0.03		0.02	% $V_O/^\circ$ C
Output Noise Voltage	V_N	V_1	0.1A	0-125° C		$10^{(3)}$		$10^{(3)}$	μ Vrms/V
Ripple Attenuation	R_A	V_1	2.0A	0-125° C	60 ⁽⁴⁾		60 ⁽⁴⁾		dB
Control Voltage	V_C	V_1 to V_2	10mA	25° C			3.6	4.0	volts

⁽¹⁾ Nominal output voltages are specified under ordering information.

⁽²⁾ Instantaneous regulation, average chip temperature changes must be accounted for separately.

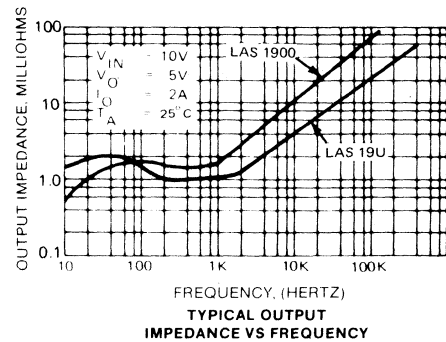
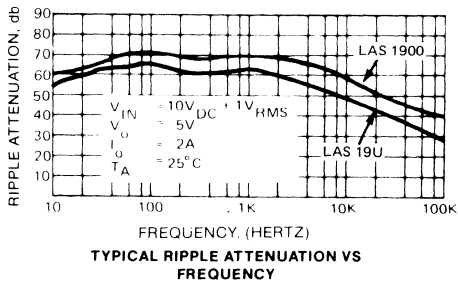
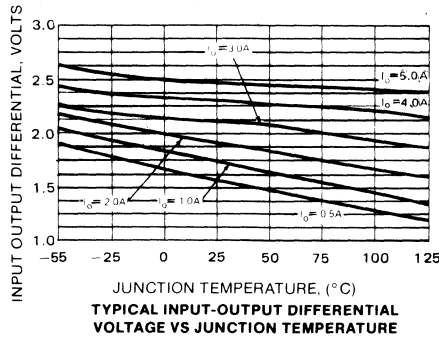
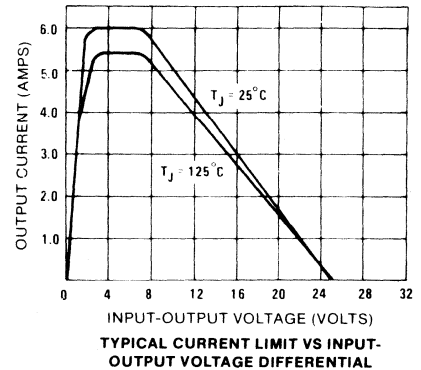
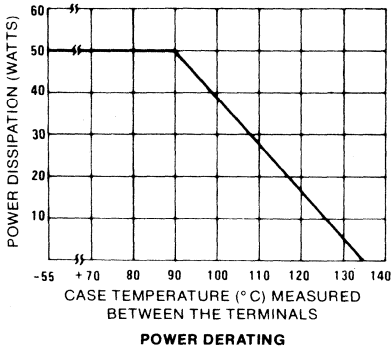
⁽³⁾ BW = 10Hz - 100KHz.

⁽⁴⁾ Ripple attenuation is specified for a 1Vrms, 120 Hz input ripple. Ripple attenuation is a minimum of 60dB (60dB for LAS-19U) at a 5 volt output, and is 1dB less for each volt increase in the output voltage. $V_{IN} = V_O + 5$

⁽⁵⁾ $V_O = V_C (1 + R_1/R_2)$

R1 = Resistance from output to control.
 R2 = Resistance from control to common.

OPERATIONAL DATA

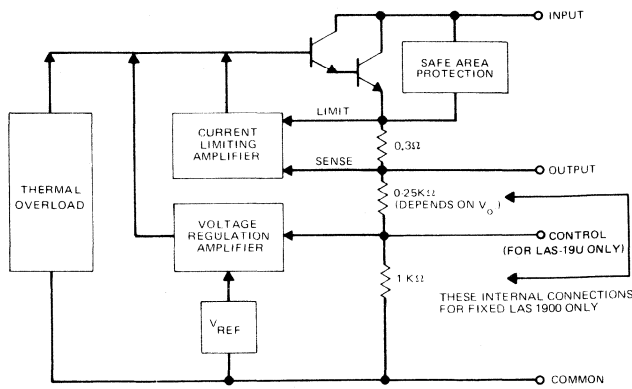


ORDERING INFORMATION

LAS 1900 SERIES 5.0 AMPS, 50 WATTS

MODEL	NOMINAL V_O VOLTS	PRICE QTY							
		1-24	25-49	50-99	100-249	250-499	500-999	1000-2499	2500-4999
LAS 1905	5	\$14.00	\$12.50	\$11.75	\$10.00	\$7.85	\$7.85	\$7.00	\$6.85
LAS 1912	12	14.00	12.50	11.75	10.00	7.85	7.85	7.00	6.85
LAS 1915	15	14.00	12.50	11.75	10.00	7.85	7.85	7.00	6.85
LAS 19U	4-30	16.00	14.50	13.75	10.25	9.75	9.00	8.50	8.00

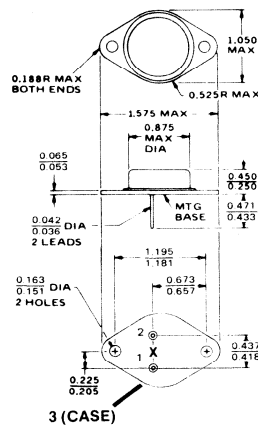
BLOCK DIAGRAM



LAS-1900 Series, LAS 19U

PACKAGE OUTLINE

LAS 1900 SERIES

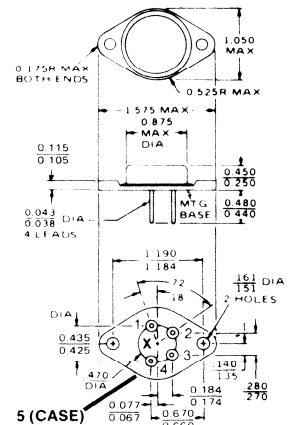


TO-3 3-TERMINAL (COPPER)

PIN	FUNCTION
1	INPUT
2	OUTPUT
3	COMMON

NOTE (X) CASE TEMPERATURE MEASURED AT THIS POINT.

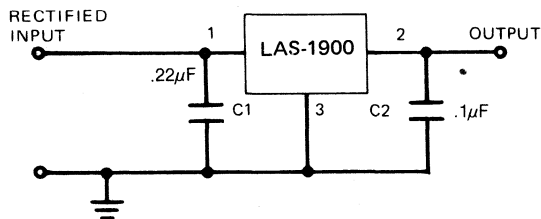
LAS 19U



TO-3 4-TERMINAL (COPPER)

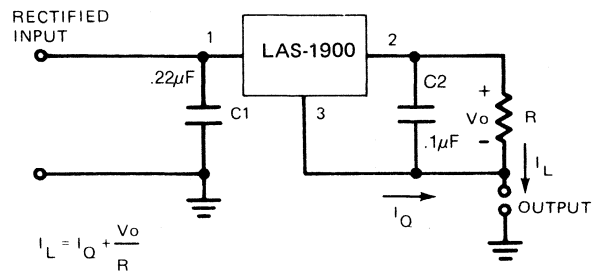
PIN	FUNCTION
1	COMMON (E.I.F.C.)
2	CONTROL
3	OUTPUT
4	INPUT
5	COMMON

TYPICAL APPLICATIONS



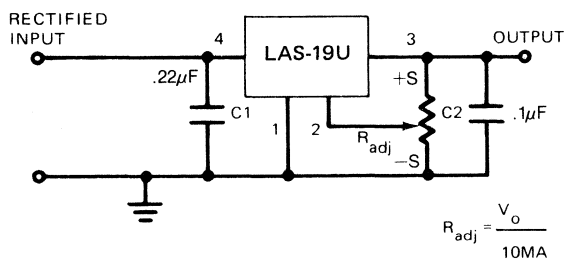
C1 AND C2 TO BE PLACED AS CLOSE TO THE DEVICE AS POSSIBLE.
 FILTER CAPACITOR = 2000μF/AMP.

5.0 AMP POSITIVE REGULATOR CIRCUIT



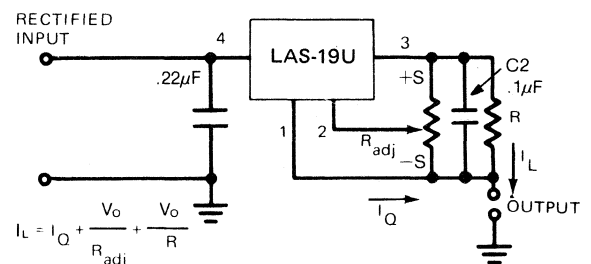
C1 AND C2 TO BE PLACED AS CLOSE TO THE DEVICE AS POSSIBLE.
 FILTER CAPACITOR = 2000μF/AMP.

5.0 AMP FIXED CURRENT REGULATOR



PIN 1 IS CONNECTED TO CASE.
 C1 AND C2 TO BE PLACED AS CLOSE TO THE DEVICE AS POSSIBLE.
 FILTER CAPACITOR = 2000μF/AMP.

5.0 AMP POSITIVE ADJUSTABLE REGULATOR



PIN 1 IS CONNECTED TO CASE.
 C1 AND C2 TO BE PLACED AS CLOSE TO THE DEVICE AS POSSIBLE.
 FILTER CAPACITOR = 2000μF/AMP.

5.0 AMP POSITIVE ADJUSTABLE CURRENT REGULATOR

LAMBDA LLM 338

5 AMP 3-TERMINAL ADJUSTABLE REGULATOR

FEATURES

- Guaranteed 7A peak output current
- Guaranteed 5A output current
- Adjustable output down to 1.2V
- Line regulation typically 0.005%/V
- Load regulation typically 0.1%
- Guaranteed thermal regulation
- Current limit constant with temperature
- Standard 3-lead transistor package

DESCRIPTION

The LLM338 is an adjustable 3-terminal positive voltage regulator capable of supplying in excess of 5A over a 1.2V to 32V output range. This voltage regulator is easy to use and requires only two external resistors to get the output voltage. Outstanding features include internal current limiting, thermal shutdown and safe area protection on the chip. Hermetically sealed steel TO-3 packages are utilized for high reliability and low thermal resistance when used with an appropriate heat sink.

The LLM338 serves a wide variety of applications including local, on card regulation. This device also makes a simple adjustable switching regulator, a programmable output regulator, or by connecting a fixed resistor between the adjustment and output, the LLM338 can be used as a precision current regulator.

ABSOLUTE MAXIMUM RATINGS

Power Dissipation	Internally limited
Input—Output Voltage Differential	35V
Operating Junction Temperature Range	
LLM338	0° C to + 125° C
Storage Temperature	-65° C to + 150° C
Lead Temperature (Soldering, 10 second)	300° C

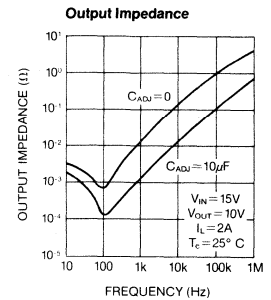
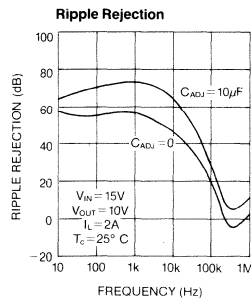
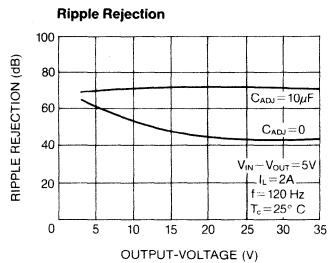
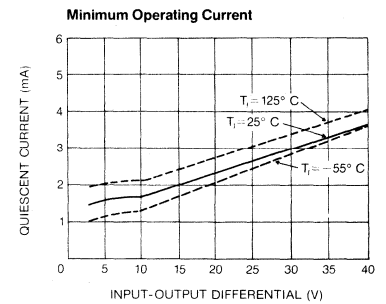
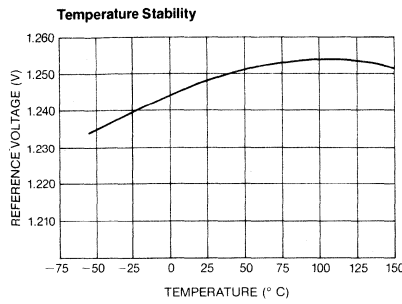
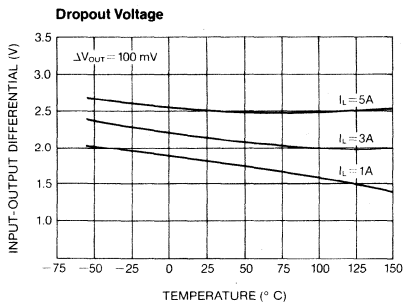
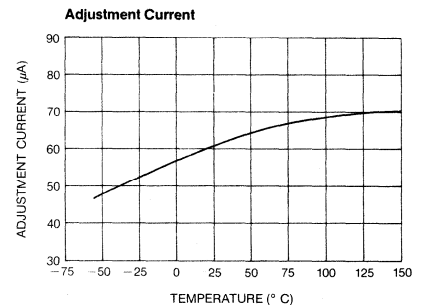
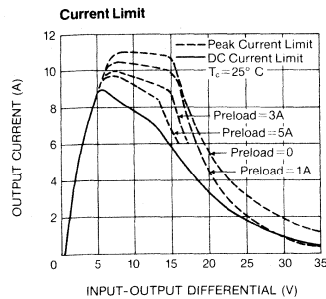
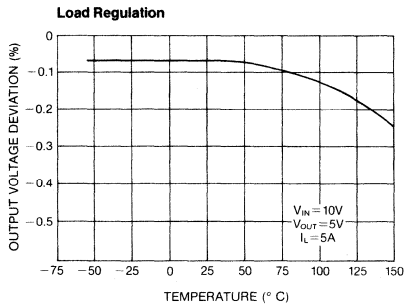
ELECTRICAL CHARACTERISTICS (NOTE 1)

PARAMETER	CONDITIONS	LLM338			UNITS
		MIN	TYP	MAX	
Line Regulation	$T_A = 25^\circ\text{C}$, $3\text{V} \leq V_{\text{IN}} - V_{\text{OUT}} \leq 35\text{V}$, (Note 2)		0.005	0.03	%/V
Load Regulation	$T_A = 25^\circ\text{C}$, $10\text{mA} \leq I_{\text{OUT}} \leq 5\text{A}$ $V_{\text{OUT}} \leq 5\text{V}$, (Note 2) $V_{\text{OUT}} \geq 5\text{V}$, (Note 2)		5	25	mV
			0.1	0.5	%
Thermal Regulation	Pulse = 20 ms		0.002	0.02	%/W
Adjustment Pin Current			45	100	μA
Adjustment Pin Current Change	$10\text{mA} \leq I_L \leq 5\text{A}$ $3\text{V} \leq (V_{\text{IN}} - V_{\text{OUT}}) \leq 35\text{V}$		0.2	5	μA
Reference Voltage	$3 \leq (V_{\text{IN}} - V_{\text{OUT}}) \leq 35\text{V}$, $10\text{mA} \leq I_{\text{OUT}} \leq 5\text{A}$, $P \leq 50\text{W}$	1.19	1.24	1.29	V
Line Regulation	$3\text{V} \leq V_{\text{IN}} - V_{\text{OUT}} \leq 35\text{V}$, (Note 2)		0.02	0.06	%/V
Load Regulation	$10\text{mA} \leq I_{\text{OUT}} \leq 5\text{A}$, (Note 2) $V_{\text{OUT}} \leq 5\text{V}$ $V_{\text{OUT}} \geq 5\text{V}$		20	50	mV
			0.3	1.0	%
Temperature Stability	$T_{\text{MIN}} \leq T_j \leq T_{\text{MAX}}$		1		%
Minimum Load Current	$V_{\text{IN}} - V_{\text{OUT}} = 35\text{V}$		3.5	10	mA
Current Limit	$V_{\text{IN}} - V_{\text{OUT}} \leq 10\text{V}$ DC	5.0	8		A
		7	12		A
			1		A
RMS Output Noise, % of V_{OUT}	$T_A = 25^\circ\text{C}$, $10\text{Hz} \leq f \leq 10\text{kHz}$		0.003		%
Ripple Rejection Ratio	$V_{\text{OUT}} = 10\text{V}$, $f = 120\text{Hz}$ $C_{\text{ADJ}} = 10\mu\text{F}$		60		dB
		60	75		dB
Thermal Resistance, Junction to Case				1.0	$^\circ\text{C}/\text{W}$

Note 1: Unless otherwise specified, these specifications apply: $0^\circ\text{C} \leq T_j \leq +125^\circ\text{C}$ for the LLM338; $V_{\text{IN}} - V_{\text{OUT}} = 5\text{V}$ and $I_{\text{OUT}} = 2.5\text{A}$. Although power dissipation is internally limited, these specifications are applicable for power dissipations up to 50W.

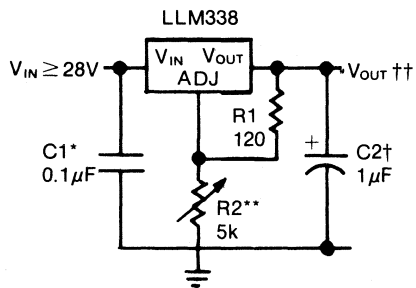
Note 2: Regulation is measured at constant junction temperature. Changes in output voltage due to heating effects must be taken into account separately by thermal regulation.

TYPICAL PERFORMANCE CHARACTERISTICS



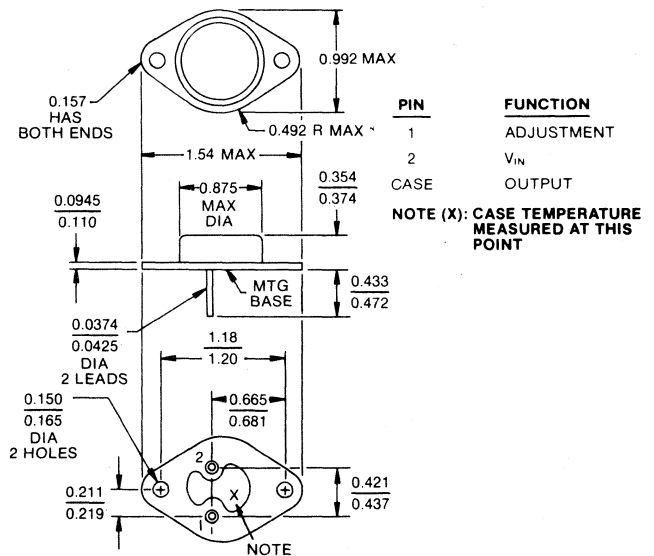
TYPICAL APPLICATIONS

1.2V—25V Adjustable Regulator



†Optional—improves transient response
 *Needed if device is far from filter capacitors
 $\dagger\dagger V_{OUT} = 1.25V \left(1 + \frac{R2}{R1} \right)$

PACKAGE OUTLINE



ORDERING INFORMATION

MODEL	V_O VOLTS	I_O	PRICE QTY							
			1-24	25-49	50-99	100-249	250-499	500-999	1000-2499	2500-4999
LLM 338	1.2-32.0	5.0A	\$7.60	\$6.39	\$6.39	\$5.00	\$5.00	\$5.00	\$4.15	\$4.00

LAMBDA LAS 3900

8 AMP POSITIVE VOLTAGE REGULATOR

FEATURES

- **Guaranteed input-output differential: +2.6 Volts**
- **Guaranteed line regulation: 2.0% V_{OUT} , $\Delta V_{IN} = 12$ Volts**
- **Guaranteed load regulation: 0.6% V_{OUT} @ 10mA to 8.0 Amps**
- **Guaranteed ripple attenuation: 60 dB @ 120 Hz**
- **Guaranteed temperature coefficient: 0 to 125° C — 0.03% $V_{OUT}/^{\circ}$ C**
- **Guaranteed thermal resistance junction to case: 0.7° C/W**
- **100% power cycled under load**
- **Sense terminal pin eliminates line drop problems**

DESCRIPTION

The LAS-3900 series voltage regulators are monolithic integrated circuits designed for use in applications requiring a well regulated positive output voltage. Outstanding features include full power usage up to 3.0 amperes of load current, internal current limiting, thermal shutdown, and safe area protection on the chip, providing protection of the series pass darlington, under most operating conditions. Hermetically sealed copper TO-3 packages are utilized for high reliability and low thermal resistance when used with an appropriate heat sink. A low-noise temperature stable diode reference is the key design factor insuring excellent temperature regulation of the LAS-3900 series. This coupled to a very low output impedance insures superior performance and load regulation.

Device	Pins	V_{OUT} Tolerance
LAS 3905	4	±5%
LAS 39U	4	Adjustable
LAS 3905K	2	+5%, -3%

ABSOLUTE MAXIMUM RATINGS

PARAMETER	SYMBOL	MINIMUM	MAXIMUM	UNITS
Input Voltage	V_{IN}	0	25 ⁽¹⁾	Volts
Input/Output Differential	$V_{IN}-V_{OUT}$	0	20 ⁽¹⁾	Volts
Power Dissipation @ $T_C \leq 94^{\circ}$ C	P_D		80 ^(1) 2 3)	Watts
Thermal Resistance Junction To Case	θ_{JC}		0.7	°C/Watt
Operating Junction Temperature Range	T_J	-55	150	°C
Storage Temperature Range	T_{STG}	-65	150	°C
Lead Temperature (Soldering, 60 Seconds Time Limit)	T_{LEAD}		300	°C

- (1) The maximum input voltage of the LAS-3900 Series is limited by the maximum input-output differential, maximum power dissipation, or the maximum current limit safe operating area, whichever is less.
- (2) For operation above 94° C T_{CASE} , derate @ 1.42 watt/° C.
- (3) In case of a short circuit, with input-output voltages approaching the maximum input-output differential, this regulator may require the removal of the input voltage to restart.

ELECTRICAL CHARACTERISTICS

INPUT VOLTAGE TEST CONDITIONS ARE AS FOLLOWS: $V_1 = V_0 + 3$ VOLTS, $V_2 = V_0 + 10$ VOLTS, $V_3 = V_0 + 15$ VOLTS, OR THE MAXIMUM INPUT WHICHEVER IS LESS.

PARAMETER	SYMBOL	TEST CONDITIONS			LAS 3905, 3905K		LAS 39U		UNITS
		V_{IN}	I_O	T_J	MIN.	MAX.	MIN.	MAX.	
Output Voltage	V_O	V_1 to V_2	10mA to 8.0A	25° C	0.95 V_O ⁽¹⁾⁽⁵⁾	1.05 V_O	4.0 ⁽⁵⁾	16	volts
Input-Output Differential	$V_{IN}-V_O$		8.0A	0-125° C	2.6	2.6	2.6		volts
Output Current	I_O	V_1		25° C	10	8.0	10mA	8.0	amps
Line Regulation ⁽²⁾	$REG_{(LINE)}$	V_1 to V_3	5A	25° C	2.0	2.0	2.0		% V_O
Load Regulation ⁽²⁾	$REG_{(LOAD)}$	V_1	10mA to 8.0A	25° C	0.6 ⁽⁶⁾	0.6 ⁽⁶⁾	0.6		% V_O
Quiescent Current	I_Q	V_1		25° C	20	20	25.0		mA
Quiescent Current Line	$I_{Q(LINE)}$	V_1 to V_2	10mA	25° C	5	5	5.0		mA
Quiescent Current Load	$I_{Q(LOAD)}$	V_1	10mA to 8.0A	25° C	5	5	5.0		mA
Current Limit	I_{LIM}	$V_0 + 5V$		25° C	14	14	14.0		amps
Short Circuit Current	I_S	$V_0 + 5V$		25° C	14	14	14.0		amps
Temperature Coefficient	T_C	V_1	0.1A	0-125° C	0.03	0.03	0.03		% $V_O/^{\circ}$ C
Output Noise Voltage	V_N	V_1	0.1A	0-125° C	10 ⁽³⁾	10 ⁽³⁾	10 ⁽³⁾		μ Vrms/V
Ripple Attenuation	R_A	V_1	2.0A	0-125° C	60 ⁽⁴⁾	60 ⁽⁴⁾	60 ⁽⁴⁾		dB
Control Voltage	V_C	V_1 to V_2	10mA	25° C			3.6	4.0	volts

⁽¹⁾ Nominal output voltages are specified under ordering information.

⁽²⁾ Instantaneous regulation, average chip temperature changes must be accounted for separately.

⁽³⁾ BW=10Hz—100KHz.

⁽⁴⁾ Ripple attenuation is specified for a 1Vrms, 120 Hz input ripple. Ripple attenuation is a minimum of 60 dB (60 dB for LAS-39U) at a 5 volt output, and is .1dB less for each volt increase in the output voltage. $V_{IN} = V_0 + 5$

R_1

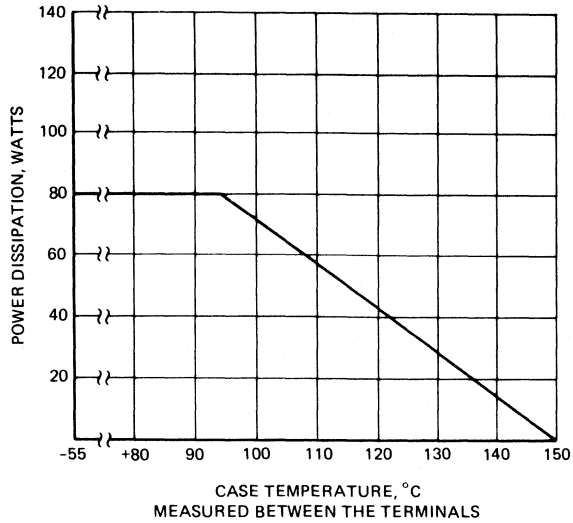
⁽⁵⁾ $V_O = V_C(1 + R_2)$

R_1 —Resistance from output to control.

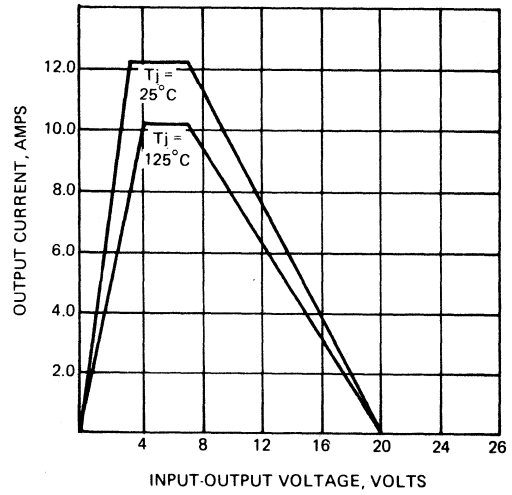
R_2 —Resistance from control to common.

⁽⁶⁾ LAS 3905K $V_{O, MIN} = 0.97 V_O$, $REG_{(LOAD), MAX} = 2\%$

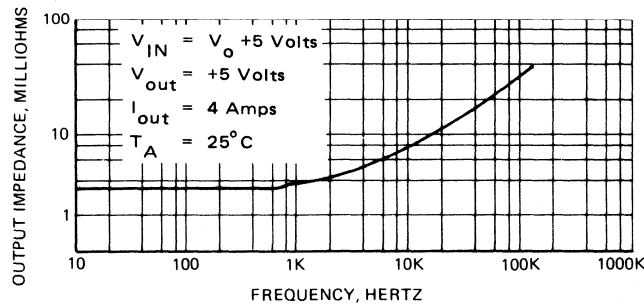
OPERATIONAL DATA



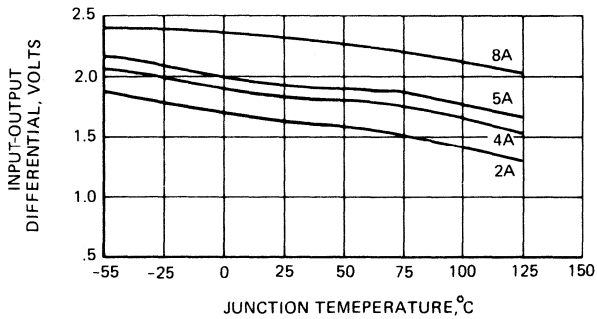
POWER DERATING



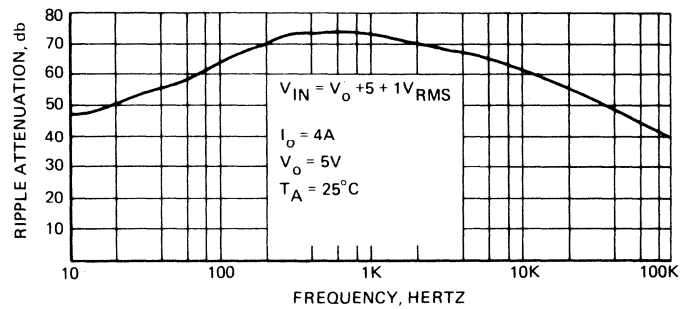
TYPICAL CURRENT LIMIT VS INPUT-OUTPUT VOLTAGE DIFFERENTIAL



TYPICAL OUTPUT IMPEDANCE vs FREQUENCY



TYPICAL INPUT-OUTPUT DIFFERENTIAL VOLTAGE vs JUNCTION TEMPERATURE



TYPICAL RIPPLE ATTENUATION vs FREQUENCY

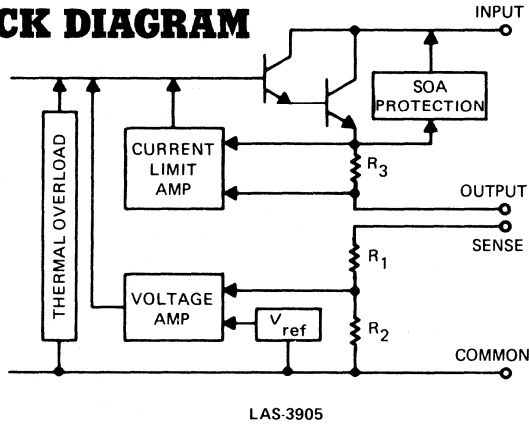
ORDERING INFORMATION

MODEL	NOMINAL V_o VOLTS	PRICE QTY							
		1-24	25-49	50-99	100-249	250-499	500-999	1000-2499	2500-4999
LAS 3905	5	\$18.00	\$16.50	\$15.75	\$13.00	\$13.00	\$11.90	\$10.65	\$10.00
LAS 3905K	5	16.00	14.25	13.80	11.65	10.75	9.85	9.50	9.25
LAS 39U	4-16	19.00	17.50	16.75	14.00	14.00	12.90	11.65	11.00

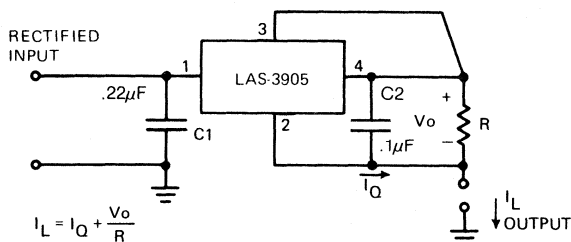
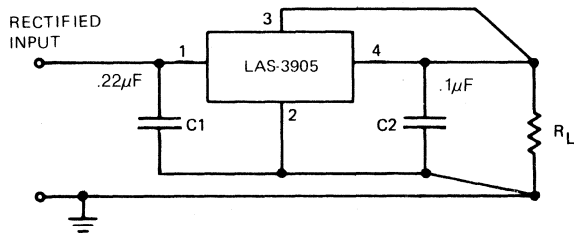
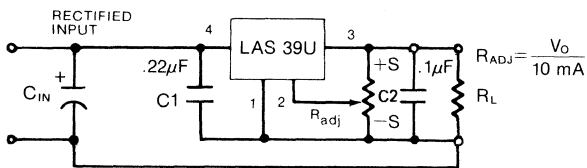
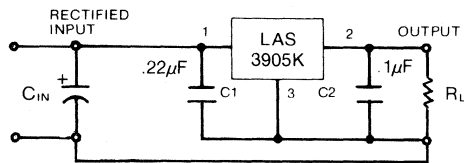
△ LAMBDA LAS 3900 SERIES

8 AMP, 80 WATT MONOLITHIC POSITIVE VOLTAGE REGULATORS

BLOCK DIAGRAM

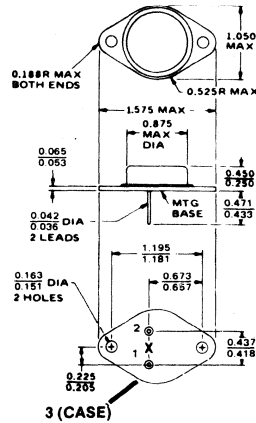


TYPICAL APPLICATIONS

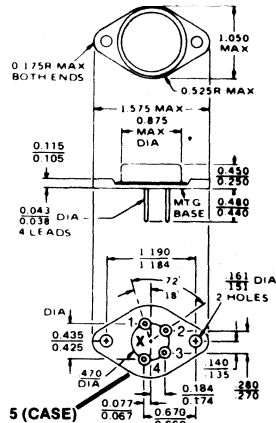


OUTLINE DRAWING

LAS 3905K SERIES



LAS 39U



TO-3 3-TERMINAL (COPPER)

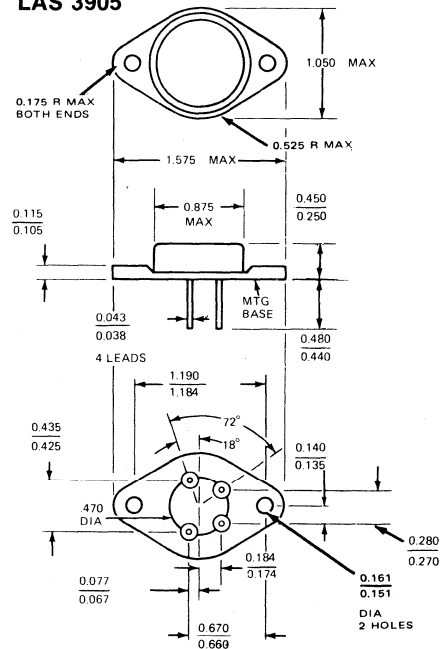
PIN	FUNCTION
1	INPUT
2	OUTPUT
3	COMMON

NOTE (X) - CASE TEMPERATURE MEASURED AT THIS POINT.

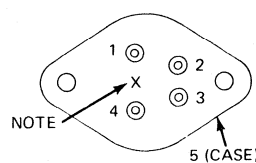
TO-3 4-TERMINAL (COPPER)

PIN	FUNCTION
1	COMMON (ELEC)
2	CONTROL
3	OUTPUT
4	INPUT
5	COMMON

LAS 3905



COPPER 4 TERMINAL/TO-3



NOTE (X) = Case temperature measured at this point.

PIN FUNCTION

PIN	FUNCTION
1	INPUT
2	COMMON (ELEC)
3	SENSE
4	OUTPUT
5	COMMON

LAMBDA LAS 3700

FLOATING REGULATOR

FEATURES

- Voltage/current regulation with automatic crossover
- Load regulation — .003% ± 1mV
- Line regulation — .01% ± 1mV $V_{IN} = 15-25V$
- Short circuit protection
- On chip stabilized heater
- Adjustable current source for programming voltage
- Fixed current source for current regulator
- Hermetic TO5 package

DESCRIPTION

The LAS 3700 floating regulator is designed for applications requiring "laboratory" power supply performance at voltages limited only by the external series pass transistor. Both output voltage and output current are adjustable. A unique feature of the LAS 3700 is the incorporation of an on chip heater which maintains a constant junction temperature, of 115C. This feature is responsible for the excellent temperature performance of the regulator.

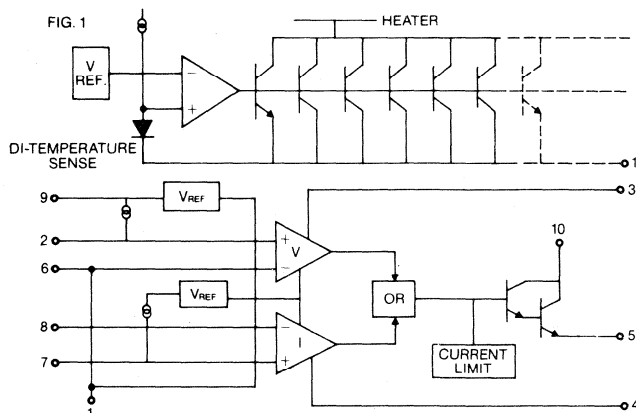
ABSOLUTE MAXIMUM RATINGS

PARAMETER	SYMBOL	MINIMUM	MAXIMUM	UNITS
Input Voltage	V_{AUX}	0	40	Volts
Thermal Resistance Junction to Ambient	θ_{JA}		220	°C/Watt
Operating Ambient Temperature Range	T_{AMB}	-30	115	°C
Storage Temperature	T_{STG}	-65	150	°C
Lead Soldering Temperature			300	°C

ELECTRICAL CHARACTERISTICS

PARAMETER	SYMBOL	MINIMUM	MAXIMUM	UNITS
Auxiliary Input Voltage ⁽¹⁾	V_{AUX}	14	40	Volts
Auxiliary Input Current ^{(2), (6)}	I_{AUX}	8	150	mA
Reference Current-Voltage Loop ⁽³⁾	I_{REF-V}		1	mA
Reference Current-Current Loop ⁽⁴⁾	I_{REF-C}		1.5	mA
Thermal Resistance	θ_{JA}		220	C/Watt
Output Current	I_{OUT}		1	mA
Output Short Circuit Current	I_{SC}		3	mA
Regulation — Line for a 5V Change in V_{AUX}	$REG_{(LINE)}$.05% ± 1	mV
Regulation — Load for a 0-1 mA Change in Output Current	$REG_{(LOAD)}$.003% $V_{AUX} \pm 1$	mV
Temperature Coefficient	T_C		.005% V_{OUT}	Volts °C
Common Mode Rejection		60		dB
Input Offset Voltage ⁽⁵⁾ (Voltage Amp)			40	mV
Input Offset Voltage ⁽⁵⁾ (Current Amp)			40	mV

FUNCTIONAL BLOCK DIAGRAM



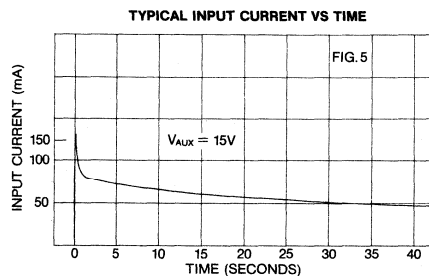
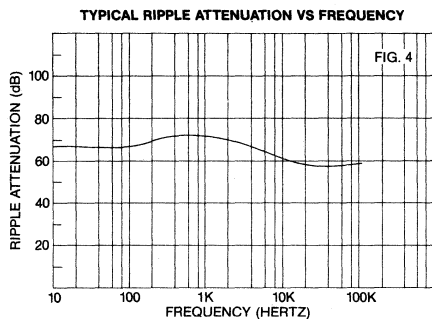
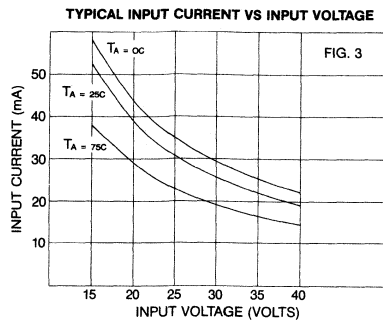
PIN CONNECTIONS

1. V_{AUX} COMMON
2. V_{ADJ}
3. VOLTAGE LOOP COMPENSATION
4. CURRENT LOOP COMPENSATION
5. OUTPUT
6. VOLTAGE SENSE
7. CURRENT OFFSET
8. CURRENT SENSE
9. VOLTAGE REFERENCE CURRENT ADJUST
10. V_{AUX} POSITIVE

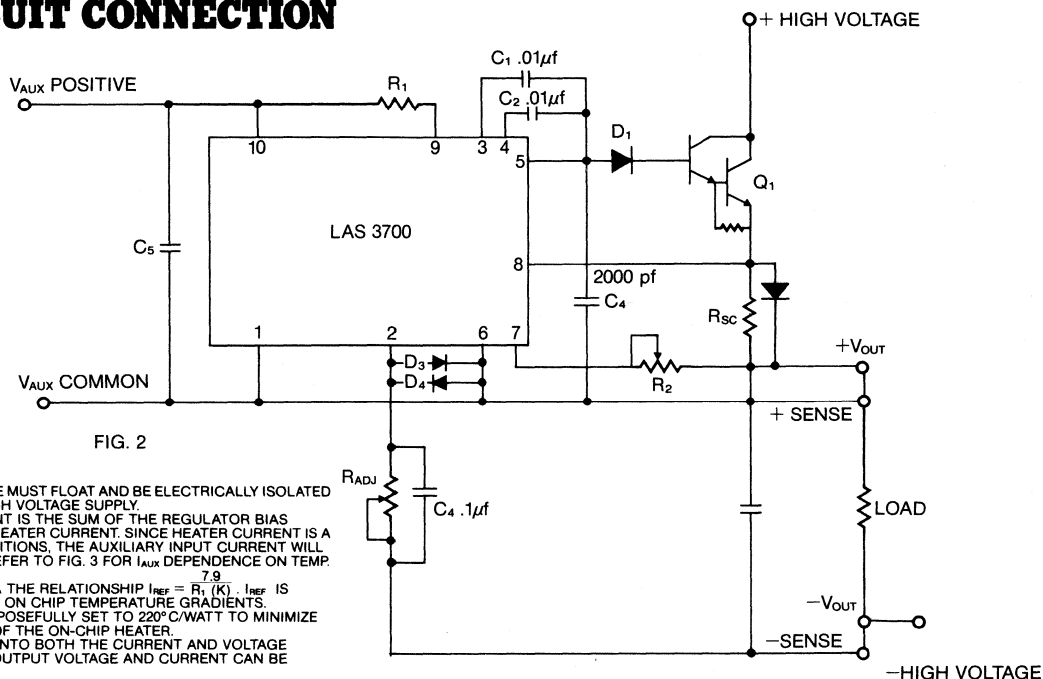
ORDERING INFORMATION

MODEL	PRICE QTY							
	1-24	25-49	50-99	100-249	250-499	500-999	1000-2499	2500-4999
LAS 3700	\$14.00	\$13.75	\$13.15	\$12.05	\$11.65	\$10.75	\$9.80	\$9.00

OPERATIONAL DATA



TYPICAL CIRCUIT CONNECTION



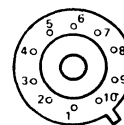
NOTES

1. THE AUXILIARY INPUT VOLTAGE MUST FLOAT AND BE ELECTRICALLY ISOLATED FROM THE UNREGULATED HIGH VOLTAGE SUPPLY.
2. THE AUXILIARY INPUT CURRENT IS THE SUM OF THE REGULATOR BIAS CURRENT AND THE ON CHIP HEATER CURRENT. SINCE HEATER CURRENT IS A FUNCTION OF AMBIENT CONDITIONS, THE AUXILIARY INPUT CURRENT WILL VARY OVER A WIDE RANGE. REFER TO FIG. 3 FOR I_{AUX} DEPENDENCE ON TEMP. AND VOLTAGE.
3. I_{REF} IS PROGRAMMED BY R_1 VIA THE RELATIONSHIP $I_{REF} = \frac{7.9}{R_1 (K)}$. I_{REF} IS LIMITED TO 1 mA TO MINIMIZE ON CHIP TEMPERATURE GRADIENTS.
4. THERMAL RESISTANCE IS PURPOSEFULLY SET TO 220° C/WATT TO MINIMIZE THE POWER REQUIREMENTS OF THE ON-CHIP HEATER.
5. AN OFFSET IS PROGRAMMED INTO BOTH THE CURRENT AND VOLTAGE AMPLIFIER TO INSURE BOTH OUTPUT VOLTAGE AND CURRENT CAN BE ADJUSTED TO ZERO.
6. THE AUXILIARY CURRENT SUPPLY SHOULD BE CAPABLE OF SUPPLYING A 150 mA TRANSIENT AT TURN ON TO INSURE RAPID HEATER TEMPERATURE STABILIZATION.

DESIGN CONSIDERATIONS

1. CONSTANT VOLTAGE—SELECT R_1 ACCORDING TO THE RELATIONSHIP $I_{REF1} = \frac{7.9}{R_1 (K)}$.
 $V_{OUT} = (I_{REF}) (R_{ADJ})$ I_{REF} IS LIMITED TO 1 mA MAX.
2. CONSTANT CURRENT—SELECT R_2 TO GIVE A DESIRED OFFSET ACCORDING TO THE RELATIONSHIP $V_{OFFSET} = (1 \text{ mA}) (R_2)$ —SELECT R_{SC} ACCORDING TO THE RELATIONSHIP $I_{LOAD \text{ RSC}} = V_{OFFSET}$ WHERE I_{LOAD} IS THE DESIRED CONSTANT CURRENT (R_2 IN K Ω).
3. D_1 , D_3 , AND D_4 ARE NECESSARY TO PROTECT THE LAS 3700 FROM TRANSIENT OR SHORT CIRCUIT CONDITIONS.
4. D_2 IS NECESSARY TO PROTECT THE 3700 FROM A FAILURE OF THE SERIES PASS ELEMENT.
5. C_1 , C_2 , AND C_4 ARE COMPENSATION CAPACITORS FOR THE VOLTAGE AND CURRENT AMPLIFIERS.
6. FOR OPTIMUM REGULATION, LIMIT THE CURRENT FROM PIN 5 TO AS LOW A VALUE AS POSSIBLE (1 mA MAX) TO PREVENT REGULATION DEGRADATION DUE TO THERMAL GRADIENT EFFECTS.
7. TO PREVENT EXCESSIVE ON CHIP HEATER CURRENTS, THE LAS 3700 SHOULD BE LOCATED IN AN AREA PROTECTED FROM AIR CURRENTS SUCH AS CAUSED BY COOLING FANS.
8. Q_1 MAY BE REPLACED BY A TRIPLE DARLINGTON OR OTHER GAIN CONFIGURATION IF MORE CURRENT IS REQUIRED. THE REGULATOR OUTPUT VOLTAGE IS LIMITED BY THE BV_{CEO} OF Q_1 .
9. C_4 IS NECESSARY IN APPLICATIONS REQUIRING VERY LOW OUTPUT NOISE.

TO-96 10-TERMINAL



Bottom View

LAS 2000 SERIES REGULATORS

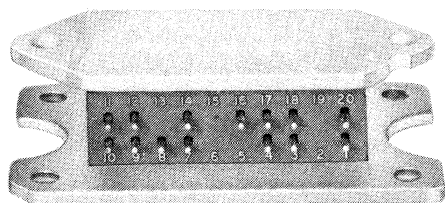
5 AMP, 85 WATT POSITIVE AND NEGATIVE

FEATURES

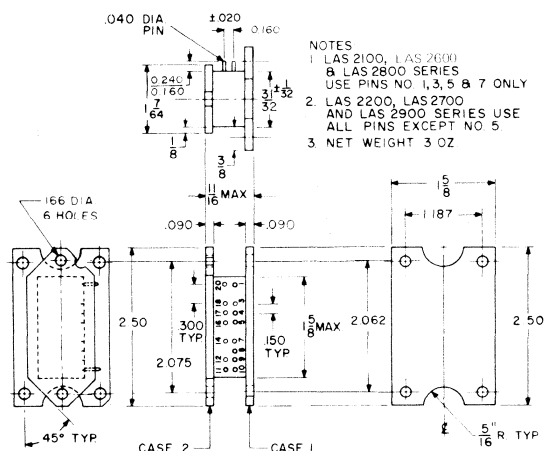
- 0.01% regulation — line
- 0.2% regulation — load
- 0.007%/°C temperature coefficient
- Short circuit and overload protection
- Thermal protection
- Remote programming
- Remote sensing
- Electrically isolated case
- Dual-tracking connection
- Increased power loading capability with external

DESCRIPTION

The LAS 2100, 2200 and 2700-2800 Series of Power Hybrid Voltage Regulators is designed for applications requiring a well regulated output voltage for load current variations up to 5.0 amperes. A key feature of the LAS series of Power Hybrid Voltage Regulators is its construction. A high degree of thermal isolation between the heat generating power elements and the heat sensitive control and reference elements is achieved by the placing of the power section on the heat-dissipating base of the unit, and the control stage on the heat-dissipating upper surface. This thermal isolation results in extremely low thermal drift characteristics for changes in power levels. In addition, a unique thermal power limiting circuit is built into the power section of the unit for increased operational reliability. This reliability is accentuated by a demonstrated MTBF of 100,000 hrs.



OUTLINE DRAWING



OUTLINE DRAWING, POWER HYBRID REGULATOR, LAS 2000 SERIES

ELECTRICAL CHARACTERISTICS

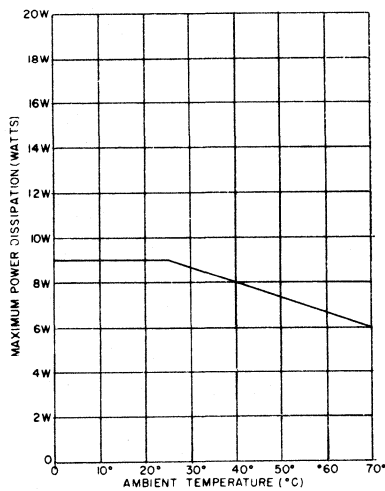
PARAMETER	SYMBOL	CONDITIONS	MIN.	MAX.	UNITS
Input voltage	V_{IN}		9.6	40.0	volts
Output voltage ⁽¹⁾	V_O		2.5	28.0	volts
Input-output differential ⁽²⁾⁽⁹⁾	$V_{IN}-V_O$		4.6 ⁽¹⁰⁾	37.5V	volts
*Input-output differential ⁽³⁾⁽⁹⁾	$V_{IN}-V_O$		2.5	37.5V	volts
Output current ⁽¹⁾	I_O		0	5.0	amps
Standby current	I_Q			10.0	mA
Power dissipation	P_D	Plate #1 @ 25°C		85	watts
Power dissipation ⁽⁶⁾	P_D	Free Air @ 25°C		9.0	watts
Thermal resistance junction—Case #1	θ_J-C1			2.0	°C/watt
Thermal resistance junction—Free Air	θ_J-FA			15.0	°C/watt
Storage temperature ⁽⁴⁾	T_S		-55	+125	°C
Power transistor junction temperature	T_J			+200	°C
Regulation line ⁽⁵⁾				0.01	%/ ΔV_{IN}
Regulation load ⁽⁶⁾				0.2	%
*Programming resistance				1000 nominal	ohms/volt
*Programming voltage				one/one	volt/volt
Temperature coefficient	T.C.			0.007	%/°C
Ripple attenuation ⁽⁷⁾		V_{IN} minimum I_O maximum	60		dB

NOTES: ⁽¹⁾ Varies with model number.
⁽²⁾ Single DC input voltage.
⁽³⁾ Separate DC input voltages for power circuit (pin 1) and control circuit (pin 20) V_{IN} minimum = 9.5 volts at pin 20.
⁽⁴⁾ Maximum storage temperature limited by tantalum capacitor.
⁽⁵⁾ I_O constant for entire range from V_{IN} minimum to V_{IN} maximum.
⁽⁶⁾ V_{IN} constant for entire range from 0 to full load.
⁽⁷⁾ Ripple attenuation is 54dB minimum for 20V, 24V, and 28V models.

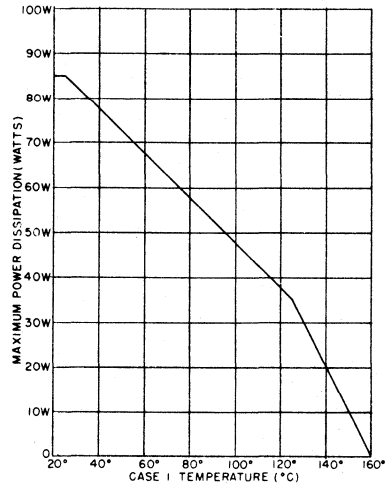
⁽⁸⁾ See figures on following pages.
⁽⁹⁾ Minimum input-output differential based on $T_J \geq 25^\circ C$.
⁽¹⁰⁾ 5.2 for Model LAS 2700 and LAS 2800 series.
⁽¹¹⁾ Voltages and currents are negative for models LAS 2700 and LAS 2800 series.
 *This parameter or note applies only to LAS 2200 and 2700 series.

LAS 2000 SERIES

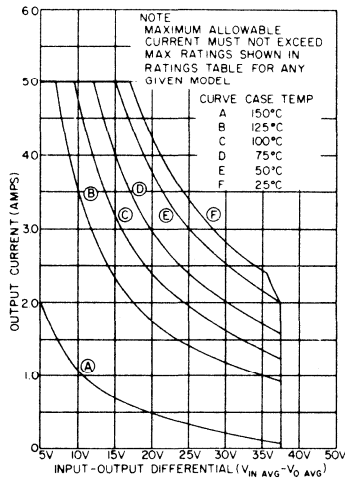
OPERATIONAL DATA



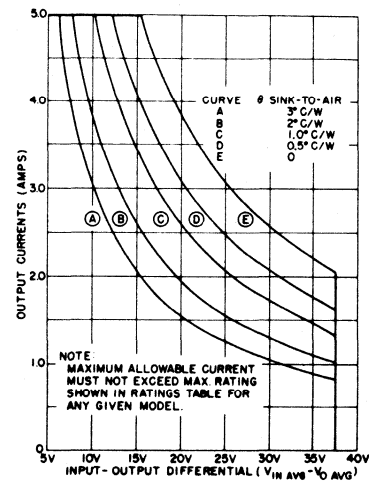
FREE AIR DERATING CURVE



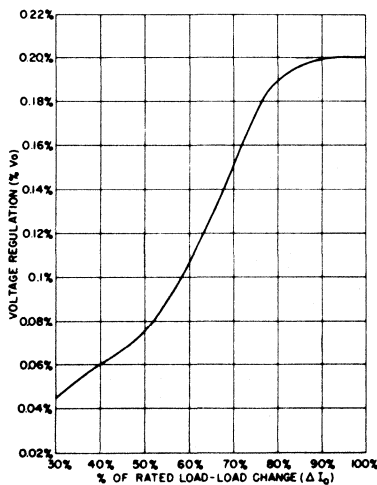
POWER DERATING CURVE AS A FUNCTION OF CASE 1 TEMPERATURE



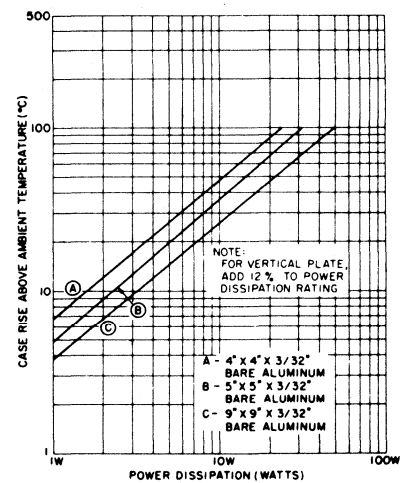
DC SAFE OPERATING AREA AS A FUNCTION OF MODULE CASE TEMPERATURE, FOR LAS 2100, LAS 2200, LAS 2600, LAS 2700, LAS 2800 AND LAS 2900 MODELS



DC SAFE OPERATING AREA AS A FUNCTION OF HEATSINK THERMAL RESISTANCE TO AIR AT 40°C AMBIENT TEMPERATURE, FOR LAS 2100, LAS 2200, LAS 2600, LAS 2700, LAS 2800, AND LAS 2900 MODELS



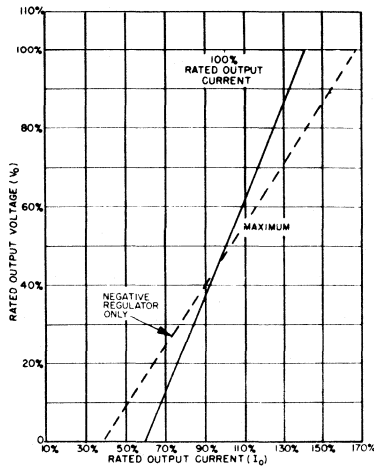
LOAD REGULATION IN % (VOLT) VS LOAD CHANGE



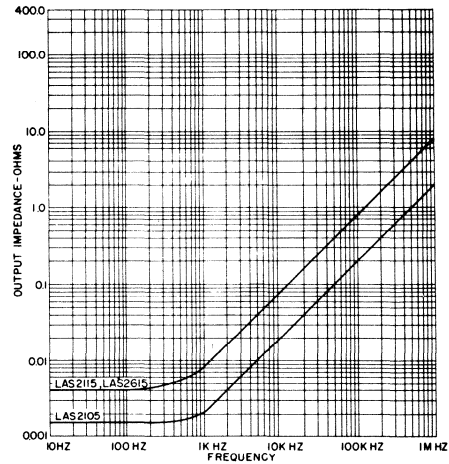
TYPICAL HEAT SINKING DATA FOR HORIZONTAL PLATE

LAS 2000 SERIES

OPERATIONAL DATA



SHORT CIRCUIT PROTECTION CHARACTERISTIC



TYPICAL OUTPUT IMPEDANCE VS FREQUENCY

ORDERING INFORMATION

LAS 2100, LAS 2200 SERIES 5 AMPS, 85 WATTS POSITIVE REGULATOR

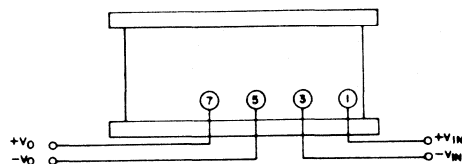
MODEL	V _o VOLTS	PRICE QTY						
		1-24	25-49	50-99	100-249	250-499	500-999	1000-2499
LAS 2105	5 fixed	\$47.00	\$47.00	\$35.00	\$31.00	\$31.00	\$26.00	\$24.00
LAS 2205	5 ADJ	47.00	43.00	35.00	31.00	31.00	26.00	24.00
LAS 2206	6 ADJ	47.00	43.00	35.00	31.00	31.00	26.00	24.00
LAS 2112	12 fixed	47.00	43.00	35.00	31.00	31.00	26.00	24.00
LAS 2212	12 ADJ	47.00	43.00	35.00	31.00	31.00	26.00	24.00
LAS 2115	15 fixed	47.00	43.00	35.00	31.00	31.00	26.00	24.00
LAS 2215	15 ADJ	47.00	43.00	35.00	31.00	31.00	26.00	24.00
LAS 2124	24 fixed	47.00	43.00	35.00	31.00	31.00	26.00	24.00
LAS 2224	24 ADJ	47.00	43.00	35.00	31.00	31.00	26.00	24.00
LAS 2228	28 ADJ	47.00	43.00	35.00	31.00	31.00	26.00	24.00

LAS 2700, LAS 2800 SERIES 5 AMPS, 85 WATTS NEGATIVE REGULATOR

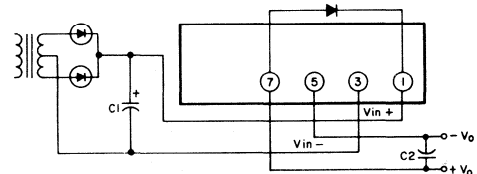
LAS 2705	- 5ADJ	\$47.00	\$44.00	\$36.00	\$32.00	\$32.00	\$28.00	\$26.00
LAS 2812	-12 fixed	47.00	44.00	36.00	32.00	32.00	28.00	26.00
LAS 2712	-12 ADJ	47.00	44.00	36.00	32.00	32.00	28.00	26.00
LAS 2715	-15 ADJ	47.00	44.00	36.00	32.00	32.00	28.00	26.00
LAS 2724	-24 ADJ	47.00	44.00	36.00	32.00	32.00	28.00	26.00

CONNECTION DIAGRAMS

POSITIVE REGULATOR



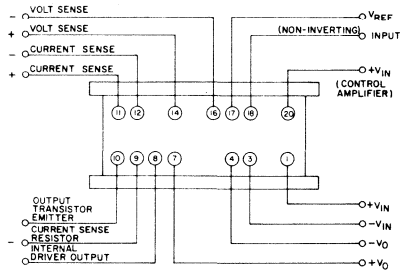
4 PIN POWER HYBRID VOLTAGE REGULATOR



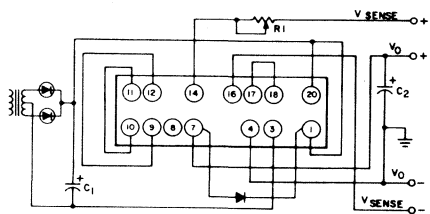
4 PIN POWER HYBRID VOLTAGE REGULATOR CIRCUIT

LAS 2000 SERIES

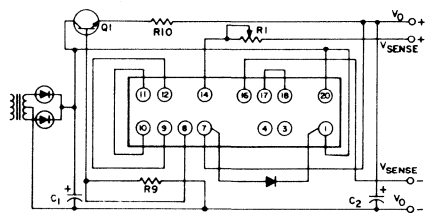
CONNECTION DIAGRAMS POSITIVE REGULATOR



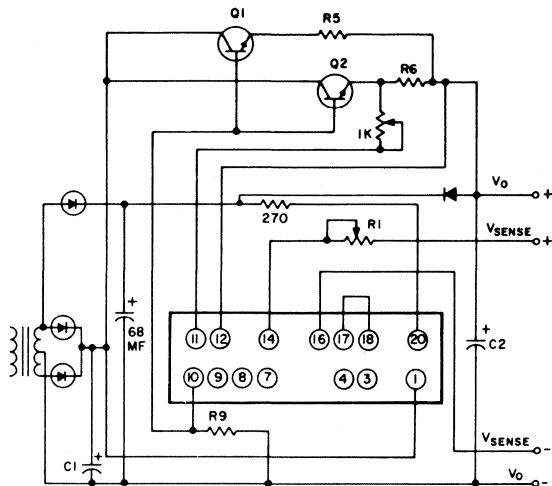
14 PIN POWER HYBRID VOLTAGE REGULATOR



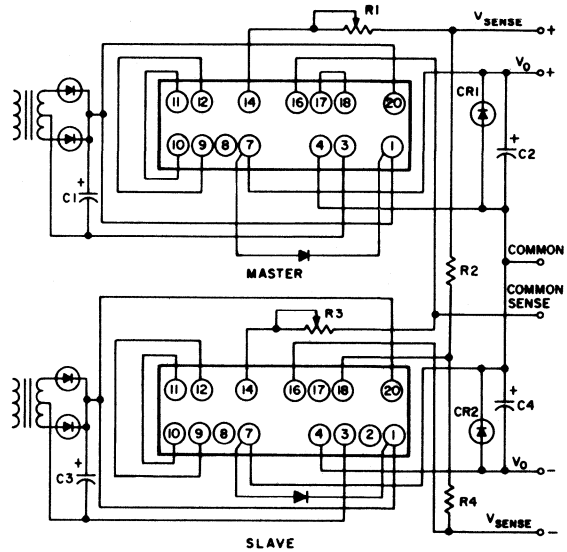
POSITIVE POWER HYBRID VOLTAGE REGULATOR CIRCUIT



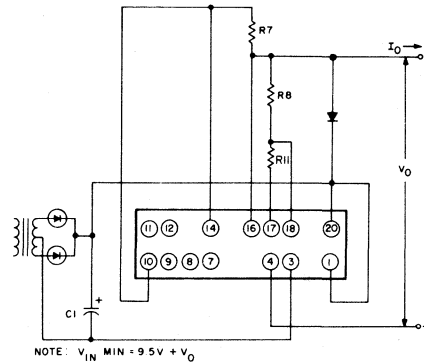
POWER HYBRID VOLTAGE REGULATOR USED WITH PARALLEL PASS TRANSISTOR FOR HIGHER OUTPUT CURRENT



POWER HYBRID VOLTAGE REGULATOR USED AS A DRIVER FOR HIGHER CURRENT OUTPUTS USING PEAK DETECTOR FOR CONTROL AMPLIFIER INPUT VOLTAGE

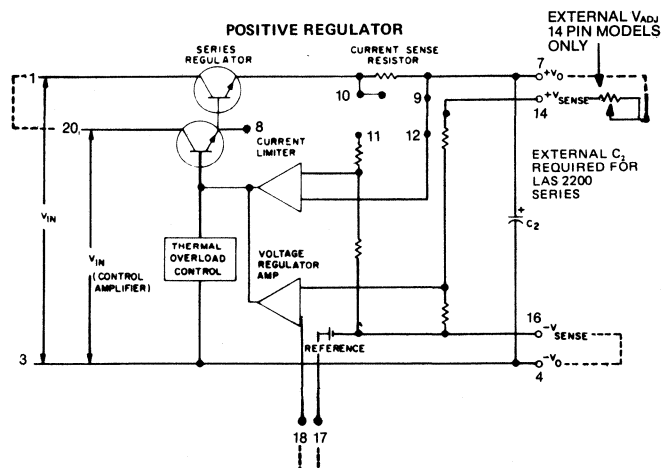


DUAL TRACKING POWER HYBRID VOLTAGE REGULATOR CIRCUIT



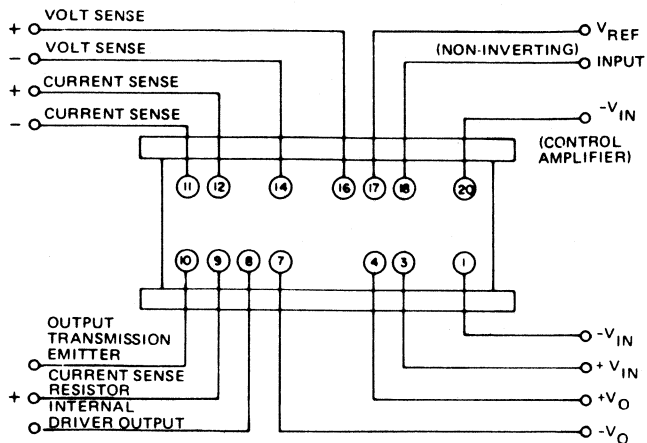
POWER HYBRID CURRENT REGULATOR CIRCUIT

FUNCTIONAL BLOCK DIAGRAM

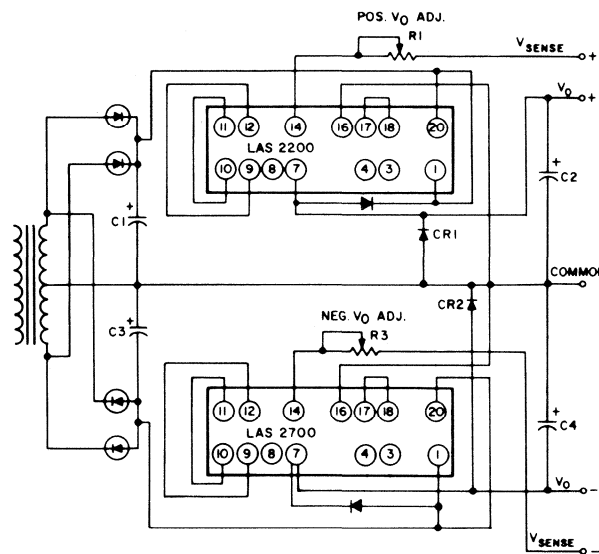


LAS 2000 SERIES

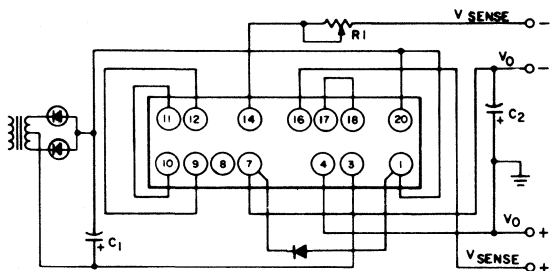
NEGATIVE REGULATORS CONNECTION DIAGRAMS



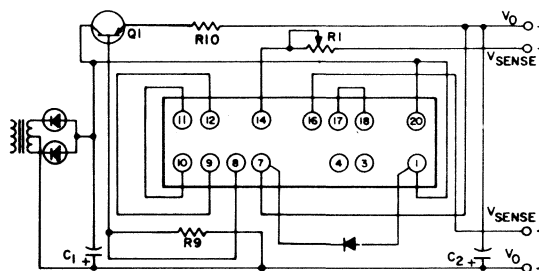
14 PIN POWER HYBRID VOLTAGE REGULATOR



DUAL, POS/NEG REGULATORS USING COMPLEMENTARY REGULATOR WITH COMMON INPUT

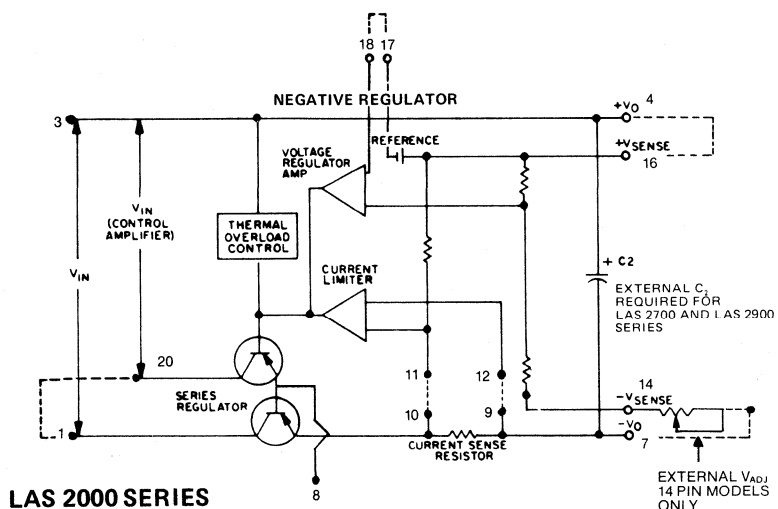


NEGATIVE POWER HYBRID VOLTAGE REGULATOR CIRCUIT



POWER HYBRID VOLTAGE REGULATOR USED WITH PARALLEL PASS TRANSISTOR FOR HIGHER OUTPUT CURRENT

FUNCTIONAL BLOCK DIAGRAM

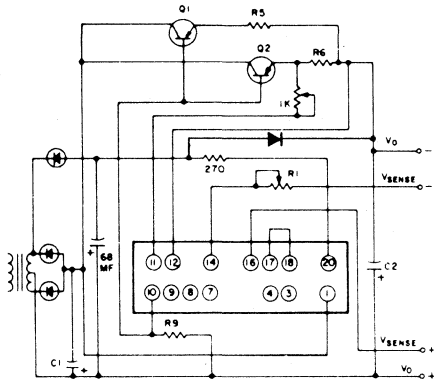


LAS 2000 SERIES

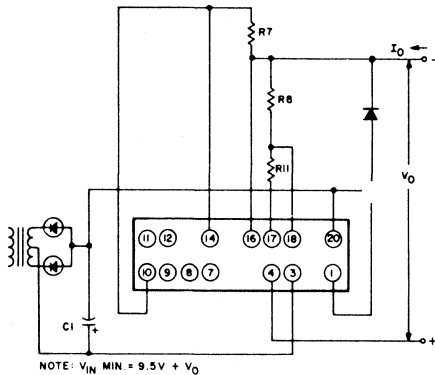
External Jumpers Required

LAS 2000 SERIES

CONNECTION DIAGRAMS NEGATIVE REGULATORS



**POWER HYBRID VOLTAGE REGULATOR
USED AS A DRIVER FOR HIGHER
CURRENT OUTPUTS USING PEAK DE-
TECTOR FOR CONTROL AMPLIFIER
INPUT VOLTAGE**



**POWER HYBRID CURRENT REGULATOR
CIRCUIT**

NOTES

- Minimum value of input filter capacitors C1 and C3 is determined by: $C1, C3 = I_O$ (1000 mfd/amp) recommended.
- Minimum value of output capacitors C2 and C4 is determined by $C2, C4 = I_O$ (100mfd/amp).
- Minimum value of output voltage adjust resistors R1 and R3 for LAS 2205 and LAS2905 is 3K ohms. See note 4 to determine value for all other models.
- Minimum value of output voltage adjust resistors R1 and R3 is determined by: $R1, R3 = (0.25V_O \times 1000\Omega/V)$ ohms wirewound. Use next highest standard value.
- Values of tracking reference voltage divider resistors R2 and R4 for all models except LAS 2205, and 2206 are determined by:
 - $R2 = (2000V_O - 7150)$ ohms, $\pm 1\%$, $\frac{1}{2}W$ film
 - $R4 = 7.15K$ ohms, $\pm 1\%$, $\frac{1}{2}W$ film
- Value of tracking reference voltage divider resistor R2 and R4 for LAS 2205 is:
 - $R2 = 7.50K \pm 1\%$ $\frac{1}{2}W$ film
 - $R4 = 2.43K \pm 1\%$ $\frac{1}{2}W$ film
- Value of tracking reference voltage divider resistor R2 and R4 for LAS 2206 and LAS 2406 is:
 - $R2 = 8.06K \pm 1\%$ $\frac{1}{2}W$ film
 - $R4 = 4.02K \pm 1\%$ $\frac{1}{2}W$ film
- Values of current sharing resistors R5 and R6 are determined by: $R5, R6 = (N \times 0.5V)/MAX I_O$ ohms $\pm 3\%$ wirewound where N = number of emitter current sharing resistors required.
- Nominal value of the current sharing resistor R10 for Figs. 58 & 64 is determined from table below:

R10

Model

.100 Ω	LAS 2205, 2206, 2212, 2215, 2705, 2712, 2715
.150 Ω	LAS 2220
.200 Ω	LAS 2224, 2724
.250 Ω	LAS 2228, 2728

- Value of current sensing resistor R7 is determined by $R7 = (2.4V)/I_O$ ohms, nominal, wirewound.
- R8 (part of reference voltage divider) = 750 ohms nominal, $\frac{1}{2}W$ film resistor for all models except LAS 2205 and 2206.
- R8 for LAS 2206 is 3.2K nominal, $\frac{1}{2}W$ film.
- R8 for LAS 2205 is not required.
- Rectifiers CR1 and CR2 should be rated at peak inverse voltage of 50 V and forward current equal at least to maximum rated I_O .
- Value of I_{cbo} drain resistor R9, is determined by: $R9 = (V_O)/(N \times MAX I_{cbo})$ ohms, $\pm 5\%$ composition where N = number of external series pass transistors.
- All fixed resistors shown on diagrams with given values in ohms are $\frac{1}{4}W$ composition.
- Temperature rise of case 2, ΔTC_2 , is given by the following:
 - For no external heat sink on case 2.
 $\Delta TC_2 = 0.25P_D\theta_{J1-A}$
 - For an external heat sink on case 2 with thermal resistance θ

$$\Delta TC_2 = \frac{\theta - HS_2}{50^\circ C/W + \theta HS_2} \times P_D \theta_{J1-A}$$

- Nominal values of R8 and R11 $\frac{1}{2}W$ film resistors (Fig. 66):

R8 Ohms	R11 Ohms	Model
∞ (open)	0(short)	LAS 2705,
15,500	4,500	LAS 2706;
9,700	12,300	LAS 2712,
9,700	11,700	LAS 2715,
9,700	10,700	LAS 2724
9,700	10,300	LAS 2728

LAS 3000 SERIES REGULATOR

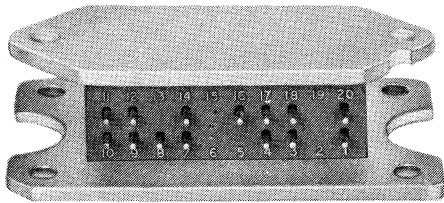
10 AMP, 140 WATT POSITIVE REGULATOR

FEATURES

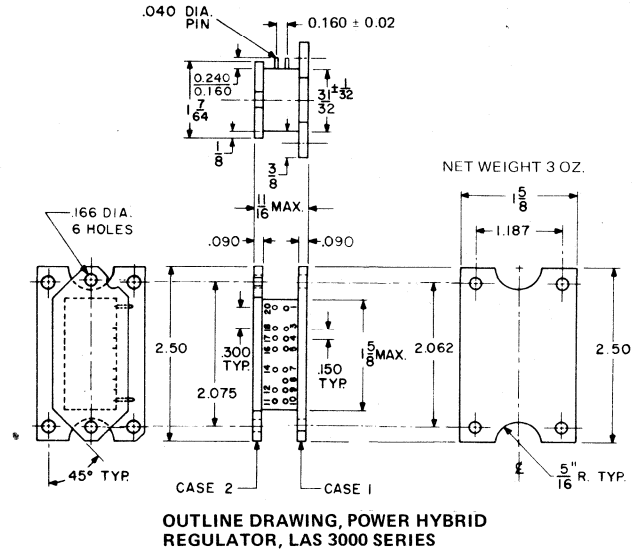
- 0.01% regulation line
- 0.2% regulation load
- 0.015%/°C temperature coefficient
- Short circuit and overload protection
- Remote programming
- Remote sensing
- Electrically isolated case
- Dual-tracking connection

DESCRIPTION

The LAS 3200 Series of Power Hybrid Voltage Regulators is designed for applications requiring a well regulated output voltage for load current variations up to 10 amperes. A key feature of the LAS series of Power Hybrid Voltage Regulators is its construction. A high degree of thermal isolation between the heat generating power elements and the heat sensitive control and reference elements is achieved by the placing of the power section on the heat-dissipating base of the unit, and the control stage on the heat-dissipating upper surface. This thermal isolation results in extremely low thermal drift characteristics for changes in power levels.



OUTLINE DRAWING



ELECTRICAL CHARACTERISTICS

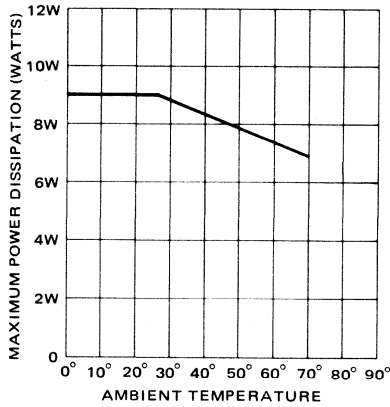
PARAMETER	SYMBOL	CONDITION	MIN.	MAX.	UNITS
Input voltage to pin (1) ^(A) ^(G)	$V_{IN}(1)$		5.2	40.0	volts
Input voltage to pin (20) ^(A)	$V_{IN}(20)$		7.9	40.0	volts
Output voltage	V_O		2.7	29.4	volts
Input-output differential ^{(A) (F)}	$(V_{IN}(1) - V_O)$		2.5	28.6	volts
Input-output differential ^{(B) (F)}	$(V_{IN}(20) - V_O)$		5.2	28.6	volts
Output current	I_O			10.0	amps
Standby current	$I_O(1)$			20.0	mA
Standby current	$I_O(20)$			7.0	mA
Power dissipation	P_D	Plate #1 @ 25°C		140	watts
Power dissipation	P_D	Free Air @ 25°C Amb		9	watts
Thermal resistance junction—Case #1	$\theta_J - C1$			1.25	°C/watt
Thermal resistance junction—Free Air	$\theta_J - FA$			15	°C/watt
Storage temperature	T_S		-55	+150	°C
Power transistor junction temperature	T_J			+200	°C
Regulation line ^(C)				0.01	%/ ΔV_{IN}
Regulation load ^(D)				0.2	%
Programming resistance				1000 nominal	ohms/volt
Programming voltage				one/one	volt/volt
Temperature coefficient				0.015	%/°C
Ripple attenuation ^(E)		$V_{IN}(1)$ minimum I_O maximum	60		dB

NOTES:

- ^(A) Separate DC input voltages for power circuit (pin 1) and control circuit pin (20).
- ^(B) Common input voltages for power circuit (pin 1) and control pin (20).
- ^(C) I_O constant for entire input voltage range from $[V_{IN}(1) \& V_{IN}(20) \text{ min.}]$ to $[V_{IN}(1) (20) \text{ max.}]$
- ^(D) V_{IN} constant for entire range from 0 to full load.
- ^(E) Ripple attenuation is 54 dB min. for 24V and 28V models.
- ^(F) Minimum input-output differential based on $T_J > 25^\circ\text{C}$.
- ^(G) Maximum input voltage is 30V for LAS 3205 and LAS 3206.

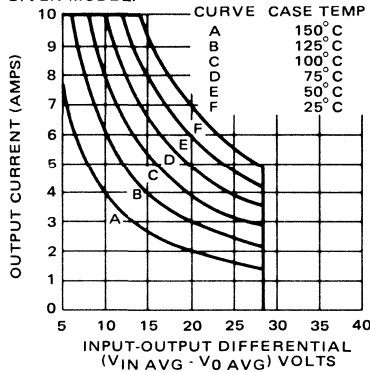
LAS 3000 SERIES

OPERATIONAL DATA

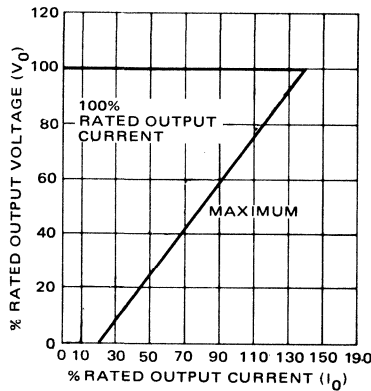


FREE AIR DERATING CURVE

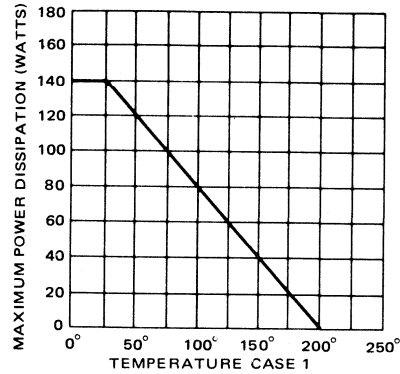
NOTE: MAXIMUM ALLOWABLE CURRENT MUST NOT EXCEED MAX. RATINGS SHOWN IN RATINGS TABLE FOR ANY GIVEN MODEL.



DC SAFE OPERATING AREA AS A FUNCTION OF MODULE CASE TEMPERATURE, FOR LAS 3000 SERIES

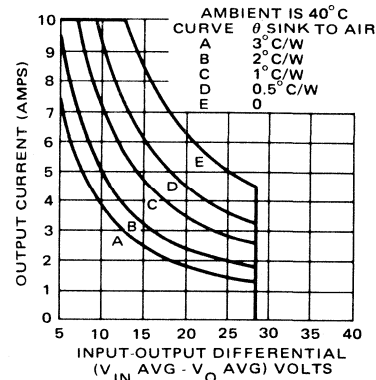


SHORT CIRCUIT PROTECTION CHARACTERISTIC

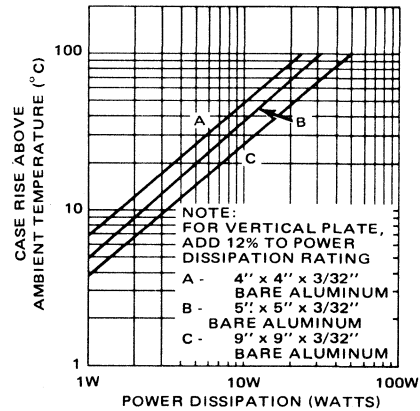


POWER DERATING CURVE AS A FUNCTION OF CASE 1 TEMPERATURE

NOTE: MAXIMUM ALLOWABLE CURRENT MUST NOT EXCEED MAX. RATINGS SHOWN IN RATINGS TABLE FOR ANY GIVEN MODEL.



DC SAFE OPERATING AREA AS A FUNCTION OF HEATSINK THERMAL RESISTANCE TO AIR AT 40°C AMBIENT TEMPERATURE FOR LAS 3000 SERIES



TYPICAL HEAT SINKING DATA FOR HORIZONTAL PLATE

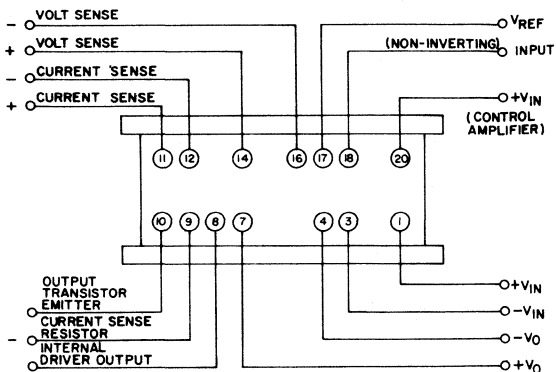
ORDERING INFORMATION

LAS 3200 10 AMPS 140 WATTS

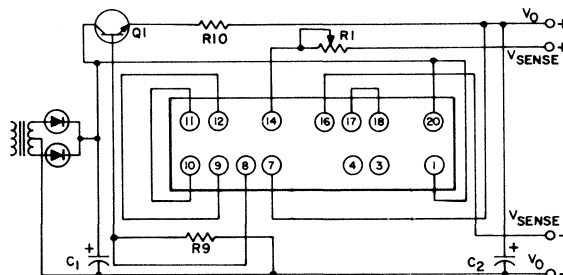
MODEL	V_0 —ADJ VOLTS	PRICE QTY						
		1-24	25-49	50-99	100-249	250-499	500-999	1000-2499
LAS 3205	5	\$54.00	\$47.00	\$41.00	\$35.00	\$35.00	\$30.00	\$29.00
LAS 3212	12	54.00	47.00	41.00	35.00	35.00	30.00	29.00
LAS 3215	15	54.00	47.00	41.00	35.00	35.00	30.00	29.00
LAS 3224	24	54.00	47.00	41.00	35.00	35.00	30.00	29.00
LAS 3228	28	54.00	47.00	41.00	35.00	35.00	30.00	29.00

LAS 3000 SERIES

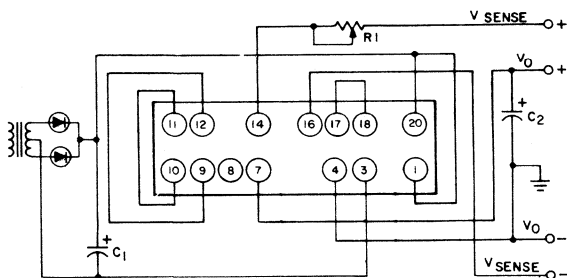
CONNECTION DIAGRAMS



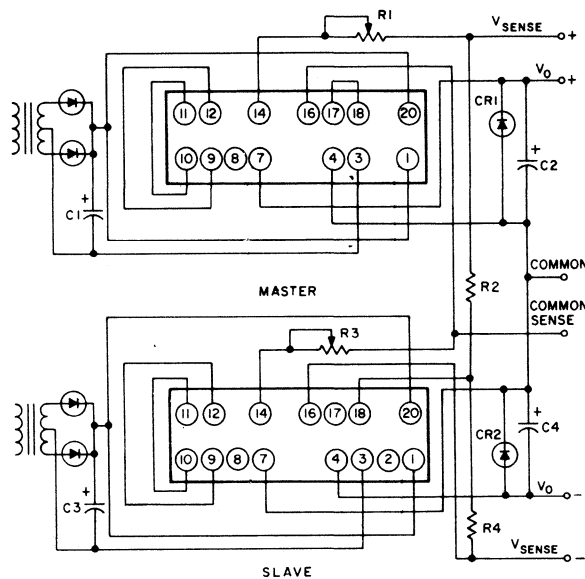
14 PIN POWER HYBRID VOLTAGE REGULATOR



POWER HYBRID VOLTAGE REGULATOR USED WITH PARALLEL PASS TRANSISTOR FOR HIGHER OUTPUT CURRENT

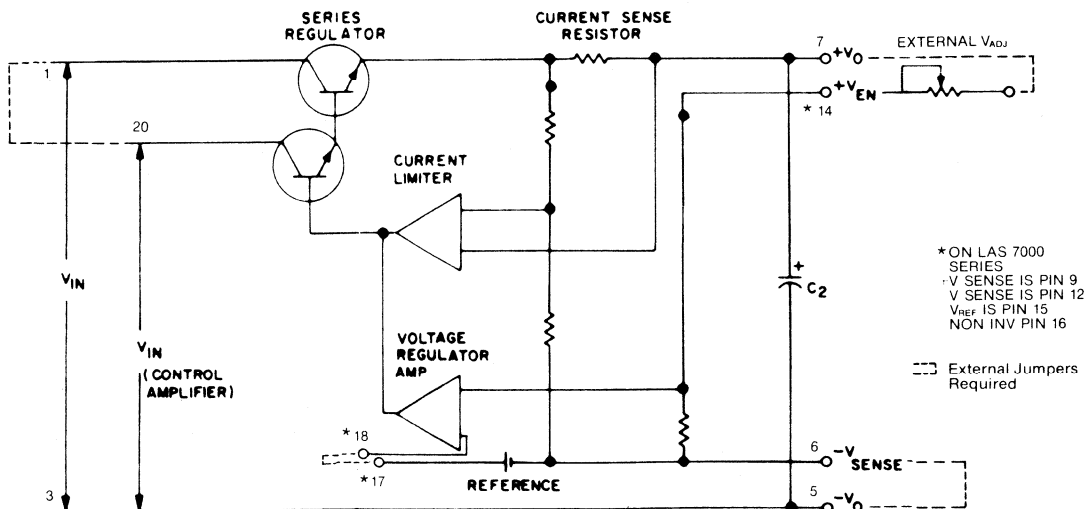


POSITIVE POWER HYBRID VOLTAGE REGULATOR CIRCUIT



DUAL TRACKING POWER HYBRID VOLTAGE REGULATOR CIRCUIT

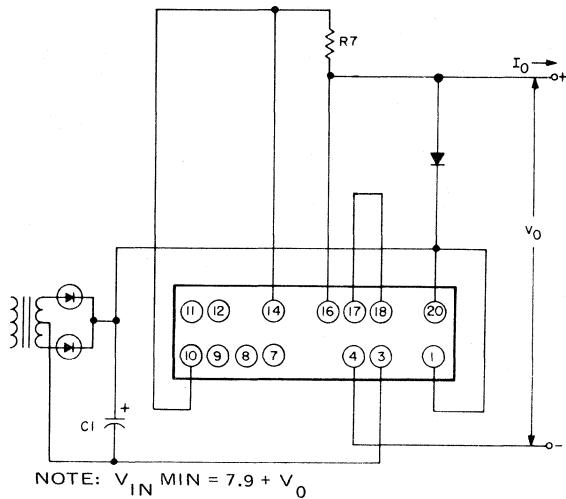
FUNCTIONAL BLOCK DIAGRAM



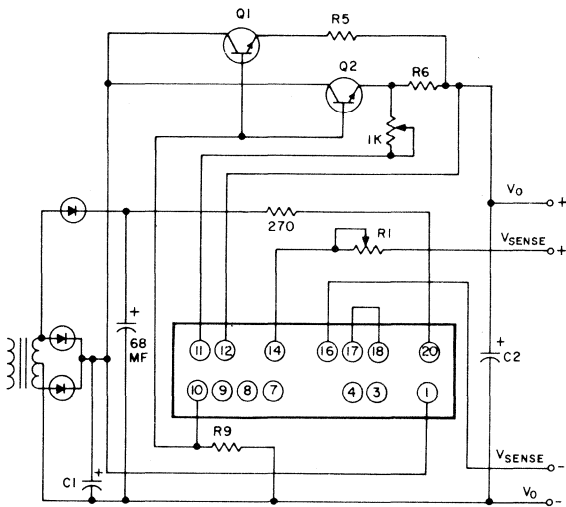
LAS 3000, LAS 5000 AND LAS 7000
 SEE LAS 2000 SERIES FOR LAS 3000 PIN CORRECTION.

LAS 3000 SERIES

CONNECTION DIAGRAMS



**POWER HYBRID CURRENT REGULATOR
CIRCUIT**



**POWER HYBRID VOLTAGE REGULATOR
USED AS A DRIVER FOR HIGHER
CURRENT OUTPUTS USING PEAK DE-
TECTOR FOR CONTROL AMPLIFIER
INPUT VOLTAGE**

NOTES

1. Minimum value of input filter capacitors C1 and C3 is determined by: $C1, C3 = I_o \times (1000 \text{ mfd/amp})$ recommended.
2. Minimum value of output capacitors C2 and C4 is determined by: $C2, C4 = I_o \times (100 \text{ mfd/amp})$.
3. Minimum value of output voltage adjust resistors R1 and R3 for LAS 3205 is 3K ohms. See note 4 to determine value for all other models.
4. Minimum value of output voltage adjust resistors R1 and R3 is determined by: $R1, R3 = (0.25V_o \times 1000 \Omega/V)$ ohms
5. Values of tracking reference voltage divider resistors R2 and R4 for all models are determined by:
 - a) $R2 = (2000V_o - 2490)$ ohms $\pm 1\%$, $\frac{1}{2}W$ film
 - b) $R4 = 2.49K$ ohms, $\pm 1\%$, $\frac{1}{2}W$ film
6. Values of current sharing resistors R5 and R6 are determined by: $R5, R6 = (N \times 0.5V)/\text{MAX } I_o$ ohms $\pm 3\%$ wirewound where N = number of emitter current sharing resistors required.
7. Nominal value of the current sharing resistor R10 for Fig. 85 is determined from table below:

R10	Model
.05 Ω	LAS 3205, 3206
.064 Ω	LAS 3212, 3215, 3224
.075 Ω	LAS 3228

8. Value of current sensing resistor R7 is determined by $R7 = 2.5V/I_o$ ohms, nominal, wirewound.
9. Rectifiers CR1 and CR2 should be rated at peak inverse voltage of 50V and forward current equal at least to maximum rated I_o .
10. Value of I_{cb0} drain resistor R9, is determined by $R9 = (V_o)/(N \times \text{MAX } I_{cb0})$ ohms, $\pm 5\%$ composition where N = number of external series pass transistors.
11. All fixed resistors shown on diagrams with given values in ohms are $\frac{1}{4}W$ composition.
12. Temperature rise of Case 2, ΔTC_2 , is given by the following:
 - a) For no external heat sink on case 2.
 $\Delta TC_2 = 0.25 P_D \theta_{J1-A}$
 - b) For an external heat sink on case 2 with thermal resistance θ

$$\Delta TC_2 = \frac{\theta \cdot HS_2}{50^\circ C/W + \theta HS_2} \times P_D \theta_{J1-A}$$

LAS 5000 SERIES REGULATOR

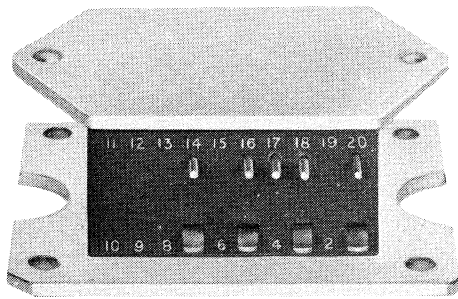
20 AMP, 270 WATT POSITIVE

FEATURES

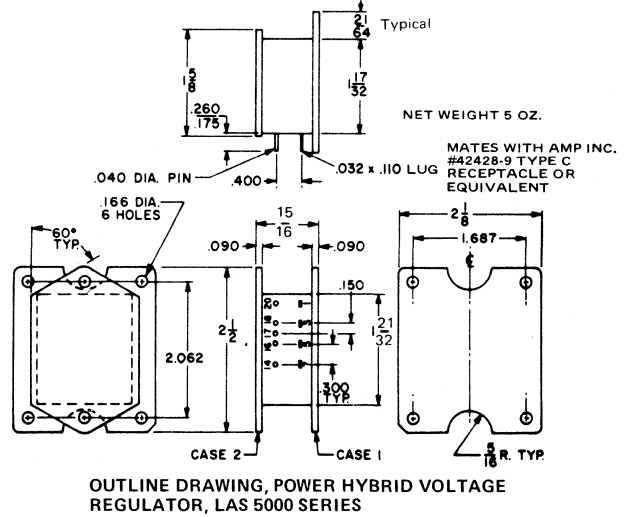
- 0.01% regulation line
- Remote programming
- 0.2% regulation load
- Remote sensing
- 0.015%/° C temperature coefficient
- Electrically isolated case
- Short circuit and overload protection
- Dual-tracking connection

DESCRIPTION

The LAS 5200 Series of Power Hybrid Voltage Regulators is designed for applications requiring a well regulated output voltage for load current variations up to 20 amperes. A key feature of the LAS series of Power Hybrid Voltage Regulators is its construction. A high degree of thermal isolation between the heat generating power elements and the heat sensitive control and reference elements is achieved by the placing of the power section on the heat-dissipating base of the unit, and the control stage on the heat-dissipating upper surface. This thermal isolation results in extremely low thermal drift characteristics for changes in power levels.



OUTLINE DRAWING



ELECTRICAL CHARACTERISTICS

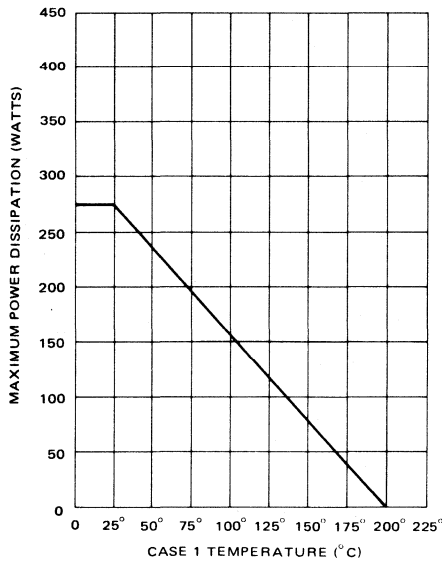
PARAMETER	SYMBOL	CONDITION	MIN.	MAX.	UNITS
Input voltage to pin (1) ^{(A) (H)}	V _{IN} (1)		7.25	40.0	volts
Input voltage to pin (20) ^{(A) (G)}	V _{IN} (20)		11.9	40.0	volts
Output voltage	V _O		4.75	29.4	volts
Input-output differential ^{(A) (F)}	(V _{IN} (1)-V _O)		2.5	28.6	volts
Input-output differential ^{(B) (F)}	(V _{IN} (20)-V _O)		7.20	28.6	volts
Output current	I _O			20.0	amps
Standby current	I _O (1)			30.0	mA
Standby current	I _O (20)			7.0	mA
Power dissipation	P _D	Plate #1 @ 25° C		270	watts
Power dissipation	P _D	Free Air @ 25° C Amb		11	watts
Thermal resistance junction— Case #1	θ _{J-C1}			0.65	°C/watt
Thermal resistance junction— Free Air	θ _{J-FA}			12.0	°C/watt
Storage temperature	T _S		-55	+125	°C
Power transistor junction temperature	T _J			+200	°C
Regulation line ^(C)				0.014	%/ΔV _{IN}
Regulation load ^(D)				0.2	%
Programming resistance				1000 nominal	ohms/volt
Programming voltage				one/one	volt/volt
Temperature coefficient				0.015	%/°C
Ripple attenuation ^(E)		V _{IN} (1) minimum I _O maximum	60		dB

NOTES:

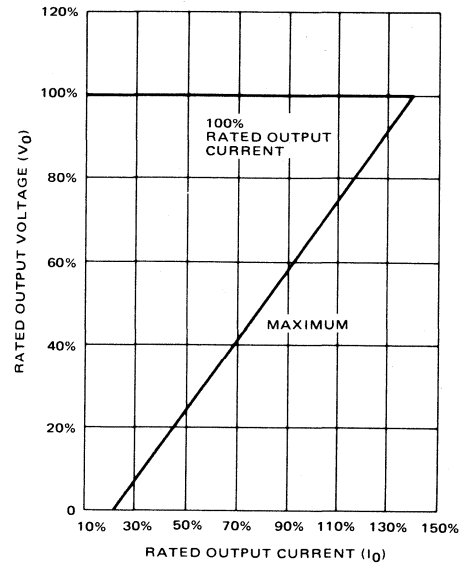
^(A) Separate DC input voltages for power circuit (pin 1) and control circuit pin (20).
^(B) Common input voltages for power circuit (pin 1) and control pin (20).
^(C) I_O constant for entire input voltage range from [V_{IN} (1) & V_{IN} (20) min.] to [V_{IN} (1) (20) max.].
^(D) V_{IN} constant for entire range from 0 to full load.
^(E) Ripple attenuation is 54 dB min. for 24V and 28V models.
^(F) Minimum input-output differential based on T_J > 25° C.
^(G) For AC source to Pin 20 with source resistance less than 10 ohms, minimum VAC = 12V rms. For other conditions consult factory.
^(H) Maximum input voltage is 30V for LAS 5205 and 5206.

LAS 5000 SERIES

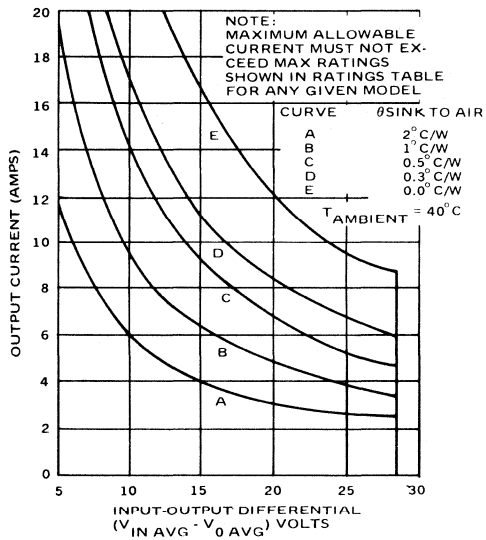
OPERATIONAL DATA



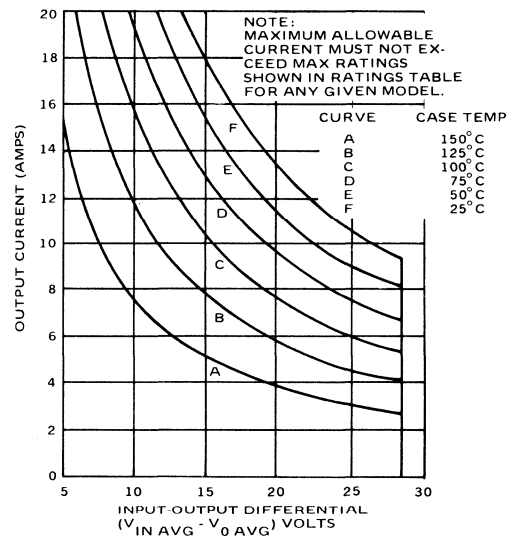
POWER DERATING CURVE AS A FUNCTION OF CASE 1 TEMPERATURE



SHORT CIRCUIT PROTECTION CHARACTERISTIC, LAS 5000 SERIES



DC SAFE OPERATING AREA AS A FUNCTION OF HEATSINK THERMAL RESISTANCE TO AIR AT 40° AMBIENT TEMPERATURE FOR LAS 5000 SERIES



DC SAFE OPERATING AREA AS A FUNCTION OF MODULE CASE TEMPERATURE FOR LAS 5000 SERIES

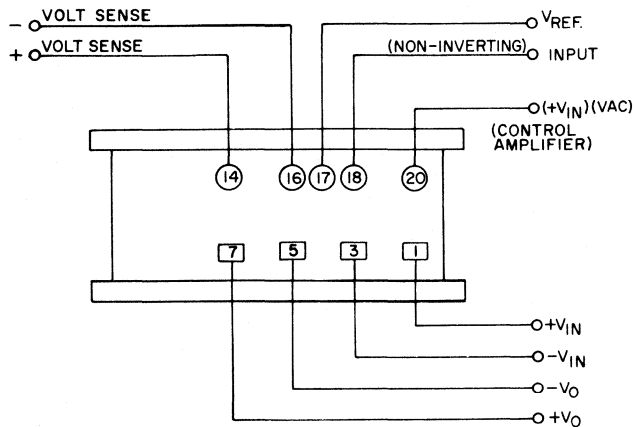
ORDERING INFORMATION

LAS 5200 20 AMPS, 270 WATTS

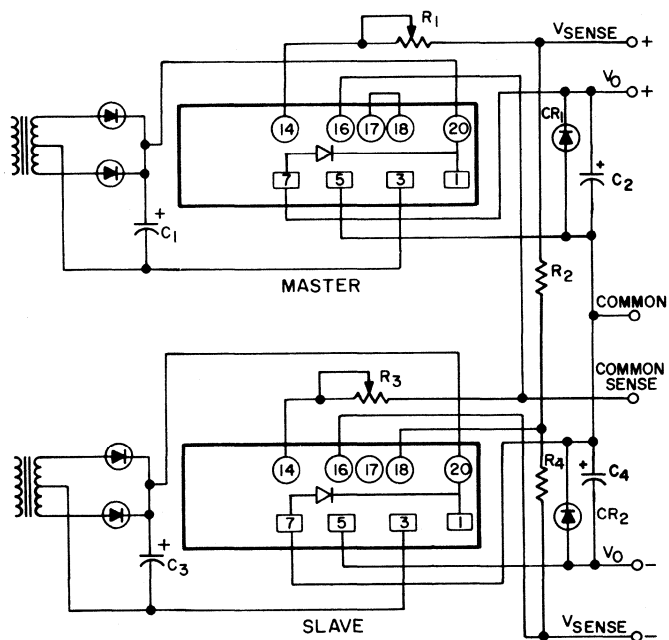
MODEL	V ₀ -ADJ VOLTS	PRICE QTY						
		1-24	25-49	50-99	100-249	250-499	500-999	1000-2499
LAS 5205	5	\$112.00	\$91.00	\$77.00	\$65.00	\$65.00	\$58.00	\$54.00
LAS 5212	12	112.00	91.00	77.00	65.00	65.00	58.00	54.00
LAS 5224	24	112.00	91.00	77.00	65.00	65.00	58.00	54.00

LAS 5000 SERIES

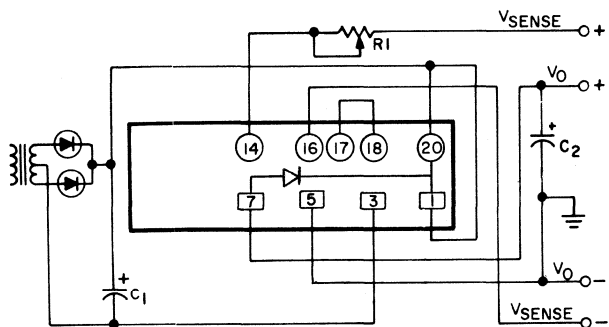
CONNECTION DIAGRAMS



9-PIN POWER HYBRID VOLTAGE REGULATOR



DUAL TRACKING POWER HYBRID VOLTAGE REGULATOR CIRCUIT

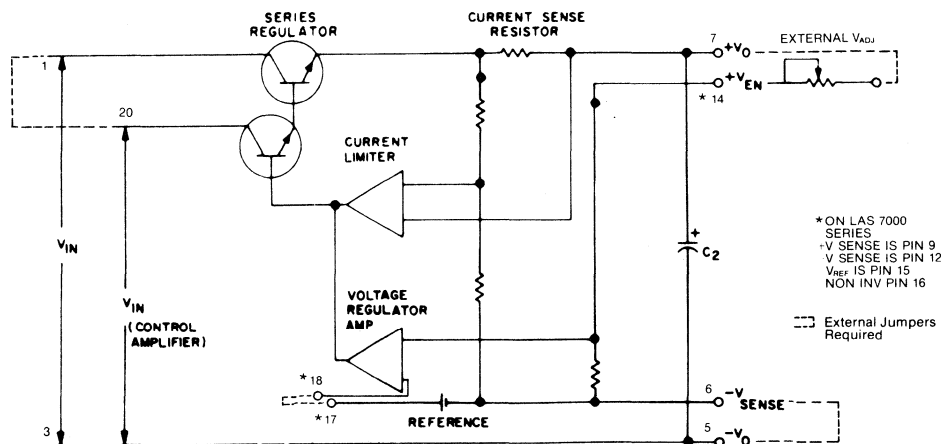


POSITIVE POWER HYBRID VOLTAGE REGULATOR CIRCUIT

NOTES

- Minimum value of input filter capacitors C1 and C3 is determined by: $C_1, C_3 = I_o \times (1000 \text{ mfd/amp})$ recommended.
- Minimum value of output capacitors C2 and C4 is determined by: $C_2, C_4 = I_o \times (100 \text{ mfd/amp})$.
- Minimum value of output voltage adjust resistors R1 and R3 is determined by: $R_1, R_3 = (0.25V_o \times 1000\Omega/V)$ ohms wirewound. Use next highest standard value.
- Values of tracking reference voltage divider resistors R2 and R4 for all models are determined by:
 - $R_2 = (2000V_o - 2490)$ ohms, $\pm 1\%$, $\frac{1}{2}W$ film
 - $R_4 = 2.49K$ ohms, $\pm 1\%$, $\frac{1}{2}W$ film
- Rectifiers CR1 and CR2 should be rated at peak inverse voltage of 50V and forward current equal at least to maximum rated I_o .

FUNCTIONAL BLOCK DIAGRAM



LAS 3000, LAS 5000 AND LAS 7000
SEE LAS 2000 SERIES FOR LAS 3000 PIN CORRECTION.

LAS 7000 SERIES REGULATOR

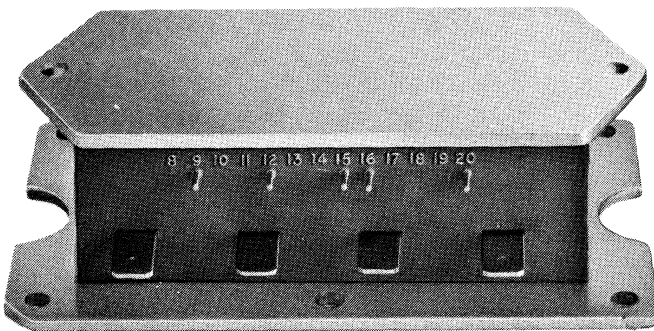
30 AMP 400 WATT POSITIVE REGULATOR

FEATURES

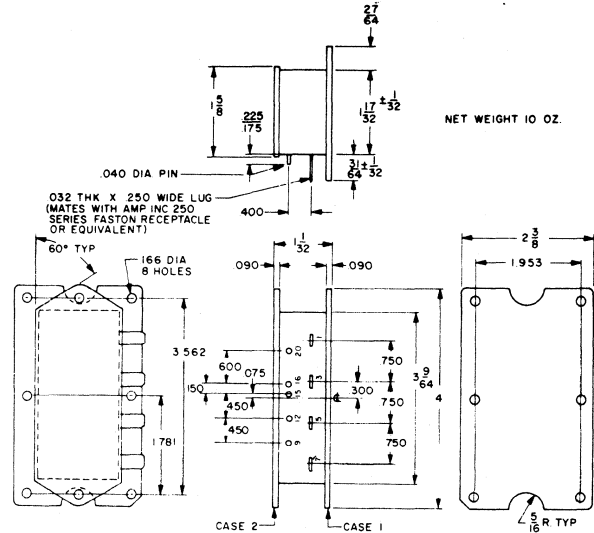
- 0.016% regulation line
- 0.2% regulation load
- 0.015%/°C temperature coefficient
- Short circuit and overload protection
- Remote programming
- Remote sensing
- Electrically isolated case
- Dual-tracking connection

DESCRIPTION

The LAS 7200 Series of Power Hybrid Voltage Regulators is designed for applications requiring a well regulated output voltage for load current variations up to 30 amperes. A key feature of the LAS series of Power Hybrid Voltage Regulators is its construction. A high degree of thermal isolation between the heat generating power elements and the heat sensitive control and reference elements is achieved by the placing of the power section on the heat-dissipating base of the unit, and the control stage on the heat-dissipating upper surface. This thermal isolation results in extremely low thermal drift characteristics for changes in power levels.



OUTLINE DRAWING



OUTLINE DRAWING, POWER HYBRID VOLTAGE REGULATOR, LAS 7000 SERIES

ELECTRICAL CHARACTERISTICS

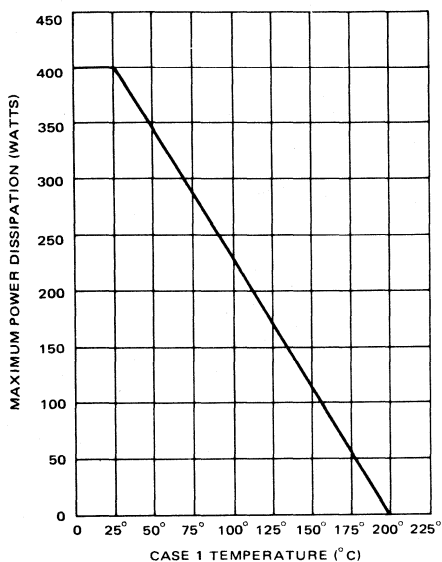
PARAMETER	SYMBOL	CONDITION	MIN.	MAX.	UNITS
Input voltage to pin (1) ^(A)	$V_{IN}(1)$		7.25	40.0	volts
Input voltage to pin (20) ^(A)	$V_{IN}(20)$		12.3	40.0	volts
Output voltage	V_O		4.75	29.4	volts
Input-output differential ^(A)	$(V_{IN}(1) - V_O)$		2.5	28.6	volts
Input-output differential ^(B)	$(V_{IN}(20) - V_O)$		7.60	28.6	volts
Output current	I_O			30.0	amps
Standby current	$I_O(1)$			40.0	mA
Standby current	$I_O(20)$			7.0	mA
Power dissipation	P_D	Plate #1 @ 25°C		400	watts
Thermal resistance junction — Case #1	θ_{J-C1}			0.44	°C/watt
Storage temperature	T_S		-55	+125	°C
Power transistor junction temperature	T_J			+200	°C
Regulation line ^(C)				0.016	%/ ΔV_{IN}
Regulation load ^(D)				0.2	%
Programming resistance				1000 nominal	ohms/volt
Programming voltage				one/one	volt/volt
Temperature coefficient				0.015	%/°C
Ripple attenuation ^(E)		$V_{IN}(1)$ minimum I_O maximum	.60		dB

NOTES:

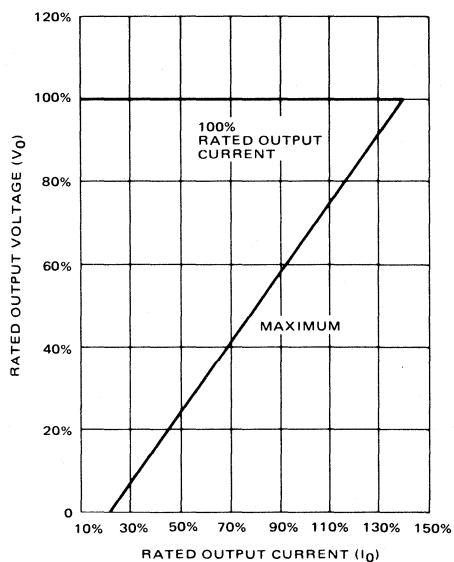
- ^(A) Separate DC input voltages for power circuit (pin 1) and control circuit pin (20).
- ^(B) Common input voltages for power circuit (pin 1) and control pin (20).
- ^(C) I_O constant for entire input voltage range from $[V_{IN}(1) \& V_{IN}(20) \text{ min.}]$ to $[V_{IN}(1) (20) \text{ max.}]$
- ^(D) V_{IN} constant for entire range from 0 to full load.
- ^(E) Ripple attenuation is 54 dB min. for 24V and 28V models.
- ^(F) Minimum input-output differential based on $T_J > 25^\circ\text{C}$.
- ^(G) For AC source to Pin 20 with source resistance less than 10 ohms, minimum VAC = 12V rms. For other conditions consult factory.
- ^(H) Maximum input voltage is 30V for LAS 7205 and 7206.

LAS 7000 SERIES

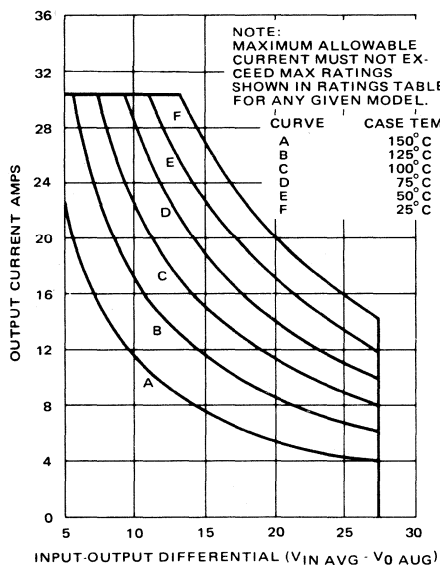
OPERATIONAL DATA



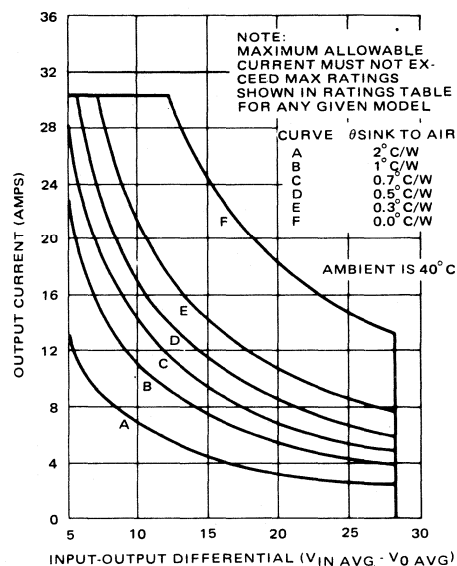
POWER DERATING CURVE AS A FUNCTION OF CASE 1 TEMPERATURE



SHORT CIRCUIT PROTECTION CHARACTERISTIC LAS 7000 SERIES



DC SAFE OPERATING AREA AS A FUNCTION OF MODULE CASE TEMPERATURE FOR LAS 7000 SERIES



DC SAFE OPERATING AREA AS A FUNCTION OF HEATSINK THERMAL RESISTANCE TO AIR AT 40° AMBIENT TEMPERATURE FOR LAS 7000 SERIES

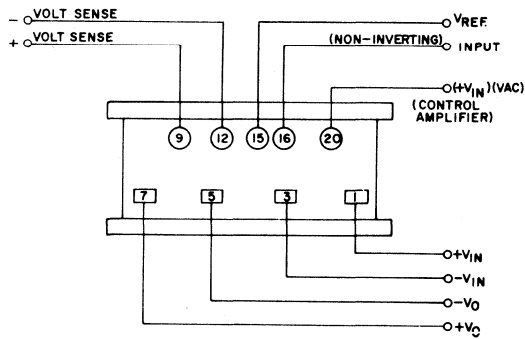
ORDERING INFORMATION

LAS 7200 30 AMPS, 400 WATTS

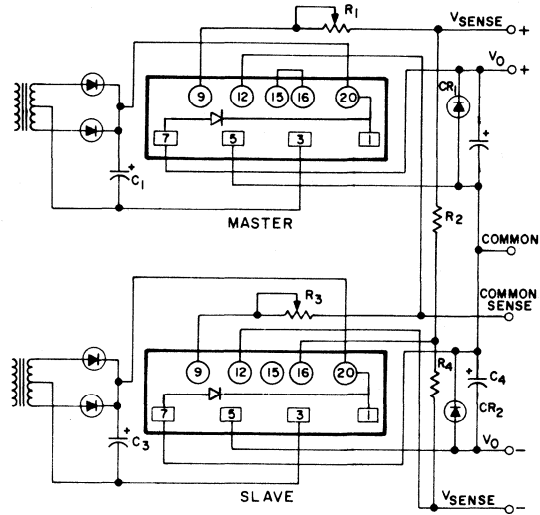
MODEL	V _o —ADJ VOLTS	PRICE QTY						
		1-24	25-49	50-99	100-249	250-499	500-999	1000-2499
LAS 7205	5	\$161.00	\$132.00	\$112.00	\$95.00	\$95.00	\$90.00	\$78.00
LAS 7215	15	161.00	132.00	112.00	95.00	95.00	90.00	78.00
LAS 7224	24	161.00	132.00	112.00	95.00	95.00	90.00	78.00

LAS 7000 SERIES

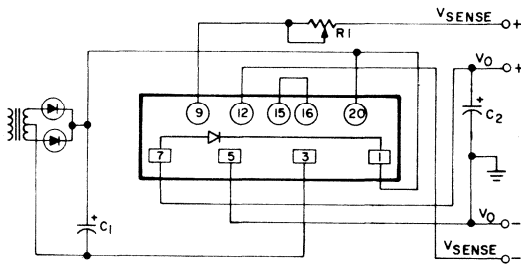
CONNECTION DIAGRAMS



9-PIN POWER HYBRID VOLTAGE REGULATOR



DUAL TRACKING POWER HYBRID VOLTAGE REGULATOR

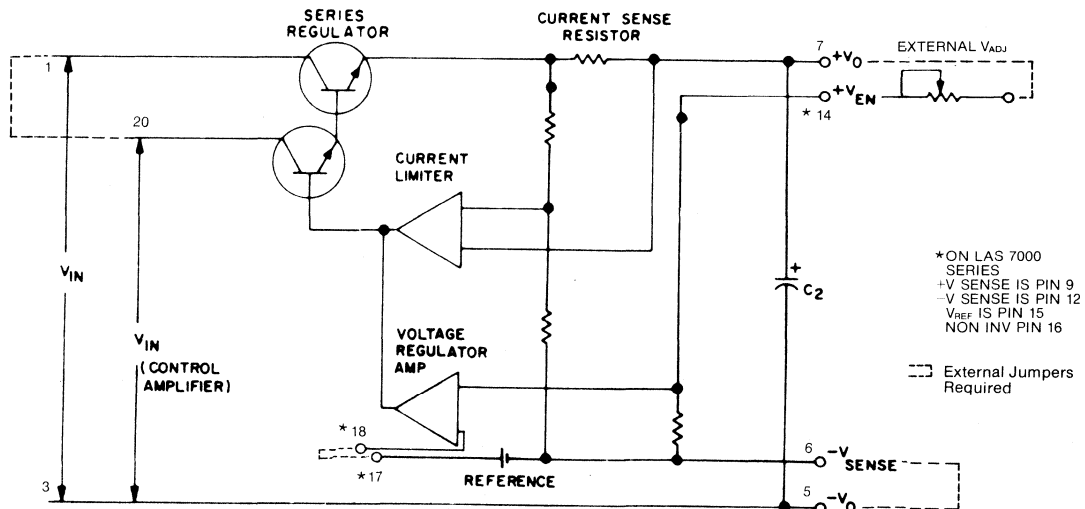


POSITIVE POWER HYBRID VOLTAGE REGULATOR CIRCUIT

NOTES:

1. Minimum value of input filter capacitors C1 and C3 is determined by: $C1, C3 = I_O \times (2000 \text{ mfd/amp})$ recommended.
2. Minimum value of output capacitors C2 and C4 is determined by: $C2, C4 = I_O \times (100 \text{ mfd/amp})$.
4. Minimum value of output voltage adjust resistors R1 and R3 is determined by: $R1, R3 = (0.25V_O \times 1000\Omega/V)$ ohms wirewound. Use next highest standard value.
5. Values of tracking reference voltage divider resistors R2 and R4 for all models are determined by:
 - a) $R2 = (2000V_O - 2490)$ ohms, $\pm 1\%$, $\frac{1}{2}W$ film
 - b) $R4 = 2.49K$ ohms, $\pm 1\%$, $\frac{1}{2}W$ film
6. Rectifiers CR1 and CR2 should be rated at peak inverse voltage of 50V and forward current equal at least to maximum rated I_O .

FUNCTIONAL BLOCK DIAGRAM



LAS 3000, LAS 5000 AND LAS 7000
SEE LAS 2000 SERIES FOR LAS 3000 PIN CORRECTION.

SECTION II – POWER DARLINGTONS

FEATURES

- The only 200° C operating junction temperature Darlington power transistors available in the industry.
- The only hard glass passivated 200° C Darlington power transistors available in the industry.
- Guaranteed leakage current stability at 200° C — I_{CER} is equal to 5mA or less, up to 150 watts; I_{CER} is equal to 7mA or less, at 225 watts; I_{CER} is equal to 10mA or less, at 240 watts.
- All Darlington power transistors are hermetically sealed and subject to leak testing.
- Guaranteed DC current gain greater than or equal to 1000 for NPN and 800 for PNP at collector currents of 4, 6, 10, and 15 amps.
- All Darlington power transistors are guaranteed and 100% tested for $I_{s,b}$ (Secondary breakdown Current) insuring maximum performance at high energy levels.
- Excellent thermal resistance ($R_{\theta JC}$) on all Darlington power transistors providing for more usable power and lower operating temperatures.

DESCRIPTION

The PMD-10K, 12K, 16K, 1600K, and 18K series of devices are three-terminal NPN Darlington Power Transistors. The PMD-11K, 13K, 17K, 1700K, and 19K series of devices are three-terminal PNP Darlington Power Transistors. These devices are monolithic epitaxial base structures with built-in base to emitter shunt resistors. The devices are CVD glass passivated to increase reliability and provide reduced high temperature reverse leakage current. This important feature helps to enable this series of Darlington devices to meet guaranteed operating junction temperatures of 200°C. Internal diode protection (D1) of the Darlington configura-

tion is built into the structure to limit the device power dissipation during negative overshoot.

The five different series of NPN and five different series of PNP Darlington Power Transistors are available in sustaining voltages ranging from +40 to +100 volts, and power dissipation levels from 100 to 240 watts. All Darlington devices are hermetically sealed steel or copper TO-3 packages, depending on power dissipation requirements, providing high reliability and low thermal resistance, when used with appropriate heat sinks.

ABSOLUTE MAXIMUM RATINGS

RATING	SYMBOL	12K 40 13K 40	12K 60 13K 60	12K 80 13K 80	12K 100 13K 100	10K 40 11K 40	10K 60 11K 60	10K 80 11K 80	10K 100 11K 100	1601K 1701K	1602K 1702K	1603K 1703K	UNITS
Collector Emitter Voltage	V_{CEO}	40	60	80	100	40	60	80	100	60	80	100	Vdc
Collector Base Voltage	V_{CBO}	40	60	80	100	40	60	80	100	60	80	100	Vdc
Emitter Base Voltage	V_{EBO}	5	5	5	5	5	5	5	5	5	5	5	Vdc
Collector Current Cont. Peak	I_C I_C	8 16	8 16	8 16	8 16	12 20	12 20	12 20	12 20	20 40	20 40	20 40	Adc
Base Current	I_B	0.12	0.12	0.12	0.12	0.2	0.2	0.2	0.2	0.5	0.5	0.5	Adc
Thermal Resistance	$R_{\theta JC}$	1.5	1.5	1.5	1.5	1.0	1.0	1.0	1.0	.83	.83	.83	°C/W
Total Internal Dissipation	P_D @ $T_C = 50^\circ C$ Derate	100	100	100	100	150	150	150	150	180	180	180	Watts
		Derate at 0.667 W/°C, $T_A \geq 50^\circ C$				Derate at 1 W/°C, $T_A \geq 50^\circ C$				Derate at 1.2 W/°C $T_A \geq 50^\circ C$			
Temp. Range Operating Storage	T_J T_{STG}	-65° to +200°C				-65° to +200°C				-65° to +200°C			°C

RATING	SYMBOL	16K 60 17K 60	16K 80 17K 80	16K 100 17K 100	18K 60 19K 60	18K 80 19K 80	18K 100 19K 100	UNITS
Collector Emitter Voltage	V_{CEO}	60	80	100	60	80	100	Vdc
Collector Base Voltage	V_{CBO}	60	80	100	60	80	100	Vdc
Emitter Base Voltage	V_{EBO}	5	5	5	5	5	5	Vdc
Collector Current Cont. Peak	I_C I_C	20 40	20 40	20 40	30 60	30 60	30 60	Adc
Base Current	I_B	0.5	0.5	0.5	.75	.75	.75	Adc
Thermal Resistance	$R_{\theta JC}$	0.67	0.67	0.67	.625	.625	.625	°C/W
Total Internal Power Dissipation	P_D @ $T_C = 50^\circ C$ Derate	225	225	225	240	240	240	Watts
		Derate at 1.5 W/°C $T_A \geq 50^\circ C$			Derate at 1.6 W/°C $T_A \geq 50^\circ C$			
Temp. Range Operating Storage	T_J T_{STG}	-65° to +200°C			-65° to +200°C			

ORDERING INFORMATION

Type	Voltage Rating	Rated Power (watts)	Price—Quantity				Type	Voltage Rating	Rated Power (watts)	Price—Quantity			
			1-99	100-999	1000-2499	2500-4999				1-99	100-999	1000-2499	2500-4999
PMD 10K 40	40V	150	\$1.90	\$1.55	\$1.39	\$1.33	PMD 11K 40	40V	150	\$1.90	\$1.55	\$1.39	\$1.33
PMD 10K 60	60V	150	2.10	1.68	1.50	1.44	PMD 11K 60	60V	150	2.10	1.68	1.50	1.44
PMD 10K 80	80V	150	2.32	1.86	1.67	1.60	PMD 11K 80	80V	150	2.32	1.86	1.67	1.60
PMD 10K 100	100V	150	2.70	2.16	1.94	1.86	PMD 11K 100	100V	150	2.70	2.16	1.94	1.86
PMD 12K 40	40V	100	1.10	.88	.81	.78	PMD 13K 40	40V	100	1.10	.88	.81	.78
PMD 12K 60	60V	100	1.14	.91	.84	.81	PMD 13K 60	60V	100	1.14	.91	.84	.81
PMD 12K 80	80V	100	1.20	.95	.87	.85	PMD 13K 80	80V	100	1.20	.95	.87	.85
PMD 12K 100	100V	100	1.25	1.00	.92	.89	PMD 13K 100	100V	100	1.25	1.00	.92	.89
PMD 1601 K	60V	180	2.95	2.35	2.10	2.02	PMD 1701 K	60V	180	2.95	2.35	2.10	2.02
PMD 1602 K	80V	180	3.35	2.67	2.40	2.30	PMD 1702 K	80V	180	3.35	2.67	2.40	2.30
PMD 1603 K	100V	180	3.90	3.10	2.78	2.66	PMD 1703 K	100V	180	3.90	3.10	2.78	2.66
PMD 16K 60	60V	225	4.68	3.78	3.42	3.33	PMD 17K 60	60V	225	5.20	4.20	3.80	3.70
PMD 16K 80	80V	225	5.05	4.15	3.50	3.42	PMD 17K 80	80V	225	5.60	4.60	3.90	3.80
PMD 16K 100	100V	225	5.60	4.60	3.90	3.80	PMD 17K 100	100V	225	6.16	5.06	4.30	4.18
PMD 18K 60	60V	240	5.38	4.35	3.93	3.83	PMD 19K 60	60V	240	5.92	4.78	4.33	4.21
PMD 18K 80	80V	240	5.80	4.77	4.03	3.93	PMD 19K 80	80V	240	6.38	5.25	4.43	4.37
PMD 18K 100	100V	240	6.44	5.29	4.49	4.37	PMD 19K 100	100V	240	7.08	5.82	4.93	4.81

Contact your area sales representative for higher quantity prices. Device configurations, specifications, and prices subject to change without notice.

CROSS REFERENCE GUIDE FOR IMPROVED EQUIVALENTS

MOTOROLA PART NUMBER	LAMBDA EQUIVALENT	LAMBDA IMPROVED REPLACEMENT	FAIRCHILD PART NUMBER	LAMBDA EQUIVALENT	LAMBDA IMPROVED REPLACEMENT	RCA PART NUMBER	LAMBDA EQUIVALENT	LAMBDA IMPROVED REPLACEMENT
MJ 3000	LMJ 3000	PMD 10K 60	2N 6050	2N 6050	PMD 11K 60	2N 6282	2N 6282	PMD 16K 60
MJ 4033	LMJ 4033	PMD 10K 60	SE 9403	LSE 9403	PMD 11K 60	2N 6283	2N 6283	PMS 16K 80
MJ 3001	LMJ 3001	PMD 10K 80	SE 9404	LSE 9404	PMD 11K 80	2N 6284	2N 6284	PMD 16K 100
MJ 4034	LMJ 4034	PMD 10K 80	SE 9405	LSE 9405	PMD 11K 100			
MJ 4035	LMJ 4035	PMD 10K 100				2N 6285	2N 6285	PMD 1701K
			SE 9303	LSE 9303	PMD 12K 60	2N 6286	2N 6286	PMD 1702K
			SE 9304	LSE 9304	PMD 12K 80	2N 6287	2N 6287	PMD 1703K
			SE 9305	LSE 9305	PMD 12K 100			
MJ 2500	LMJ 2500	PMD 11K 60				2N 6285	2N 6285	PMD 17K 60
MJ 4030	LMJ 4030	PMD 11K 60				2N 6286	2N 6286	PMD 17K 80
MJ 2501	LMJ 2501	PMD 11K 80	SE 9403	LSE 9403	PMD 13K 60	2N 6287	2N 6287	PMD 17K 100
MJ 4031	LMJ 4031	PMD 11K 80	SE 9404	LSE 9404	PMD 13K 80			
MJ 4032	LMJ 4032	PMD 11K 100	SE 9405	LSE 9405	PMD 13K 100			
MJ 1000	LMJ 1000	PMD 12K 60	2N 6282	2N 6282	PMD 1601K	TEXAS INSTRUMENTS PART NUMBER LAMBDA EQUIVALENT LAMBDA IMPROVED REPLACEMENT		
MJ 1001	LMJ 1001	PMD 12K 80	2N 6283	2N 6283	PMD 1602K			
MJ 900	LMJ 900	PMD 13K 60	2N 6284	2N 6284	PMD 1603K	TIP 640	LTIP 640	PMD 10K 60
MJ 901	LMJ 901	PMD 13K 80	2N 6282	2N 6282	PMD 16K 60	TIP 641	LTIP 641	PMD 10K 80
			2N 6283	2N 6283	PMD 16K 80	TIP 641	LTIP 642	PMD 10K 100
			2N 6284	2N 6284	PMD 16K 100			
2N 6282	2N 6282	PMD 1601K	2N 6285	2N 6285	PMD 1701K	TIP 645	LTIP 645	PMD 11K 60
2N 6283	2N 6283	PMD 1602K	2N 6286	2N 6286	PMD 1702K	TIP 646	LTIP 646	PMD 11K 80
2N 6284	2N 6284	PMD 1603K	2N 6287	2N 6287	PMD 1703K	TIP 647	LTIP 647	PMD 11K 100
2N 6282	2N 6282	PMD 16K 60						
2N 6283	2N 6283	PMD 16K 80	2N 6285	2N 6285	PMD 17K 60	STANDARD COMMERCIAL PART NUMBER LAMBDA EQUIVALENT LAMBDA IMPROVED REPLACEMENT		
2N 6284	2N 6284	PMD 16K 100	2N 6286	2N 6286	PMD 17K 80			
			2N 6287	2N 6287	PMD 17K 100	2N 6057	2N 6057	PMD 10K 60
2N 6285	2N 6285	PMD 1701K	RCA PART NUMBER LAMBDA EQUIVALENT LAMBDA IMPROVED REPLACEMENT			2N 6058	2N 6058	PMD 10K 80
2N 6286	2N 6286	PMD 1702K	RCA 8350	LCA 8350	PMD 11K 40	2N 6059	2N 6059	PMD 10K 100
2N 6287	2N 6287	PMD 1703K	RCA 8350A	LCA 8350A	PMD 11K 60			
			RCA 8350B	LCA 8350B	PMD 11K 80	2N 6050	2N 6050	PMD 11K 60
						2N 6051	2N 6051	PMD 11K 80
						2N 6052	2N 6052	PMD 11K 100
			RCA 1000	LCA 1000	PMD 12K 60			
			RCA 1001	LCA 1001	PMD 12K 80	2N 6055	2N 6055	PMD 12K 60
						2N 6056	2N 6056	PMD 12K 80
			RCA 8350	LCA 8350	PMD 13K 40			
			RCA 8350A	LCA 8350A	PMD 13K 60	2N 6053	2N 6053	PMD 13K 60
			RCA 8350B	LCA 8350B	PMD 13K 80	2N 6054	2N 6054	PMD 13K 80
2N 6057	2N 6057	PMD 10K 60	2N 6282	2N 6282	PMD 1601K	2N 6383	2N 6383	PMD 12K 40
SE 9303	LSE 9303	PMD 10K 60	2N 6283	2N 6283	PMD 1602K	2N 6384	2N 6384	PMD 12K 60
SE 9304	LSE 9304	PMD 10K 80	2N 6284	2N 6284	PMD 1603K	2N 6385	2N 6385	PMD 12K 80
SE 9305	LSE 9305	PMD 10K 100						

△ LAMBDA

POWER DARLINGTON NPN SERIES—

PMD 10K, 12K, 16K, 1600K, 18K

ELECTRICAL CHARACTERISTICS

ALL PARAMETERS ARE GUARANTEED AT T_J of 0° C to 200° C, UNLESS OTHERWISE SPECIFIED

DEVICE TYPE	TEST	SYMBOL	TEST CONDITIONS	MIN.	MAX.	UNITS	NOTE
PMD 10K SERIES PMD 12K SERIES PMD 16K SERIES PMD 1600K SERIES PMD 18K SERIES	Collector Emitter Saturation Voltage	$V_{CE} (Sat)$	$I_C = 6A_{dc}; I_B = 24mA_{dc}$ $I_C = 4A_{dc}; I_B = 16mA_{dc}$ $I_C = 10A_{dc}; I_B = 40mA_{dc}$ $I_C = 10A_{dc}; I_B = 40mA_{dc}$ $I_C = 5A_{dc}; I_B = 60mA_{dc}$		2.0	Vdc	1
PMD 10K SERIES PMD 12K SERIES PMD 16K SERIES PMD 1600K SERIES PMD 18K SERIES	Base Emitter Turn-on Voltage	$V_{BE} (on)$	$I_C = 6A_{dc}$ $I_C = 4A_{dc}$ $I_C = 10A_{dc}$ $I_C = 10A_{dc}$ $I_C = 15A_{dc}$	$V_{CE} = 3V_{dc}$	2.8	Vdc	1
PMD 10K SERIES PMD 12K SERIES PMD 16K SERIES PMD 1600K SERIES PMD 18K SERIES	Base Emitter Saturation Voltage	$V_{BE} (Sat)$	$I_C = 6A_{dc}; I_B = 24mA_{dc}$ $I_C = 4A_{dc}; I_B = 16mA_{dc}$ $I_C = 10A_{dc}; I_B = 40mA_{dc}$ $I_C = 10A_{dc}; I_B = 40mA_{dc}$ $I_C = 15A_{dc}; I_B = 60mA_{dc}$		2.8	Vdc	1
PMD 10K SERIES PMD 12K SERIES PMD 16K SERIES PMD 1600K SERIES PMD 18K SERIES	DC Current Gain	h_{FE}	$I_C = 6A_{dc}$ $I_C = 4A_{dc}$ $I_C = 10A_{dc}$ $I_C = 10A_{dc}$ $I_C = 15A_{dc}$	$V_{CE} = 3V_{dc}$ $* T_J = 25^\circ C$	1000 1000 1000 750 1000	20,000*	1
PMD 10K SERIES PMD 12K SERIES PMD 16K SERIES PMD 1600K SERIES PMD 18K SERIES	Forward Bias Second Breakdown current	$I_{s/b}$	$T_A = 25^\circ C$ $V_{CE} = 30V_{dc}$ 1 second nonrepetitive pulse	5.0 3.0 7.5 6.0 8.0		A _{dc}	
PMD 10K, 12K-40 PMD 10K, 12K 16K, 18K-60 PMD 1601K PMD 10K, 12K, 16K, 18K-80 PMD 1602 PMD 10K, 12K 16K, 18K-100 PMD 1603K	Collector Emitter Breakdown Voltage (Base Open)	$V_{(BR) CEO}$	$I_{CE} = 100mA_{dc}$ $T_J = 25^\circ C$	40 60 80 100		Vdc	1
PMD 10K, 12K-40 PMD 10K, 12K, 16K, 18K-60 PMD 1601K PMD 10K, 12K 16K, 18K-80 PMD 1602K PMD 10K, 12K, 16K, 18K-100 PMD 1603K	Collector Emitter Sustaining Voltage	$V_{(BR) CEO}$	$I_{CE} = 100mA_{dc}$ $R_{BE} = 2.2K\ ohms$	40 60 80 100		Vdc	1
All NPN Series	Emitter Base Leakage Current	I_{EBO}	$V_{EB} = 5V_{dc}$ $I_C = 0A$		3.0	mA _{dc}	1
PMD 10K, 12K-40 PMD 10K, 12K-60 PMD 10K, 12K-80 PMD 10K, 12K-100	Collector Emitter Leakage Current	I_{CER}	$V_{CE} = 20V_{dc}$ $V_{CE} = 40V_{dc}$ $V_{CE} = 54V_{dc}$ $V_{CE} = 67V_{dc}$	$R_{BE} =$ 2.2Kohms	5.0 5.0 5.0 5.0	mA _{dc}	1

continued on next page

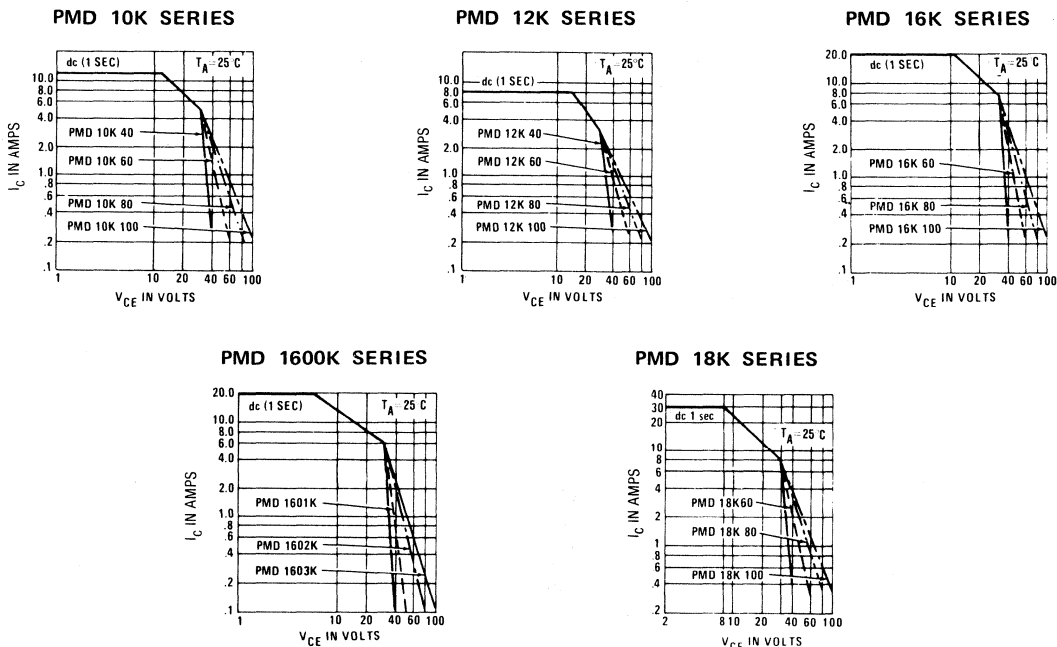
ELECTRICAL CHARACTERISTICS (continued)

DEVICE TYPE	TEST	SYMBOL	TEST CONDITIONS	MIN.	MAX.	UNITS	NOTE
PMD 16K-60 PMD 1601K PMD 16K-80 PMD 1602K PMD 16K-100 PMD 1603K PMD 18K-60 PMD 18K-80 PMD 18K-100	Collector Emitter Leakage Current	I_{CER}	OFF Characteristics $V_{CE} = 40Vdc$ $V_{CE} = 54Vdc$ $V_{CE} = 67Vdc$ $V_{CE} = 40Vdc$ $V_{CE} = 54Vdc$ $V_{CE} = 67Vdc$ $R_{BE} = 2.2Kohms$		7.0 10.0	mAdc	1
PMD 10K SERIES PMD 12K SERIES PMD 16K SERIES PMD 1600K SERIES PMD 18K SERIES	Output Capacitance	Cobo	Dynamic Characteristics $V_{CB} = 10Vdc; I_E = 0Adc$ $f = 1MHz$ $T_J = 25^\circ C$		300 200 400 400 600	pf	
PMD 10K SERIES PMD 12K SERIES PMD 16K SERIES PMD 1600K SERIES PMD 18K SERIES	Small Signal Current Gain	h_{fe}	Dynamic Characteristics $I_C = 5Adc$ $I_C = 3Adc$ $I_C = 7Adc$ $I_C = 7Adc$ $I_C = 9Adc$ $V_{CE} = 3Vdc$ $f = 1kHz$ $T_J = 25^\circ C$	300			1
PMD 10K SERIES PMD 12K SERIES PMD 16K SERIES PMD 1600K SERIES PMD 18K SERIES	Common Emitter Short-Circuit Forward Transfer Ratio	h_{fe}	Dynamic Characteristics $I_C = 5Adc$ $I_C = 3Adc$ $I_C = 7Adc$ $I_C = 7Adc$ $I_C = 9Adc$ $V_{CE} = 3Vdc$ $f = 1MHz$ $T_J = 25^\circ C$	4			1

NOTES:

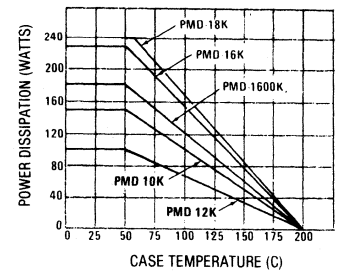
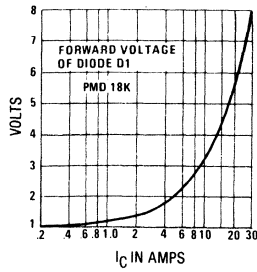
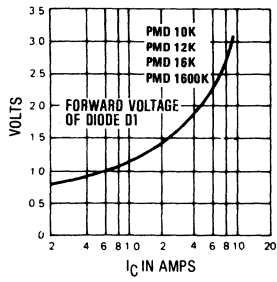
1. Pulse Tested: Pulse Width is less than or equal to $300\mu Sec.$ and the Duty Cycle is less than or equal to 2.0%.
DEVICE PRICES, SPECIFICATIONS, AND CONFIGURATIONS ARE SUBJECT TO CHANGE WITHOUT NOTICE.

OPERATIONAL DATA Safe Operating Area

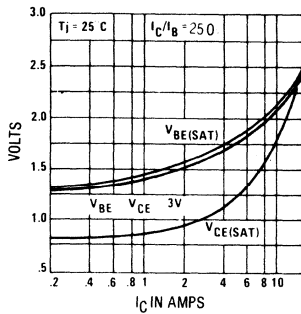


OPERATIONAL DATA (continued)

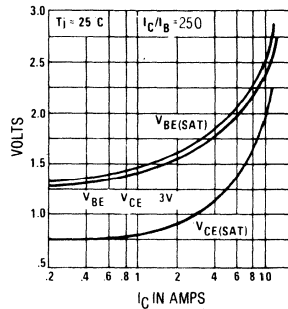
POWER DERATING



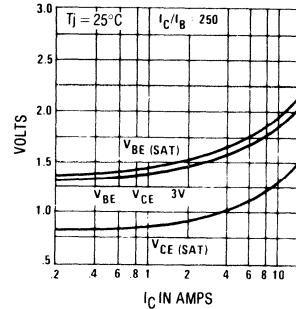
"ON" VOLTAGES VERSUS COLLECTOR CURRENT



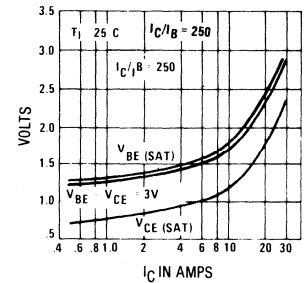
PMD 10K SERIES



PMD 12K SERIES

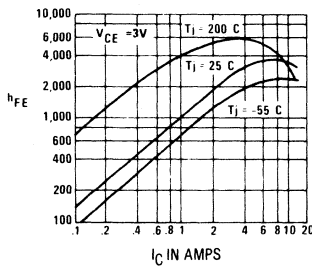


PMD 16K, 1600K SERIES

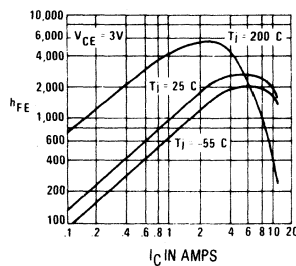


PMD 18K SERIES

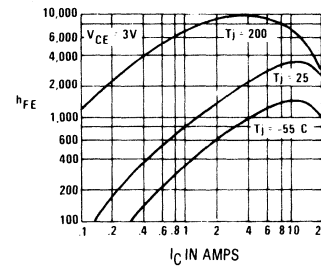
DC COLLECTOR CURRENT GAIN VERSUS COLLECTOR CURRENT



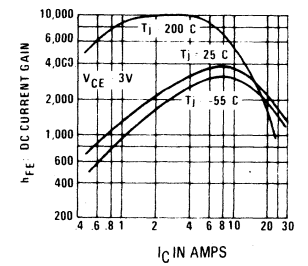
PMD 10K SERIES



PMD 12K SERIES

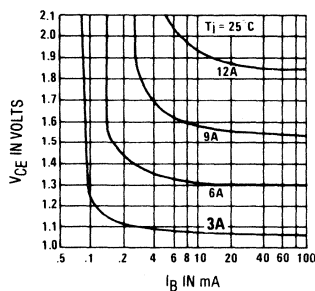


PMD 16K, 1600K SERIES

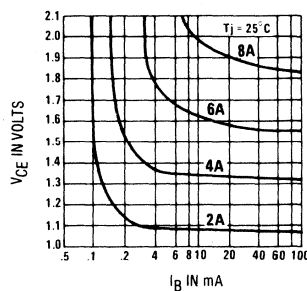


PMD 18K SERIES

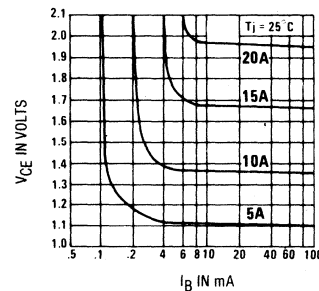
COLLECTOR SATURATION REGION



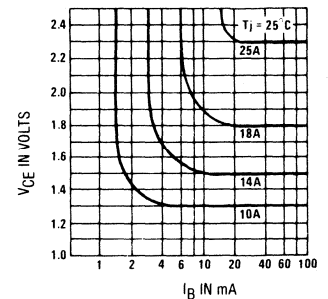
PMD 10K SERIES



PMD 12K SERIES



PMD 16K, 1600K SERIES



PMD 18K SERIES

LAMBDA **POWER DARLINGTON PNP SERIES—** **PMD 11K, 13K, 17K, 1700K, 19K**

ELECTRICAL CHARACTERISTICS

ALL PARAMETERS ARE GUARANTEED AT T_J of 0° C to 200° C, UNLESS OTHERWISE SPECIFIED

DEVICE TYPE	TEST	SYMBOL	TEST CONDITIONS	MIN.	MAX.	UNITS	NOTE
PMD 11K SERIES PMD 13K SERIES PMD 17K SERIES PMD 1700K SERIES PMD 19K SERIES	Collector Emitter Saturation Voltage	$V_{CE} (Sat)$	$I_C = 6A_{dc}; I_B = 24mA_{dc}$ $I_C = 4A_{dc}; I_B = 16mA_{dc}$ $I_C = 10A_{dc}; I_B = 40mA_{dc}$ $I_C = 10A_{dc}; I_B = 40mA_{dc}$ $I_C = 15A_{dc}; I_B = 60mA_{dc}$		2.0	Vdc	1
PMD 11K SERIES PMD 13K SERIES PMD 17K SERIES PMD 1700K SERIES PMD 19K SERIES	Base Emitter Turn-on Voltage	$V_{BE} (on)$	$I_C = 6A_{dc}$ $I_C = 4A_{dc}$ $I_C = 10A_{dc}$ $I_C = 10A_{dc}$ $I_C = 15A_{dc}$	$V_{CE} = 3V_{dc}$	2.8	Vdc	1
PMD 11K SERIES PMD 13K SERIES PMD 17K SERIES PMD 1700K SERIES PMD 19K SERIES	Base Emitter Saturation	$V_{BE} (Sat)$	$I_C = 6A_{dc}; I_B = 24mA_{dc}$ $I_C = 4A_{dc}; I_B = 16mA_{dc}$ $I_C = 10A_{dc}; I_B = 40mA_{dc}$ $I_C = 10A_{dc}; I_B = 40mA_{dc}$ $I_C = 15A_{dc}; I_B = 60mA_{dc}$		2.8	Vdc	1
PMD 11K SERIES PMD 13K SERIES PMD 17K SERIES PMD 1700K SERIES PMD 19K SERIES	DC Current Gain	h_{FE}	$I_C = 6A_{dc}$ $I_C = 4A_{dc}$ $I_C = 10A_{dc}$ $I_C = 10A_{dc}$ $I_C = 15A_{dc}$	$V_{CE} = 3V_{dc}$ $* T_J = 25^\circ C$	800 800 800 750 800	20,000*	1
PMD 11K SERIES PMD 13K SERIES PMD 17K SERIES PMD 1700K SERIES PMD 19K SERIES	Forward Bias Second Breakdown current	$I_{S/B}$	$T_A = 25^\circ C$ $V_{CE} = 30V_{dc}$ 1 second nonrepetitive pulse	5.0 3.0 7.5 6.0 8.0		A _{dc}	
PMD 11K, 13K-40 PMD 11K, 13K, 17K, 19K-60 PMD 1701K PMD 11K, 13K, 17K, 19K-80 PMD 1702 PMD 11K, 13K, 17K, 19K-100 PMD 1703K	Collector Emitter Breakdown Voltage (Base Open)	$V_{(BR)CEO}$	$I_{CE} = 100mA_{dc}$ $T_J = 25^\circ C$	40 60 80 100		Vdc	1
PMD 11K, 13K-40 PMD 11K, 13K, 17K, 19K-60 PMD 1701K PMD 11K, 13K, 17K, 19K-80 PMD 1702K PMD 11K, 13K, 17K, 19K-100 PMD 1703K	Collector Emitter Sustaining Voltage	$V_{(BR)CER(sus)}$	$I_{CE} = 100mA_{dc}$ $R_{BE} = 2.2K\ ohms$	40 60 80 100		Vdc	1
All PNP Series	Emitter Base Leakage Current	I_{EBO}	$V_{EB} = 5V_{dc}$ $I_C = 0A$		3.0	mA _{dc}	1
PMD 11K, 13K-40 PMD 11K, 13K-60 PMD 11K, 13K-80 PMD 11K, 13K-100	Collector Emitter Leakage Current	I_{CER}	$V_{CE} = 20V_{dc}$ $V_{CE} = 40V_{dc}$ $V_{CE} = 54V_{dc}$ $V_{CE} = 67V_{dc}$	$R_{BE} = 2.2Kohms$	5.0 5.0 5.0 5.0	mA _{dc}	1

continued on next page

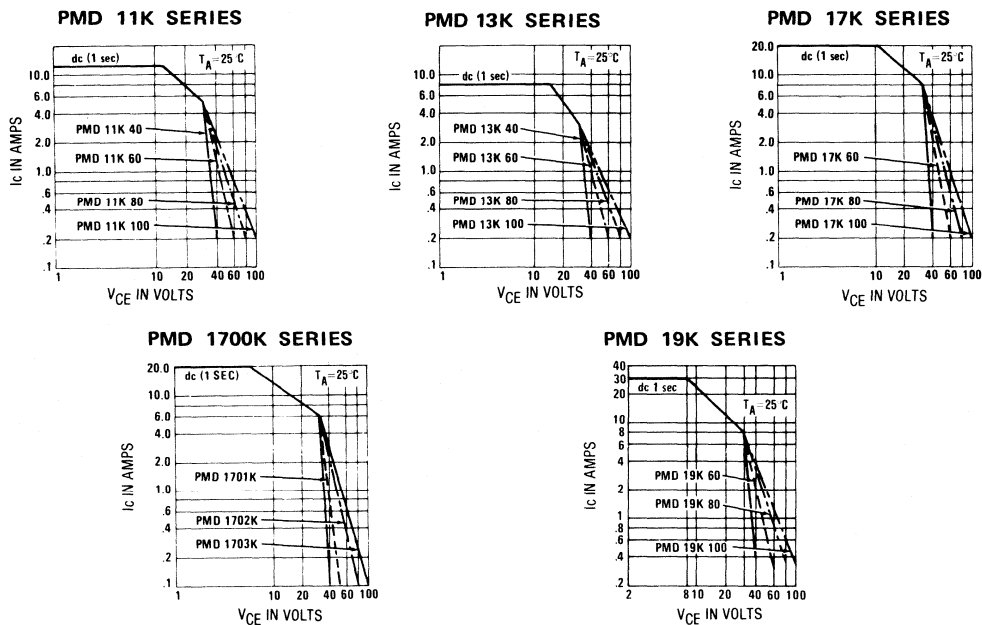
ELECTRICAL CHARACTERISTICS (continued)

DEVICE TYPE	TEST	SYMBOL	TEST CONDITIONS	MIN.	MAX.	UNITS	NOTE
PMD 17K-60 PMD 1701K PMD 17K-80 PMD 1702K PMD 17K-100 PMD 1703K PMD 19K-40 PMD 19K-60 PMD 19K-80 PMD 19K-100	Collector Emitter Leakage Current	I_{CER}	Off Characteristics $V_{CE} = 40Vdc$ $V_{CE} = 54Vdc$ $V_{CE} = 67Vdc$ $R_{BE} = 2.2Kohms$ $V_{CE} = 20Vdc$ $V_{CE} = 40Vdc$ $V_{CE} = 54Vdc$ $V_{CE} = 67Vdc$		7.0 10.0	mAdc	1
PMD 11K SERIES PMD 13K SERIES PMD 17K SERIES PMD 1700K SERIES PMD 19K SERIES	Output Capacitance	C_{obo}	Dynamic Characteristics $V_{CB} = 10Vdc; I_E = 0Adc$ $f = 1MHz$ $T_J = 25^\circ C$		300 200 400 400 600	pf	
PMD 11K SERIES PMD 13K SERIES PMD 17K SERIES PMD 1700K SERIES PMD 19K SERIES	Small Signal Current Gain	h_{fe}	Dynamic Characteristics $I_C = 5Adc$ $I_C = 3Adc$ $I_C = 7Adc$ $I_C = 7Adc$ $I_C = 9Adc$ $V_{CE} = 3Vdc$ $f = 1kHz$ $T_J = 25^\circ C$	300			1
PMD 11K SERIES PMD 13K SERIES PMD 17K SERIES PMD 1700K SERIES PMD 19K SERIES	Common Emitter Short-Circuit Forward Transfer Ratio	h_{fe}	Dynamic Characteristics $I_C = 5Adc$ $I_C = 3Adc$ $I_C = 7Adc$ $I_C = 7Adc$ $I_C = 9Adc$ $V_{CE} = 3Vdc$ $f = 1MHz$ $T_J = 25^\circ C$	4			1

NOTES:

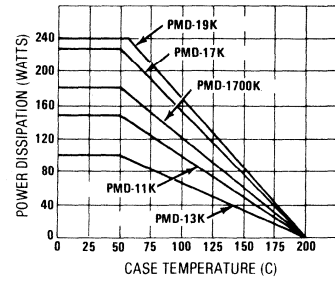
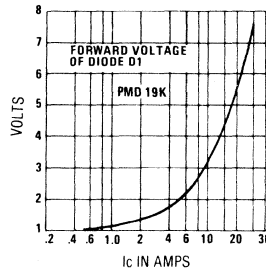
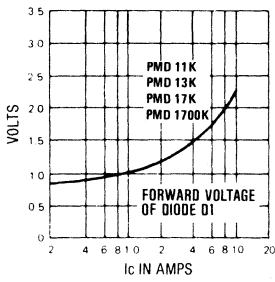
1. Pulse Tested: Pulse Width is less than or equal to $300\mu Sec.$ and the Duty Cycle is less than or equal to 2.0%.
DEVICE PRICES, SPECIFICATIONS, AND CONFIGURATIONS ARE SUBJECT TO CHANGE WITHOUT NOTICE.

OPERATIONAL DATA Safe Operating Area

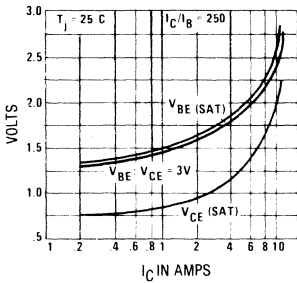


OPERATIONAL DATA (continued)

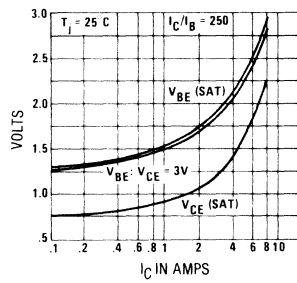
POWER DERATING



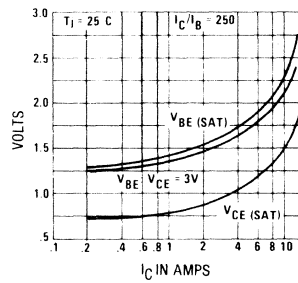
"ON" VOLTAGES VERSUS COLLECTOR CURRENT



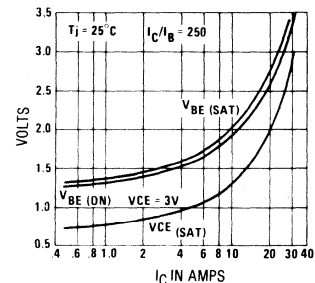
PMD 11K SERIES



PMD 13K SERIES

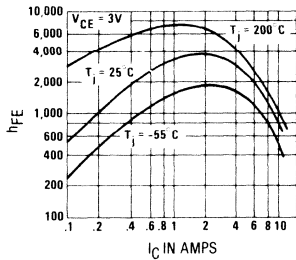


PMD 17K, 1700K SERIES

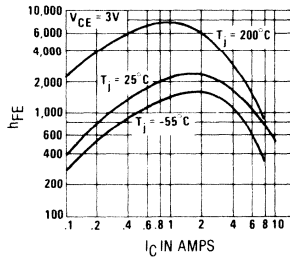


PMD 19K SERIES

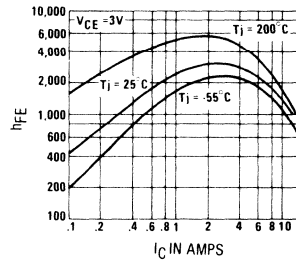
DC COLLECTOR CURRENT GAIN VERSUS COLLECTOR CURRENT



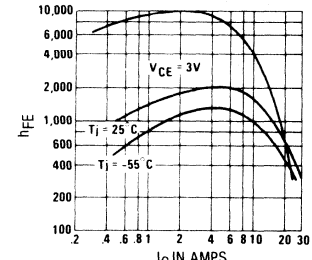
PMD 11K SERIES



PMD 13K SERIES

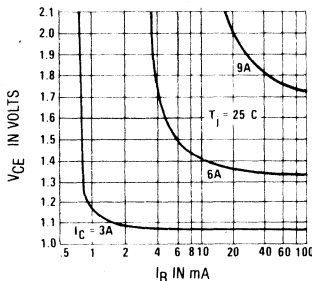


PMD 17K, 1700K SERIES

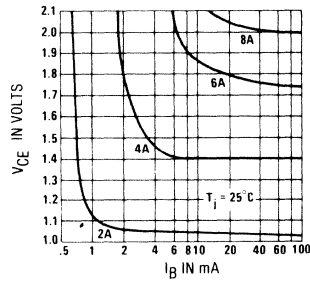


PMD 19K SERIES

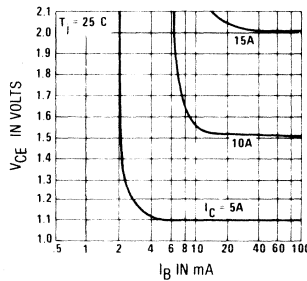
COLLECTOR SATURATION REGION



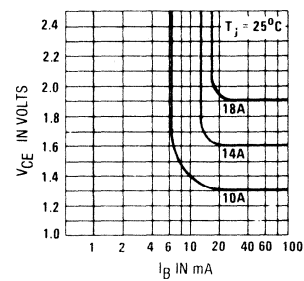
PMD 11K SERIES



PMD 13K SERIES

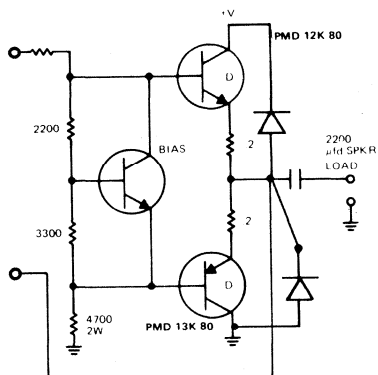


PMD 17K, 1700K SERIES

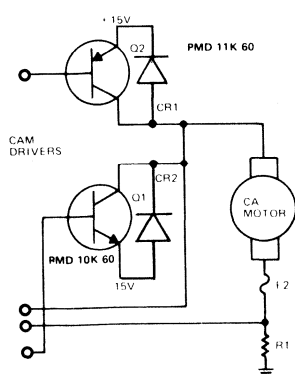


PMD 19K SERIES

TYPICAL APPLICATIONS

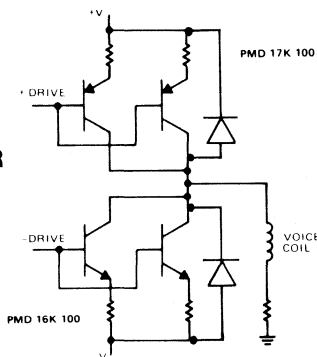


CLASS AB AMPLIFIER

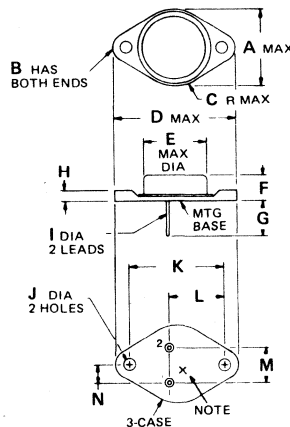


MOTOR DRIVE

DISC DRIVE FOR COMPUTERS



OUTLINE DRAWINGS



TO-3/3 TERMINAL COPPER

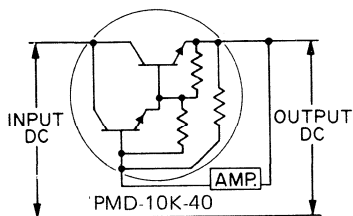
DIM	INCHES	
	MIN.	MAX.
A	—	1.050
B	—	0.188
C	—	0.525
D	—	1.575
E	—	0.875
F	0.250	0.450
G	0.400	0.480
H	0.060	0.115
I	0.035	0.045
J	0.150	0.160
K	1.175	1.195
L	0.655	0.675
M	0.420	0.440
N	0.210	0.220

TO-3/3 TERMINAL STEEL

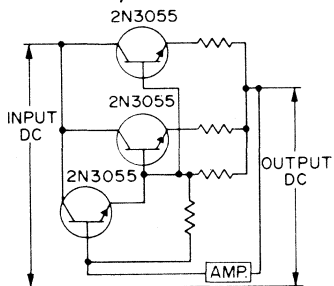
DIM.	INCHES	
	MIN.	MAX.
A	—	1.050
B	—	0.188
C	—	0.525
D	—	1.575
E	—	0.875
F	0.250	0.450
G	0.400	0.480
H	0.050	0.070
I	0.035	0.045
J	0.150	0.160
K	1.175	1.195
L	0.655	0.675
M	0.420	0.440
N	0.210	0.220

TYPICAL COST SAVINGS WITH LAMBDA DARLINGTONS

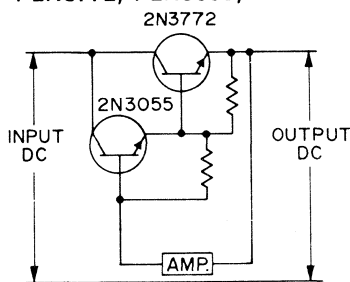
1PMD-10-K-40 Darlington



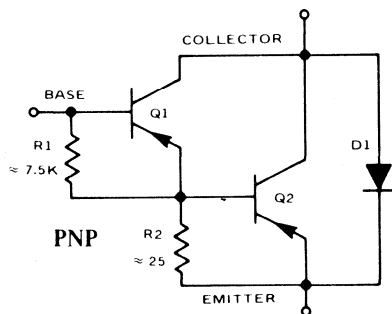
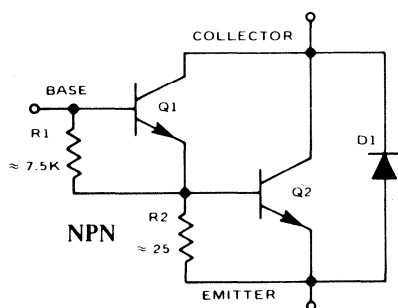
3-2N3055, 4 Resistors



1-2N3772, 1-2N3055, 2 Resistors



BLOCK DIAGRAMS



PIN	FUNCTION	NOTE (X): CASE TEMPERATURE MEASURED AT THIS POINT
1	BASE	
2	EMITTER	
CASE	COLLECTOR	

SECTION III—OVERVOLTAGE PROTECTORS

FEATURES

• True temperature compensated overvoltage protection—2A, 6A, 12A, 20A, 35A

• Complete overvoltage protectors in single package

DESCRIPTION

The L-2-OV and the L-6-OV series of devices are monolithic overvoltage protectors. These two-terminal devices are available in 2 amp and 6 amp versions. The L-12-OV, L-20-OV and the L-35-OV series of devices are hybrid overvoltage protectors. These devices are available in 12 amp, 20 amp and 35 amp versions. They have been designed to prevent damage to a load caused by excessive power supply output voltage, improper connection, a disconnected sense lead, or failure of the power supply.

Load protection is accomplished by effectively short circuiting the output terminals of the power supply when the trip point voltage limit of the overvoltage protector is exceeded. All overvoltage protectors in each series have fixed trip point voltage values and

cannot be adjusted. To reset the overvoltage protector, interrupt power to the overvoltage protector, allow the overvoltage protector to cool below 71°C, and reconnect the power.

Both the L-2-OV and the L-6-OV series are available for applications in nominal supply voltage ranges from +5 to +28 volts. The steel TO-3 and TO-66 packages that are used for these device series are hermetically sealed and offer low thermal resistance characteristics. The L-12-OV and the L-20-OV series are available for applications in nominal supply voltage range from +5 to +30 volts. The L-35-OV series are available for applications in nominal supply voltage range from +5 to +12 volts.

ABSOLUTE MAXIMUM RATINGS

PARAMETER	SYMBOL	L-2-OV SERIES		L-6-OV SERIES		L-12-OV SERIES		L-20-OV SERIES		L-35-OV SERIES	
		MIN	MAX	MIN	MAX	MIN	MAX	MIN	MAX	MIN	MAX
On State Current	I_{oc}	—	2A	—	6A	—	12A	—	20A	—	35A
On State Voltage	V_{oc}	—	2.6V	—	2.6V	—	1.6V	—	1.75V	—	2.2V
Non-Repetitive Peak Surge Current*	I_p	—	20A	—	70A	—	200A	—	260A	—	350A
Standby Current	I_s	—	35mA	—	25mA	—	30mA	—	30mA	—	30mA
Operating Temperature (Blocking)**	T_{cb}	-40°C	+100°C	-40°C	+100°C	-40°C	+100°C	-40°C	+100°C	-40°C	+100°C
Operating Temperature (Conducting)***	T_{cc}	-40°C	+150°C	-40°C	+150°C	-40°C	+140°C	-40°C	+140°C	-40°C	+140°C
Storage Temperature	T_s	-40°C	+150°C	-40°C	+150°C	-40°C	+125°C	-40°C	+125°C	-40°C	+125°C
Power Dissipation @ $T_c=25^\circ\text{C}$ Derate above 50°C	P_d		30 Watts @ 0.6 w/°C		150 Watts @ 1.5 w/°C						
Thermal Resistance	$R_{\theta JC}$		5.0°C/W		1.0°C/W						

*For sinusoidal current duration of 8.3 milliseconds max.

**Case temperature for overvoltage protector in nonconducting or "OFF" state.

***Case temperature for overvoltage protector in conducting or "ON" state. Power must be removed and case temperature allowed to drop to 71°C before application of output voltage.

The overvoltage protector requires an external heat sink to maintain case temperature below rated limit. When the overvoltage protector is used with a Lambda power supply the power supply chassis acts as the heat sink. The L-12-OV, L-20-OV, L-35-OV, overvoltage protector is supplied with mating connectors for pins on overvoltage protector (+V and -V engraved on unit).

ORDERING INFORMATION

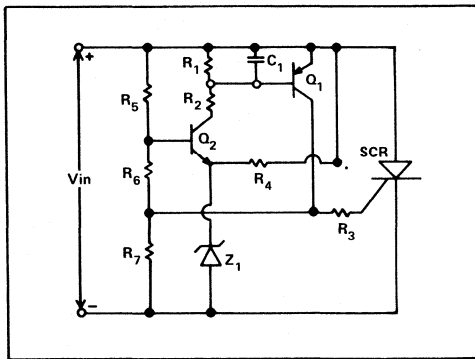
NOM SUPPLY VOLTAGE (VOLTS)	TRIP POINT VOLTAGE* (VOLTS)	2 AMP MODELS	PRICE QTY				6 AMP MODELS	PRICE QTY			
			1	100	250	1000		1	100	250	1000
5	6.6±0.2	L-2-OV-5	\$2.50	\$2.00	\$1.90	\$1.70	L-6-OV-5	\$5.00	\$4.00	\$3.75	\$3.40
6	7.3±0.2	L-2-OV-6	2.50	2.00	1.90	1.70	L-6-OV-6	5.00	4.00	3.75	3.40
12	13.7±0.4	L-2-OV-12	2.50	2.00	1.90	1.70	L-6-OV-12	5.00	4.00	3.75	3.40
15	17.0±0.5	L-2-OV-15	2.50	2.00	1.90	1.70	L-6-OV-15	5.00	4.00	3.75	3.40
18	20.5±1.0	L-2-OV-18	2.50	2.00	1.90	1.70					
20	22.8±0.7	L-2-OV-20	2.50	2.00	1.90	1.70					
24	27.3±0.8	L-2-OV-24	2.50	2.00	1.90	1.70	L-6-OV-24	5.00	4.00	3.75	3.40
28	31.9±1.0						L-6-OV-28	5.00	4.00	3.75	3.40

NOM SUPPLY VOLTAGE (VOLTS)	TRIP POINT VOLTAGE* (VOLTS)	12 AMP MODELS	PRICE QTY				20 AMP MODELS	PRICE QTY				35 AMP MODELS	PRICE QTY			
			1	100	250	1000		1	100	250	1000		1	100	250	1000
5	6.6±0.3	L-12-OV-5	\$15.30	\$11.10	\$10.45	\$9.50	L-20-OV-5	\$22.20	\$15.60	\$14.50	\$13.20	L-35-OV-5	\$27.80	\$20.13	\$18.90	\$16.90
6	7.4±0.2	L-12-OV-6	15.30	11.10	10.45	9.50	L-20-OV-6	22.20	15.60	14.50	13.20	L-35-OV-6	27.80	20.13	18.90	16.90
9	10.5±0.5	L-12-OV-9	15.30	11.10	10.45	9.50										
12	13.8±0.5	L-12-OV-12	15.30	11.10	10.45	9.50	L-20-OV-12	22.20	15.60	14.50	13.20	L-35-OV-12	27.80	20.13	18.90	16.90
15	17.0±0.5	L-12-OV-15	15.30	11.10	10.45	9.50	L-20-OV-15	22.20	15.60	14.50	13.20					
20	22.8±0.7	L-12-OV-20	15.30	11.10	10.45	9.50	L-20-OV-20	22.20	15.60	14.50	13.20					
24	27.3±0.8	L-12-OV-24	15.30	11.10	10.45	9.50	L-20-OV-24	22.20	15.60	14.50	13.20					
28	31.9±1.0	L-12-OV-28	15.30	11.10	10.45	9.50	L-20-OV-28	22.20	15.60	14.50	13.20					
30	33.5±1.0	L-12-OV-30	15.30	11.10	10.45	9.50	L-20-OV-30	22.20	15.60	14.50	13.20					

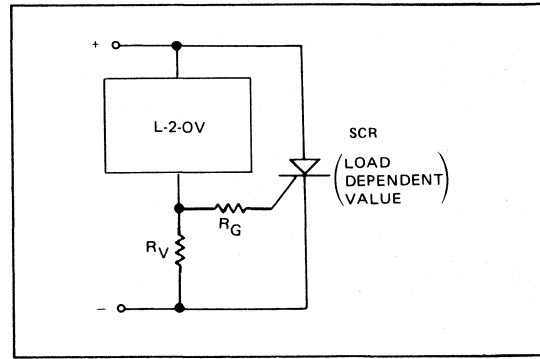
*TRIP POINT VOLTAGE TOLERANCE SPECIFIED OVER A TEMPERATURE RANGE OF 0°C TO 71°C

Contact your area sales representative for higher quantity prices. Device configurations, specifications, and prices subject to change without notice.

OVERVOLTAGE PROTECTORS



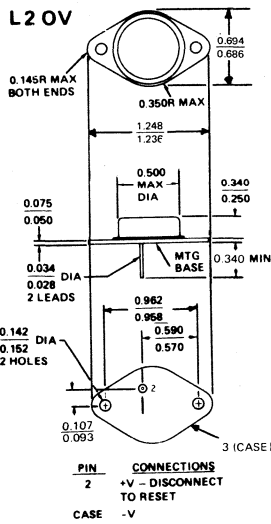
HYBRID OVERVOLTAGE PROTECTOR SCHEMATIC DIAGRAM



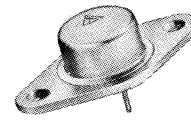
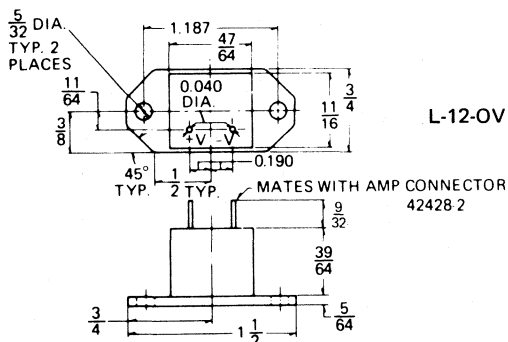
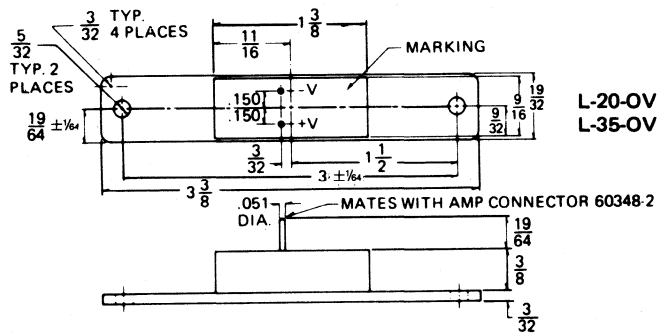
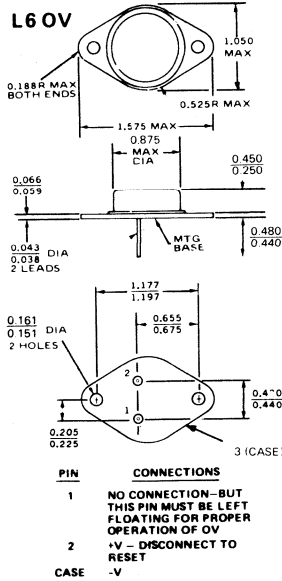
LAMBDA L-2-OV USED TO BUILD HIGHER CURRENT LEVEL OV PROTECTOR.

OUTLINE DRAWING

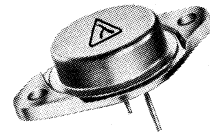
TO-66 (STEEL)



TO-3 (STEEL)



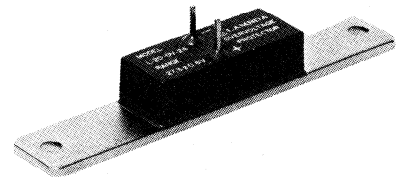
MONOLITHIC L-2-OV 2 AMP



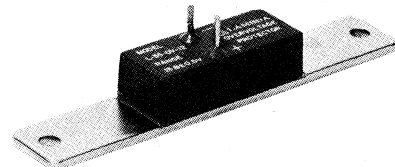
MONOLITHIC L-6-OV 6 AMP



HYBRID L-12-OV 12 AMP



HYBRID L-20-OV 20 AMP



HYBRID L-35-OV 35 AMP

SECTION IV – POWER DRIVERS

LPD 4100 SERIES

2.0 AMP POWER DRIVERS

FEATURES

- Two drivers per TO-3 package.
- Full 2 amperes of output current on each driver.
- Six dual 2 input logic configurations available. Dual AND, dual NAND, dual OR, dual NOR, Buffer, and Inverter
- Full 4 amperes of output current in a parallel mode of operation.
- 100% compatible interface with TTL, low power SCHOTTKY TTL, and CMOS
- Duty cycle up to 100% (DC)—both outputs.

DESCRIPTION

The LPD-4100 Series of Dual Power Drivers are monolithic bipolar integrated circuits capable of switching load currents up to two (2) Amperes per output. The input characteristics of these power drivers are fully compatible with TTL, Low Power Schottky-TTL, and CMOS when operated at an appropriate power supply level. The six functions available in the LPD-4100 Series include Dual 2 input AND, Dual 2 input NAND, Dual 2 input OR, Dual 2 input NOR, Dual INVERTER, and Dual BUFFER. The output of each power driver consists of an open collector transistor which is capable of sustaining 40 volts. The devices are specifically designed to interface TTL logic levels to high current, high voltage output drive requirements, such as relay drivers, lamp drivers, solenoid drivers, and stepper motors. The devices are assembled in a hermetic 8-pin TO-3 package for high reliability and low thermal resistance. Each output may be operated up to 2 Amperes at a 100% duty cycle. The outputs of the dual power driver may be paralleled to achieve a load current capability of 4 Amperes.

ABSOLUTE MAXIMUM RATINGS

PARAMETER	SYMBOL	MINIMUM	MAXIMUM	UNITS
Supply Voltage	V _{CC}	0	10	Volts
High Level Output Voltage	V _{OH}		40	Volts
Input Voltage	V _{IN}	-1.5	30	Volts
Low Level Output Current	I _{OL}		2.0	Amps
Power Dissipation (Each Driver)	P _D		3.0	Watts
Power Dissipation (Total) ⁽¹⁾	P _D (Total)		6.0	Watts
Thermal Resistance Junction to Case	θ _{Jc}		3.5	°C/Watt
Operating Junction Temperature Range	T _J	-55	150	°C
Storage Temperature Range	T _{STG}	-55	200	°C
Lead Temperature (Soldering, 60 Seconds Time Limit)	T _{LEAD}		300	°C

(1) For operation above 104°C T_{CASE}, derate @ 285mW/°C.

ELECTRICAL CHARACTERISTICS

CHARACTERISTIC	SYMBOL	JUNCTION TEMPERATURE	V _{CC}	LPD-4101 LPD-4104 LPD-4106 DRIVEN INPUT	LPD-4102 LPD-4103 LPD-4105 DRIVEN INPUT	OUTPUT	LIMITS			OTHER INPUT					
							MIN.	MAX.	UNITS	LPD-4101 DUAL AND	LPD-4104 DUAL OR	LPD-4102 DUAL NAND	LPD-4103 DUAL NOR	LPD-4105* LPD-4106**	
"1" Input Voltage	V _{IH}	0-125°C	Min				2.0	—	V						NA
"0" Input Voltage	V _{IL}	0-125°C	Min				—	0.8	V						NA
"1" Input Current	I _{IH}	0-125°C	Max	2.7V	2.7V		—	20	μA	0V	0V	0V	0V	0V	NA
"0" Input Current	I _{IL}	0-125°C	Max	0.4V	0.4V		—	50	μA	0V	0V	0V	0V	0V	NA
Input Clamp Diode Voltage	V _{CD}	0-125°C	Min	-12mA	-12mA		—	1.5	V						NA
"1" Output Reverse Current	I _{OH}	0-125°C	Min	2.0V	0V	40V	—	100	μA	2.0V	0V	2.0V	0V	0V	NA
"0" Output Voltage	V _{OL}	+25°C	Min	0.8V	2.0V	0.5A	—	0.5	V	V _{CC}	0.8V	2.0V	0.8V	0.8V	NA
"0" Output Voltage	V _{OL}	+25°C	Min	0.8V	2.0V	1.0A	—	0.8	V	V _{CC}	0.8V	2.0V	0.8V	0.8V	NA
"0" Output Voltage	V _{OL}	+25°C	Min	0.8V	2.0V	2.0A	—	1.5	V	V _{CC}	0.8V	2.0V	0.8V	0.8V	NA
"1" Output Level Supply Current	I _{CC1}	0-125°C	Max	V _{CCN}	0V		—	30	mA	V _{CCN}	V _{CCN}	0V	0V	0V	NA
"0" Output Level Supply Current	I _{CC0}	0-125°C	Max	0V	2.0V		—	400	mA	0V	0V	2.0V	2.0V	2.0V	NA
Turn-On Delay	TPD-	+25°C	V _{CCN}	See Test Figure			—	350	n Sec						
Turn-Off Delay	TPD+	+25°C	V _{CCN}	See Test Figure			—	2.0	μ Sec						

*Dual INVERTER **Dual BUFFER

SWITCHING CHARACTERISTICS

V_{CC}=5 VOLTS, V_{OH}=40 VOLTS, T_J=25°C

PARAMETER	SYMBOL	TEST CONDITIONS	MIN.	MAX.	UNITS
Turn-on Delay	TPD-	See "Switching Test Circuit"	—	350	nS
Turn-off Delay	TPD+	See "Switching Test Circuit"	—	2.0	uS

RECOMMENDED OPERATING CONDITIONS

- Supply Voltage Maximum (V_{CC} MAX) + 5.5 Volts
- Supply Voltage Nominal (V_{CCN}) + 5.0 Volts
- Supply Voltage Minimum (V_{CC} MIN) + 4.5 Volts

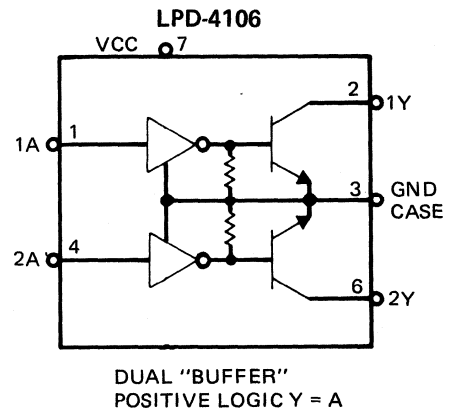
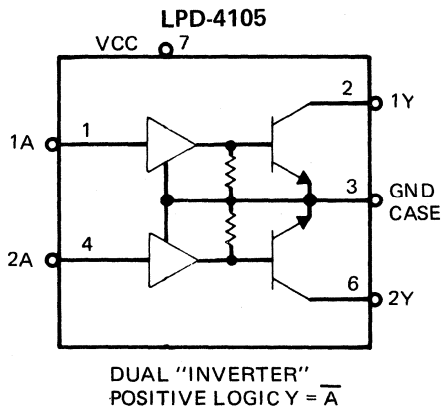
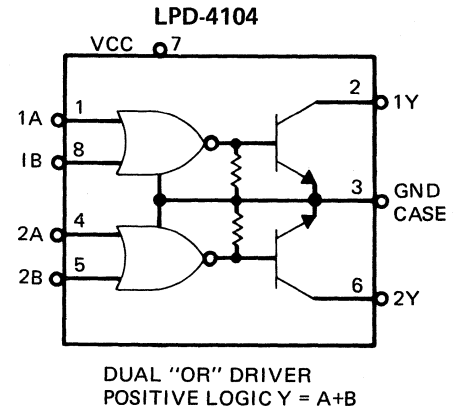
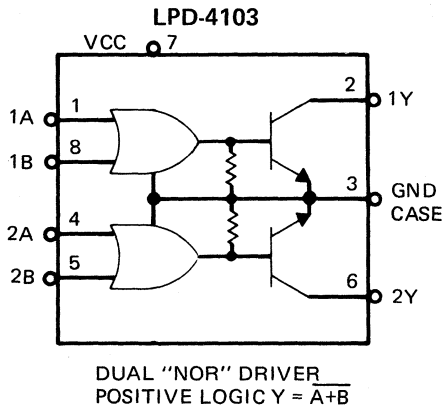
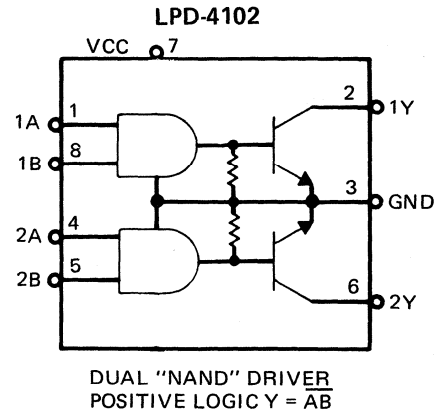
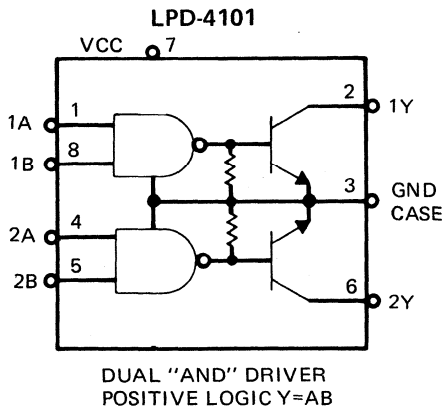
LPD 4100 SERIES

ORDERING INFORMATION

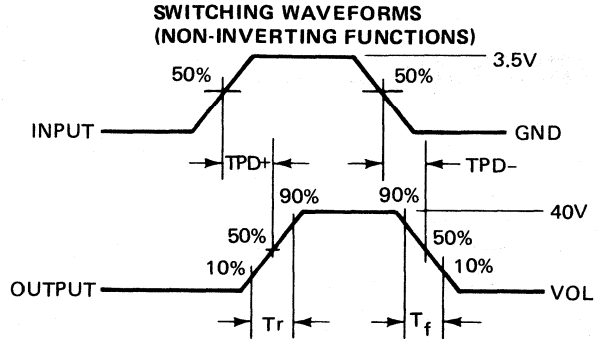
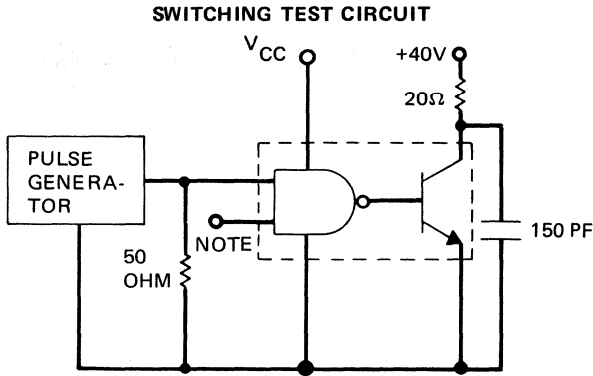
DEVICE NO.	LOGIC FUNCTION	PRICE QTY			
		1-99	100-999	1000-2499	2500-4999
LPD-4101	Dual AND	\$6.00	\$4.00	\$3.50	\$3.25
LPD-4102	Dual NAND	6.00	4.00	3.50	3.25
LPD-4103	Dual NOR	6.00	4.00	3.50	3.25
LPD-4104	Dual OR	6.00	4.00	3.50	3.25
LPD-4105	Dual INVERTER	6.00	4.00	3.50	3.25
LPD-4106	Dual BUFFER	6.00	4.00	3.50	3.25

Contact your area sales representative for higher quantity prices. Device configurations, specifications, and prices subject to change without notice.

FUNCTIONAL BLOCK DIAGRAMS



LPD 4100 SERIES



INPUT WAVEFORM: $T_R = T_F \leq 100 \text{ NS}$
DUTY CYCLE = 10%

NOTE: ALL INPUTS ARE AT EITHER LOW OR HIGH LEVELS, DEPENDING ON THE CORRECT LOGIC STATES REQUIRED IN THE TRUTH TABLES.

TRUTH TABLES

LPD-4101 "AND"		
INPUTS		OUTPUT
A	B	Y
0	0	0
0	1	0
1	0	0
1	1	1

LPD-4102 "NAND"		
INPUTS		OUTPUT
A	B	Y
0	0	1
0	1	1
1	0	1
1	1	0

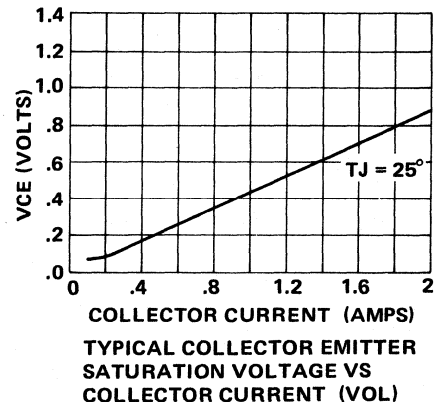
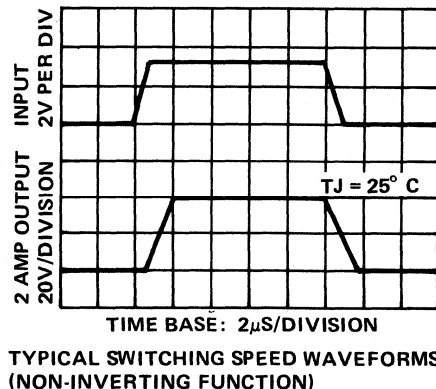
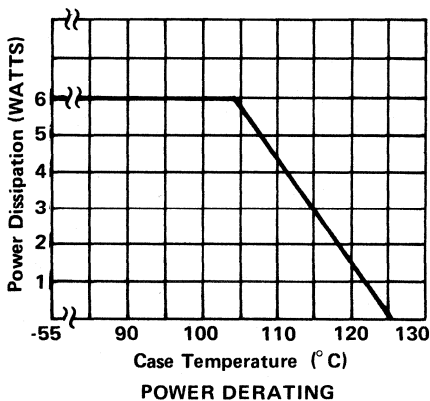
LPD-4103 "NOR"		
INPUTS		OUTPUT
A	B	Y
0	0	1
0	1	0
1	0	0
1	1	0

LPD-4104 "OR"		
INPUTS		OUTPUT
A	B	Y
0	0	0
0	1	1
1	0	1
1	1	1

LPD-4105 "INVERTER"	
INPUT	OUTPUT
A	Y
0	1
1	0

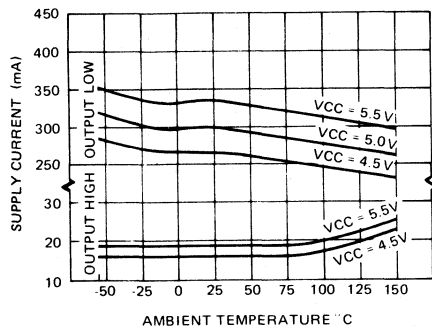
LPD-4106 "BUFFER"	
INPUT	OUTPUT
A	Y
0	0
1	1

OPERATIONAL DATA

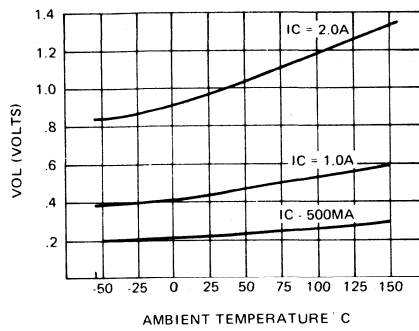


LPD 4100 SERIES

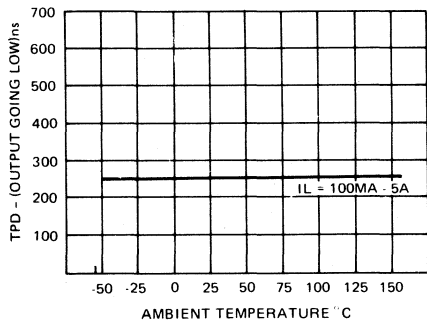
OPERATIONAL DATA



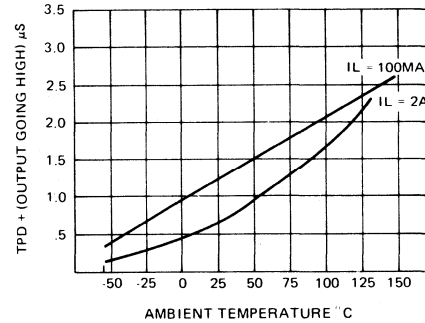
TYPICAL SUPPLY CURRENT VS AMBIENT TEMPERATURE



TYPICAL COLLECTOR-EMITTER SATURATION VOLTAGE VS AMBIENT TEMPERATURE (VOL)

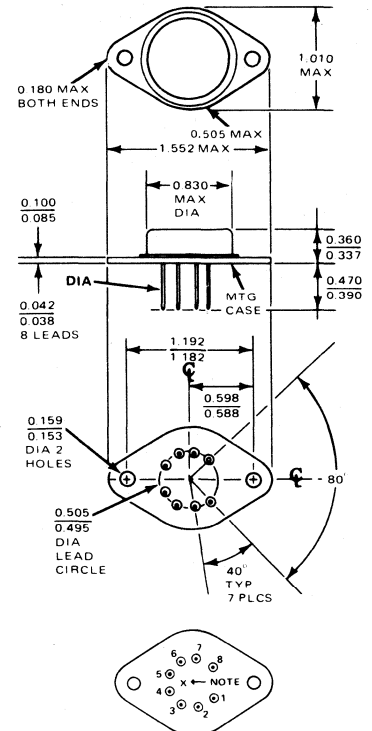


TYPICAL PROPAGATION DELAY VS TEMPERATURE



TYPICAL PROPAGATION DELAY VS TEMPERATURE

OUTLINE DRAWING

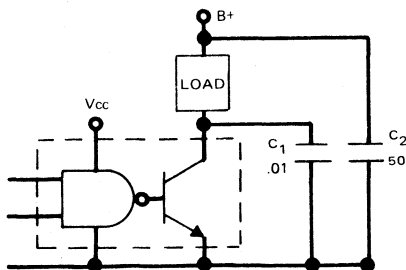


PIN	FUNCTION
1	INPUT 1
2	OUTPUT 1
3	GROUND (CASE)
4	INPUT 3
5	INPUT 4, NC FOR 4105 & 4106
6	OUTPUT 2
7	V _{CC}
8	INPUT 2, NC FOR 4105 & 4106

NOTE (X) = CASE TEMPERATURE MEASURED AT THIS POINT
NC = NO CONNECTION

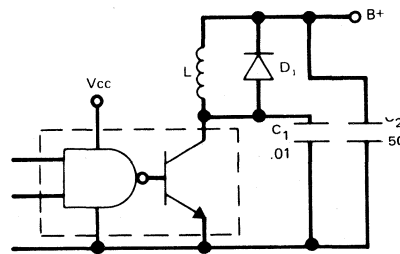
TYPICAL APPLICATIONS

BASIC DRIVING CIRCUIT



C₁ may be necessary to slow the 4100 slew rate during turn off to prevent overvoltages due to stray inductance. C₂ should be added to prevent ringing in applications where long B+ lines are used or when the B+ supply has poor transient load regulation.

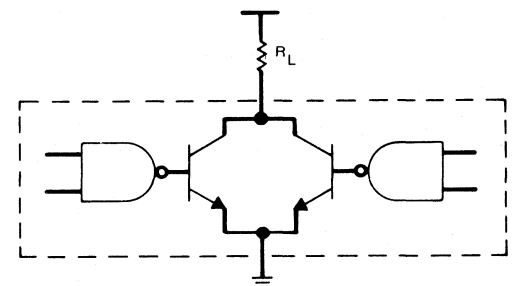
INDUCTIVE DRIVE CIRCUIT



L is an inductive load such as a relay solenoid or DC motor.

Diode D₁ is necessary to protect the 4100 output transistor from overvoltage switching transients during turnoff. D₁ should be a fast switching diode.

PARALLEL CONNECTION OF DRIVER OUTPUTS



Outputs from each driver may be connected in parallel to increase load current to 4 amps.

SECTION V- RECTIFIERS

LAMBDA PMR-27K, 31K, 35K, 36K SERIES MONOLITHIC FULL-WAVE, CENTER-TAP POWER RECTIFIERS

FEATURES

• Replaces two stud mount rectifiers

• Available in 15 and 30 Amperes per diode in each rectifier

DESCRIPTION

Four series of three-terminal monolithic Center Tap Rectifiers have been designed for use in power supplies. The PMR-27K and PMR-35K devices are a common cathode configuration; the PMR-31K and PMR-36K are a common anode configuration. Both configurations are available in average forward current ratings of 15 and 30 amperes. Peak reverse voltage ranges of 50, 100 and 200

volts are available in the PMR-27K, 31K series; 50, 100, and 200 volts are available in the PMR-35K, 36K series. All power rectifiers are hermetically sealed in either steel or copper TO-3 packages, depending on the power dissipation requirement; providing high reliability and low thermal resistance, when used with appropriate heat sinks.

ABSOLUTE MAXIMUM RATINGS

PARAMETER	SYMBOL	PMR27K050 PMR31K050	PMR27K100 PMR31K100	PMR27K200 PMR31K200	PMR35K050 PMR36K050	PMR35K100 PMR36K100	PMR35K200 PMR36K300	UNITS
Maximum Repetitive Peak Reverse Voltage	V_{RRM}	50	100	200	50	100	200	VOLTS
Maximum RMS Reverse Voltage	$V_{R(RMS)}$	35	70	141	35	70	141	VOLTS
Average Rectified Forward Current	I_o	15	15	15	30	30	30	AMPS
Peak-Repetitive Surge Current ($1/2$ cyc)	I_{FSM}	250	250	250	400	400	400	AMPS
RATING-Non-Repetitive for $1\text{mS} < t < 8.3\text{mS}$	I^2T	240	240	240	600	600	600	$A_{RMS}^2 S$
Power Dissipation Total Package ¹⁾	P_D	150	150	150	150	150	150	WATTS
Operating and Storage Temperature	T_J, T_{STG}	-65 to 200						$^{\circ}C$
Thermal Resistance Junction to Case	$R\theta_{JC}$	1.0	1.0	1.0	0.67	0.67	0.67	$^{\circ}C/WATT$

(1) For PMR-27K & 31K Series, derate @ $1W/^{\circ}C$ above $50^{\circ}C (T_C)$.
For PMR-35K & 36K Series, derate @ $1.5W/^{\circ}C$ above $100^{\circ}C (T_C)$.

ELECTRICAL CHARACTERISTICS

CROSS REFERENCE GUIDE

$T_J = 0^{\circ}C$ to $200^{\circ}C$

TEST	SYMBOL	TEST CONDITIONS	MIN.	MAX.	UNITS
Full Cycle Average Forward Voltage Drop	$V_{F(AV)}$	$I_o = 15A$ (PMR27/31) $I_o = 30A$ (PMR35/36) $f = 30\text{Hz}$, each diode		1.4	VOLTS
Maximum Reverse Current	I_{RM}	V_{RRM} , each diode		10	MA dc

R711 PMR27K100 UES2601 PMR27K050 NSR8140 PMR27K100
R711A PMR31K100 UES2601R PMR31K050 NSR8140A PMR31K100
R712 PMR27K200 UES2602 PMR27K100 NSR8141 PMR31K200
R712A PMR31K200 UES2602R PMR31K100 NSR8141A PMR31K200
UES2603 PMR27K200
UES2603R PMR31K200

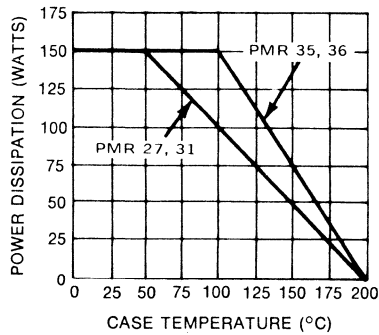
ORDERING INFORMATION

COMMON CATHODE	COMMON ANODE	V_{RRM}	I_o	PRICE QTY				
				1-99	100-249	250-999	1000-2499	2500<-5K
PMR 27K 050	PMR 31K 050	50	15	\$2.08	\$1.25	\$1.10	\$.95	\$.91
PMR 27K 100	PMR 31K 100	100	15	2.15	1.30	1.15	1.00	.96
PMR 27K 200	PMR 31K 200	200	15	2.25	1.35	1.20	1.05	1.00
PMR 35K 050	PMR 36K 050	50	30	4.75	2.85	2.57	2.18	2.10
PMR 35K 100	PMR 36K 100	100	30	5.00	3.00	2.70	2.30	2.20
PMR 35K 200	PMR 36K 200	200	30	5.25	3.15	2.83	2.42	2.30

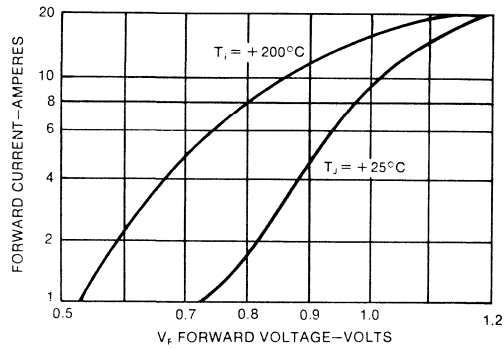
Contact your area sales representative for higher quantity prices.
Device configurations, specifications, and prices subject to change without notice.

PMR 27K, 31K, 35K, 36K SERIES

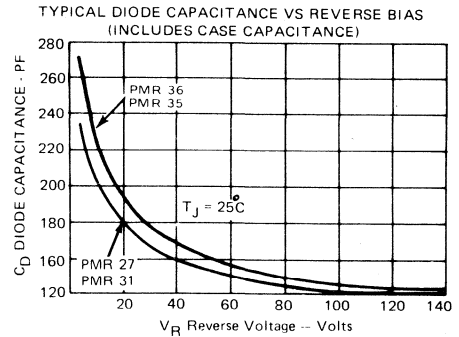
OPERATIONAL DATA



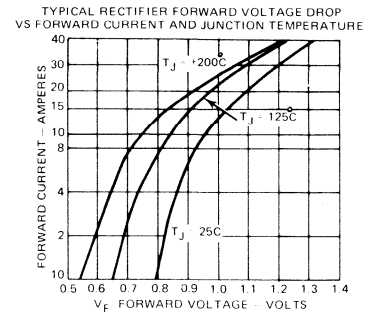
POWER DERATING



TYPICAL FORWARD VOLTAGE VS CURRENT: PMR 27,31



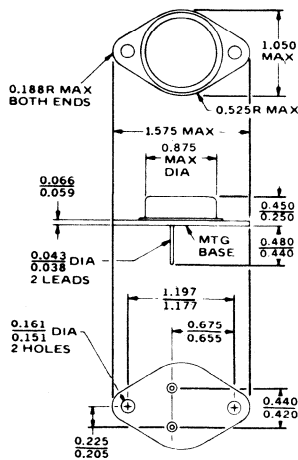
TYPICAL DIODE CAPACITANCE VS REVERSE BIAS (INCLUDES CASE CAPACITANCE)



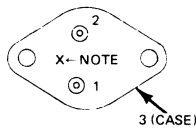
TYPICAL RECTIFIED FORWARD VOLTAGE DROP VS FORWARD CURRENT AND JUNCTION TEMPERATURE: PMR 35,36

OUTLINE DRAWINGS

TO-3 3-TERMINAL
PMR 27K, 31K



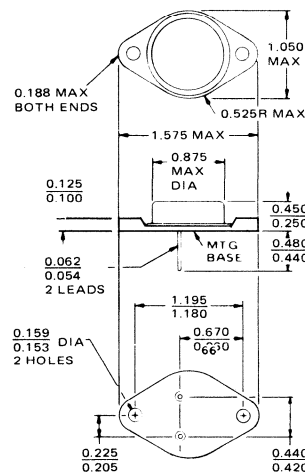
PMR-27K, 35K



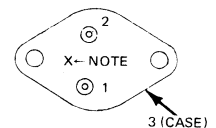
PIN	FUNCTION
1	ANODE
2	ANODE
3	COMMON CATHODE

NOTE (X) = CASE TEMPERATURE MEASURED AT THIS POINT.

TO-3 3-TERMINAL
PMR 35K, 36K



PMR-31K, 36K



PIN	FUNCTION
1	CATHODE
2	CATHODE
3	COMMON ANODE

NOTE (X) = CASE TEMPERATURE MEASURED AT THIS POINT.

PART III – DEFINITION OF TERMS

Cobo: Common base open-circuit output capacitance

The capacitance measured across the output terminals of the collector and base.

E_{s/b}: Reverse biased second breakdown energy

The maximum quantity of stored energy which can be non-destructively dissipated by a transistor during turn-off under the specified turn-off conditions.

h_{FE}: Static common emitter forward current transfer ratio

The ratio of the dc output current to the dc input current.

h_{fe}: Small-signal common emitter short-circuit forward current transfer ratio

The ratio of the ac output current to the small-signal ac input current with the output short-circuited to dc.

I_B: dc base-terminal current

The value of the dc current into the base terminal.

I_C: dc collector-terminal current

The value of the dc current into the collector terminal.

I_{CER}: Collector cutoff current, dc with a resistance between base and emitter.

The dc current into the collector terminal when it is biased into the reverse direction* with respect to the emitter terminal and the base terminal is returned to the emitter terminal through a specified resistance [R_{BE}]

*for NPN transistors—collector more positive than emitter.

*for PNP transistors—collector more negative than emitter.

I_{EBO}: Emitter cutoff current dc, collector open

The dc current into the emitter terminal when it is biased in the reverse direction with respect to the base terminal and the collector terminal is open-circuited.

I_F: Forward current

The respective value of current that flows through a semiconductor diode or rectifier diode in the forward direction.

I_{FSM}: Forward current, surge peak

The maximum (peak) surge forward current having a specified waveform and a short specified time interval.

I²T: I squared T

The measure of the maximum forward non-recurring overcurrent capability for very short pulse durations. The value is valid only for the pulse duration specified. I is in RMS AMPERES and T is in seconds.

I_O: Average forward current, 180° conduction angle, 60-Hz, half sine wave

The value of the forward current averaged over a full cycle.

I_{OPK}: Average forward current-peak

The maximum specified value of the forward current averaged over a full cycle.

I_Q: Bias current or quiescent current

Input current of the regulator that is not delivered to the load.

I_{R(REC)}: Reverse recovery current

The transient reverse current associated with a change from forward current condition to reverse condition.

I_S: Standby Current

The input current drawn by a regulator with no output load and no reference voltage load.

I_{SC}: Short-circuit current

The output current drawn with the output shorted to common.

I_{S/b}: Forward bias second breakdown collector current

The maximum amount of collector current which can be safely handled by an "ON" transistor for a specified time (usually 1 Sec.) under specified VCE conditions.

P_D: Power Dissipation

The maximum amount of power that a device can safely dissipate consistent with other device specifications limit.

P_T: Total non-reactive power input to all terminals

The sum of the products of the dc input current and voltages or the sum of the products of the instantaneous input currents and voltages.

P_{T(max)}: Total maximum non-reactive power input to all terminals

The maximum specified sum of the products of the dc input currents and voltages or the maximum specified sum of the products of the instantaneous input currents and voltages.

R_A: Ripple attenuation

The ratio of peak to peak input ripple voltage (unregulated) to the peak to peak output ripple voltage.

REG_{IN}: Line regulation

The maximum amount of output voltage change as a result of a change in the input voltage with load current and ambient temperature held constant.

REG_L: Load regulation

The maximum amount of output voltage change due to a change in the load, from no-load to full load or for a specified load change with line voltage and ambient temperature held constant.

R_{θJC}: Thermal resistance junction to case

The steady-state thermal resistance from the semiconductor or IC's junction to the case with infinite heat sink.

T_A: Ambient temperature

The temperature of ambient under operating conditions.

T_C: Temperature coefficient of regulator voltage, temperature coefficient at reference voltage

The ratio of the change in the respective regulator voltage or reference voltage, usually expressed as a percentage of the regulator or reference voltage, to the change in temperature. This ratio is the average value for the total temperature change.

T_{CASE}: Case temperature

The temperature measured at a specific location on the case of the device.

T_{CB}: Case temperature, blocking mode

The temperature measured at a specific location on the case of the device when the junction in question is reversed biased.

T_{CC}: Case temperature, conducting

The temperature measured at a specific location on the case of the device when the junction in question is forward biased.

T_J: Junction temperature

The temperature of a semiconductor junction.

T_{stg}: Storage temperature

The temperature at which the device, without any power applied, is stored.

t_d: Delay time

The time interval from the point at which the leading edge of the input pulse has reached 10% of its maximum amplitude to the point at which the leading edge of the output pulse has reached 10% of its maximum amplitude.

t_f: Fall time

The time duration during which the trailing edge of pulse is decreasing from 90% to 10% of its maximum amplitude.

t_{off}: Turn-off time

The sum of t_s + t_f.

t_{on}: Turn-on time

The sum of t_d + t_r.

t_r: Rise time

The time duration during which the leading edge of a pulse is increasing from 10% to 90% of its maximum amplitude.

t_s: Storage time

The time interval from a point 90% of the maximum amplitude on the trailing edge of the input pulse to a point 90% of the maximum amplitude on the trailing edge of the output pulse.

$V_{(BR)CBO}$: Breakdown voltage, Collector-to-base, emitter open

The breakdown voltage between the collector terminal and the base terminal when the collector terminal is biased in the reverse direction with respect to the base terminal and the emitter terminal is open circuited.

$V_{(BR)CEO}$: Breakdown voltage, Collector-to-emitter with base open

The breakdown voltage between the collector terminal and the emitter terminal when the collector terminal is biased in the reverse direction* with respect to the emitter terminal and the base terminal is open-circuited.

*for NPN transistors—collector more positive than emitter.

*for PNP transistors—collector more negative than emitter.

$V_{CER(sus)}$: Sustaining voltage, collector-to-emitter, resistance between base and emitter

The collector-to-emitter breakdown voltage at relatively high values of collector current where the breakdown voltage is relatively insensitive to changes in collector current. The base terminal is returned to the emitter terminal through a specified resistance.

$V_{(BR)EBO}$: Breakdown voltage, emitter-to-base, collector open

The breakdown voltage between the emitter and base terminals when the emitter terminal is biased

in the reverse direction with respect to the base terminal and the collector terminal is open-circuited.

$V_{BE(on)}$: Base-emitter turn-on voltage

The dc voltage between the base and emitter terminals measured in the common-emitter connections at a specified collector to emitter voltage and specified collector current.

$V_{BE(sat)}$: Saturation voltage, dc, base-to-emitter

The dc voltage between the base and emitter terminals for specified base-current conditions which are intended to insure that the collector junction is forward biased.

$V_{CE(sat)}$: Saturation voltage, dc, collector-to-emitter

The dc voltage between the collector and emitter terminals for specified saturation conditions.

V_{DC} : On-state voltage

V_{DO} : Dropout voltage

The minimum voltage between the input terminal and the output terminal at which the device ceases to regulate within specifications.

V_F : Forward voltage, dc

The dc voltage across a semiconductor diode associated with the flow of forward current.

V_{IN} : Input Voltage

The dc voltage measured between the input terminal and the common terminal.

V_N : Output noise voltage

The rms value of the random noise voltage measured between the output terminal and common with a constant load current and a dc input voltage.

V_{IN-V_O} : Input-Output differential

The voltage measured between the input terminal and the output terminal.

V_O : Output voltage

The dc voltage measured between the output terminal and the common terminal.

V_R : Reverse voltage

The voltage applied to a semiconductor diode or rectifier diode which causes the respective current to flow in the reverse direction.

V_{REF} : Reference voltage

The value of the internal voltage standard which is compared to the sense voltage during regulation.

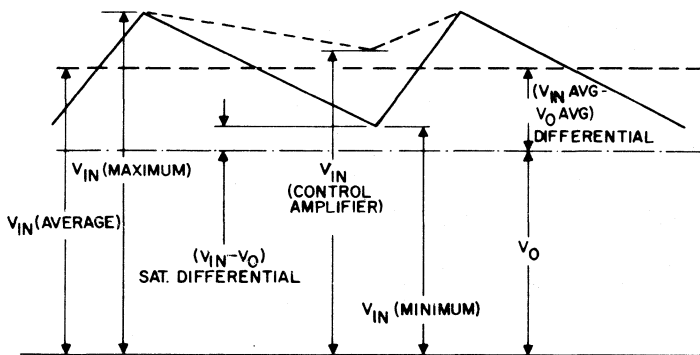
V_{RRM} : Repetitive peak reverse voltage

The maximum instantaneous value of the reverse voltage, including all repetitive transient voltages but excluding all non-repetitive transient voltages, which occurs across a semiconductor rectifier diode.

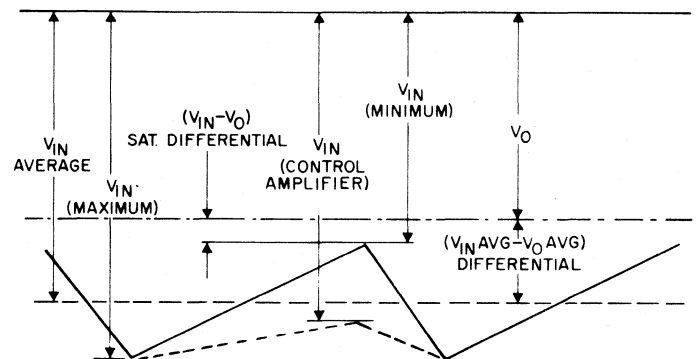
V_S : Sense voltage

Voltage that is a function of the output voltage and is used for feed back control of the regulator.

POSITIVE REGULATOR



NEGATIVE REGULATOR



GENERAL ORDERING INFORMATION

PRICES

U.S.—All prices F.O.B. Melville, N.Y.; Corpus Christi, Texas.

Canada—Please contact Montreal office for prices and handling charges of units shipped from Montreal, Canada. All prices are effective Oct. 15, 1981 and are in U.S. funds.

TERMS

Net 30 days.

TAXES

All applicable taxes, federal, state and local, are extra.

DISCOUNTS

Available to quantity buyers. For details consult with the Lambda Sales Department or Field Sales Engineer.

QUOTATIONS

All written quotations will be honored for 30 days from the date on which they are made.

DISTRIBUTION POINTS

Lambda power semiconductors are shipped from three distribution points for minimum shipping costs.

DISTRIBUTION CENTERS

1. Melville, New York
2. Corpus Christi, Texas
3. Montreal, Canada

SHIPMENT

Express truck shipment in our experience has provided the fastest and safest delivery of power devices. Unless you specifically request otherwise, your order will be forwarded in the best manner. Urgent needs can be accommodated by air express or air freight in accordance with your instructions. If devices are received damaged, it is the customer's responsibility to contact the carrier and file a claim for damages.

TESTING

Each Lambda power device is electrically tested to insure conformance with applicable specifications.

WARRANTY

We warrant each product manufactured by us, and sold by us or our authorized agents, to be free from defects in material and workmanship, and that, upon delivery, it will perform within applicable specifications. Our obligation under this warranty is limited to replacing products determined to be defective, and returned to us, in accordance with the provisions set forth below under the heading "Shipments To Factory."

This warranty is in lieu of all other warranties, express or implied and constitutes fulfillment of all our liabilities to purchaser. We do not warrant that the products can be used for any particular purpose other than those covered by the applicable specifications. We assume no liability, in any event, for consequential damages; for anticipated or lost profits, incidental damages or loss of time or other losses incurred by purchaser or any third party in connection with products covered by this warranty or otherwise.

We reserve the right to discontinue products without notice, and to make modifications in design without incurring any obligation to make such modifications to products previously sold.

SHIPMENTS TO FACTORY

If the customer notifies Lambda within 30 days after receipt of shipment that our parts have failed incoming inspection, the parts may be returned for replacement. Replacement will be issued subject to Lambda's inspection and verification of D.O.A. part. D.O.A. replacements require authorization prior to returning devices. Contact the originating factory for a return authorization number as shipments to the factory cannot be accepted without reference to this number.

All returned devices must be new, unused and in the original carton. Returned devices must be shipped to the factory within 30 days after receipt of return authorization number. Shipments must be prepaid, contain return authorization number and reason for return plus instructions for return shipment, and be packed in a manner to preclude shipping damage. Any shipping damage will be the responsibility of the customer.

PRODUCT/PRICE INDEX

VOLTAGE REGULATORS, POSITIVE MODELS

WIDE RANGE

MODEL	PAGE	V _o VOLTS	I _o	PRICE QTY							
				1-24	25-49	50-99	100-249	250-499	500-999	1000-2499	2500-4999
LAS 3700	51	0-*	6mA	\$14.00	\$13.75	\$13.15	\$12.05	\$11.65	\$10.75	\$ 9.80	\$ 9.00
LAS 723	21	2.63-38	150mA	1.15	1.15	1.15	1.04	.98	.96	.92	.86
LAS 723B	21	2.63-48	150mA	1.19	1.19	1.19	1.08	1.02	1.02	.96	.90
LAS 1000	21	2.63-38	150mA	1.39	1.39	1.39	1.24	1.17	1.10	1.10	1.03
LAS 1100	21	2.63-48	150mA	1.64	1.64	1.64	1.43	1.40	1.40	1.33	1.23
LAS 15U	25	4-30	1.5A	4.25	3.86	3.55	3.33	3.09	2.94	2.78	2.63
LLM 317	33	1.2-37.0	1.5A	3.75	3.10	3.10	2.65	2.65	2.65	2.00	1.90
LAS 16U	35	4-30	2.0A	4.50	4.20	3.90	3.60	3.40	3.25	3.08	2.85
LAS 14AU	38	4-35	3.0A	9.00	8.00	7.50	7.25	6.22	5.55	4.95	4.60
LLM 350	41	1.2-33.0	3.0A	5.75	4.80	4.80	4.00	4.00	4.00	3.35	3.20
LAS 19U	43	4-30	5.0A	16.00	14.50	13.75	10.25	9.75	9.00	8.50	8.00
LLM 338	46	1.2-32.0	5.0A	7.60	6.39	6.39	5.00	5.00	5.00	4.15	4.00
LAS 39U	48	4-16	8.0A	19.00	17.50	16.75	14.00	14.00	12.90	11.65	11.00

* Limited only by external components

FIXED

MODEL	PAGE	I _o AMPS	P _o WATTS	PRICE QTY							
				1-24	25-49	50-99	100-249	250-499	500-999	1000-2499	2500-4999
5V											
LAS 15A05	25	1.5	15	\$ 2.75	\$ 2.50	\$ 2.30	\$ 2.15	\$ 2.00	\$ 1.90	\$ 1.80	\$ 1.70
LAS 1505	25	1.5	15	2.05	1.88	1.73	1.61	1.50	1.43	1.35	1.28
LAS 1605	35	2.0	20	3.00	2.80	2.60	2.40	2.24	2.15	2.05	1.90
LAS 1405	38	3.0	30	5.00	4.75	4.50	4.30	4.10	3.90	3.70	3.40
LAS 1905	43	5.0	50	14.00	12.50	11.75	10.00	7.85	7.85	7.00	6.85
LAS 2105	53	5.0	85	47.00	43.00	35.00	31.00	31.00	26.00	24.00	—
LAS 3905	48	8.0	80	18.00	16.50	15.75	13.00	13.00	11.90	10.65	10.00
LAS 3905K	48	8.0	80	16.00	14.25	13.80	11.65	10.75	9.85	9.50	9.25
6V											
LAS 1506	25	1.5	15	\$ 2.05	\$ 1.88	\$ 1.73	\$ 1.61	\$ 1.50	\$ 1.43	\$ 1.35	\$ 1.28
LAS 1406	38	3.0	30	5.00	4.75	4.50	4.30	4.10	3.90	3.70	3.40
8V											
LAS 1508	25	1.5	15	\$ 2.05	\$ 1.88	\$ 1.73	\$ 1.61	\$ 1.50	\$ 1.43	\$ 1.35	\$ 1.28
LAS 1608	35	2.0	20	3.00	2.80	2.60	2.40	2.24	2.15	2.05	1.90
10V											
LAS 1510	25	1.5	15	\$ 2.05	\$ 1.88	\$ 1.73	\$ 1.61	\$ 1.50	\$ 1.43	\$ 1.35	\$ 1.28
12V											
LAS 15A12	25	1.5	15	\$ 2.75	\$ 2.50	\$ 2.30	\$ 2.15	\$ 2.00	\$ 1.90	\$ 1.80	\$ 1.70
LAS 1512	25	1.5	15	2.05	1.88	1.73	1.61	1.50	1.43	1.35	1.28
LAS 1612	35	2.0	20	3.00	2.80	2.60	2.40	2.24	2.15	2.05	1.90
LAS 1412	38	3.0	30	5.00	4.75	4.50	4.30	4.10	3.90	3.70	3.40
LAS 1912	43	5.0	50	14.00	12.50	11.75	10.00	7.85	7.85	7.00	6.85
LAS 2112	53	5.0	85	47.00	43.00	35.00	31.00	31.00	26.00	24.00	—
13.8V											
LAS 15CB	25	1.5	15	\$ 2.75	\$ 1.88	\$ 1.73	\$ 1.61	\$ 1.50	\$ 1.43	\$ 1.35	\$ 1.28
LAS 16CB	35	2.0	20	3.00	2.80	2.60	2.40	2.24	2.15	2.05	1.90
15V											
LAS 15A15	25	1.5	15	\$ 2.75	\$ 2.50	\$ 2.30	\$ 2.15	\$ 2.00	\$ 1.90	\$ 1.80	\$ 1.70
LAS 1515	25	1.5	15	2.05	1.88	1.73	1.61	1.50	1.43	1.35	1.28
LAS 1615	35	2.0	20	3.00	2.80	2.60	2.40	2.24	2.15	2.05	1.90
LAS 1415	38	3.0	30	5.00	4.75	4.50	4.30	4.10	3.90	3.70	3.40
LAS 1915	43	5.0	50	14.00	12.50	11.75	10.00	7.85	7.85	7.00	6.85
LAS 2115	53	5.0	85	47.00	43.00	35.00	31.00	31.00	26.00	24.00	—
18V											
LAS 1518	25	1.5	15	\$ 2.05	\$ 1.88	\$ 1.73	\$ 1.61	\$ 1.50	\$ 1.43	\$ 1.35	\$ 1.28
20V											
LAS 1520	25	1.5	15	\$ 2.05	\$ 1.88	\$ 1.73	\$ 1.61	\$ 1.50	\$ 1.43	\$ 1.35	\$ 1.28
24V											
LAS 1524	25	1.5	15	\$ 2.05	\$ 1.88	\$ 1.73	\$ 1.61	\$ 1.50	\$ 1.43	\$ 1.35	\$ 1.28
LAS 2124	53	3.0	85	47.00	43.00	35.00	31.00	31.00	26.00	24.00	—
28V											
LAS 1528	25	1.5	15	\$ 2.05	\$ 1.88	\$ 1.73	\$ 1.61	\$ 1.50	\$ 1.43	\$ 1.35	\$ 1.28

LAMBDA IMPROVED SECOND SOURCE PRODUCTS

MODEL	V _o VOLTS	I _o	PRICE QTY								
			1-24	25-49	50-99	100-249	250-499	500-999	1000-2499	2500-4999	
LMC 7805CK	5.0	1.5	\$2.15	\$2.00	\$1.88	\$1.50	\$1.50	\$1.50	\$1.45	\$1.40	\$1.40
LMC 7806CK	6.0	1.5	2.15	2.00	1.88	1.50	1.50	1.50	1.45	1.40	1.40
LMC 7808CK	8.0	1.5	2.15	2.00	1.88	1.50	1.50	1.50	1.45	1.40	1.40
LMC 7812CK	12.0	1.5	2.15	2.00	1.88	1.50	1.50	1.50	1.45	1.40	1.40
LMC 7815CK	15.0	1.5	2.15	2.00	1.88	1.50	1.50	1.50	1.45	1.40	1.40
LMC 7818CK	18.0	1.5	2.15	2.00	1.88	1.50	1.50	1.50	1.45	1.40	1.40
LMC 7824CK	24.0	1.5	2.15	2.00	1.88	1.50	1.50	1.50	1.45	1.40	1.40
LLM 340K-5	5.0	1.5	2.05	1.90	1.75	1.45	1.45	1.45	1.40	1.35	1.35
LLM 340K-6	6.0	1.5	2.05	1.90	1.75	1.45	1.45	1.45	1.40	1.35	1.35
LLM 340K-8	8.0	1.5	2.05	1.90	1.75	1.45	1.45	1.45	1.40	1.35	1.35
LLM 340K-12	12.0	1.5	2.05	1.90	1.75	1.45	1.45	1.45	1.40	1.35	1.35
LLM 340K-15	15.0	1.5	2.05	1.90	1.75	1.45	1.45	1.45	1.40	1.35	1.35
LLM 340K-18	18.0	1.5	2.05	1.90	1.75	1.45	1.45	1.45	1.40	1.35	1.35
LLM 340K-24	24.0	1.5	2.05	1.90	1.75	1.45	1.45	1.45	1.40	1.35	1.35
LLM 323K	3.0	30	4.50	4.25	4.00	3.80	3.60	3.40	3.20	2.90	2.90

VOLTAGE REGULATORS

ADJUSTABLE

MODEL	PAGE	I _o AMPS	P _D WATTS	PRICE QTY						
				1- 24	25- 49	50- 99	100- 249	250- 499	500- 999	1000
5V ± 5%										
LAS 2205	53	5.0	85	\$ 47.00	\$ 43.00	\$ 35.00	\$31.00	\$31.00	\$26.00	\$24.00
LAS 3205	59	10.0	140	54.00	47.00	41.00	35.00	35.00	30.00	29.00
LAS 5205	63	20.0	270	112.00	91.00	77.00	65.00	65.00	58.00	54.00
LAS 7205	66	30.0	400	161.00	132.00	112.00	95.00	95.00	90.00	78.00
6V ± 5%										
LAS 2206	53	5.0	85	\$ 47.00	\$ 43.00	\$ 35.00	\$31.00	\$31.00	\$26.00	\$24.00
12V ± 5%										
LAS 2212	53	5.0	85	\$ 47.00	\$ 43.00	\$ 35.00	\$31.00	\$31.00	\$26.00	\$24.00
LAS 3212	59	8.5	140	54.00	47.00	41.00	35.00	35.00	30.00	29.00
LAS 5212	63	15.0	270	112.00	91.00	77.00	65.00	65.00	58.00	54.00
15V ± 5%										
LAS 2215	53	5.0	85	\$ 47.00	\$ 43.00	\$ 35.00	\$31.00	\$31.00	\$26.00	\$24.00
LAS 3215	59	8.0	140	54.00	47.00	41.00	35.00	35.00	30.00	29.00
LAS 5212	63	15.0	270	112.00	91.00	77.00	65.00	65.00	58.00	54.00
LAS 7215	66	22.0	400	161.00	132.00	112.00	95.00	95.00	90.00	78.00
24V ± 5%										
LAS 2224	53	3.0	85	\$ 47.00	\$ 43.00	\$ 35.00	\$31.00	\$31.00	\$26.00	\$24.00
LAS 3224	59	7.5	140	54.00	47.00	41.00	35.00	35.00	30.00	29.00
LAS 5224	63	14.0	270	112.00	91.00	77.00	65.00	65.00	58.00	54.00
LAS 7224	66	21.0	400	161.00	132.00	112.00	95.00	95.00	90.00	78.00
28V ± 5%										
LAS 2228	53	2.0	85	\$ 47.00	\$ 43.00	\$ 35.00	\$31.00	\$31.00	\$26.00	\$24.00
LAS 3228	59	7.0	140	54.00	47.00	41.00	35.00	35.00	30.00	29.00

NEGATIVE MODELS

WIDE RANGE

MODEL	PAGE	V _o VOLTS	I _o	PRICE QTY							
				1- 24	25- 49	50- 99	100- 249	250- 499	500- 999	1000- 2499	2500- 4999
LAS 18U	29	-30 to -2.6	1.5A	\$ 5.40	\$ 5.03	\$ 4.73	\$ 4.50	\$ 4.28	\$ 4.05	\$ 3.82	\$ 3.60

FIXED

MODEL	PAGE	I _o AMPS	P _D	PRICE QTY							
				1- 24	25- 49	50- 99	100- 249	250- 499	500- 999	1000- 2499	2500- 4999
-2V											
LAS 1802	29	1.5	15	\$ 2.95	\$ 2.70	\$ 2.47	\$ 2.30	\$ 2.10	\$ 1.95	\$ 1.80	\$ 1.60
-5V											
LAS 18A05	29	1.5	15	\$ 3.60	\$ 3.35	\$ 3.15	\$ 3.00	\$ 2.85	\$ 2.70	\$ 2.55	\$ 2.40
LAS 1805	29	1.5	15	2.95	2.70	2.47	2.30	2.10	1.95	1.80	1.60
-5.2V											
LAS 18052	29	1.5	15	\$ 2.95	\$ 2.70	\$ 2.47	\$ 2.30	\$ 2.10	\$ 1.95	\$ 1.80	\$ 1.60
-6V											
LAS 1806	29	1.5	15	\$ 2.95	\$ 2.70	\$ 2.47	\$ 2.30	\$ 2.10	\$ 1.95	\$ 1.80	\$ 1.60
-8V											
LAS 1808	29	1.5	15	\$ 2.95	\$ 2.70	\$ 2.47	\$ 2.30	\$ 2.10	\$ 1.95	\$ 1.80	\$ 1.60
-10V											
LAS 1810	29	1.5	15	\$ 2.95	\$ 2.70	\$ 2.47	\$ 2.30	\$ 2.10	\$ 1.95	\$ 1.80	\$ 1.60
-12V											
LAS 18A12	29	1.5	15	\$ 3.60	\$ 3.35	\$ 3.15	\$ 3.00	\$ 2.85	\$ 2.70	\$ 2.55	\$ 2.40
LAS 1812	29	1.5	15	2.95	2.70	2.47	2.30	2.10	1.95	1.80	1.60
LAS 2812	53	2.5	85	47.00	44.00	36.00	32.00	32.00	28.00	26.00	—
-15V											
LAS 18A15	29	1.5	15	\$ 3.60	\$ 3.35	\$ 3.15	\$ 3.00	\$ 2.85	\$ 2.70	\$ 2.55	\$ 2.40
LAS 1815	29	1.5	15	2.95	2.70	2.47	2.30	2.10	1.95	1.80	1.60
-18V											
LAS 1818	29	1.5	15	\$ 2.95	\$ 2.70	\$ 2.47	\$ 2.30	\$ 2.10	\$ 1.95	\$ 1.80	\$ 1.60
-20V											
LAS 1820	29	1.5	15	\$ 2.95	\$ 2.70	\$ 2.47	\$ 2.30	\$ 2.10	\$ 1.95	\$ 1.80	\$ 1.60
-24V											
LAS 1824	29	1.5	15	\$ 2.95	\$ 2.70	\$ 2.47	\$ 2.30	\$ 2.10	\$ 1.95	\$ 1.80	\$ 1.60
-28V											
LAS 1828	29	1.5	15	\$ 2.95	\$ 2.70	\$ 2.47	\$ 2.30	\$ 2.10	\$ 1.95	\$ 1.80	\$ 1.60

VOLTAGE REGULATORS

ADJUSTABLE

MODEL	PAGE	I _o AMPS	P _D WATTS	PRICE QTY						
				1- 24	25- 49	50- 99	100- 249	250- 499	500- 999	1000
-5V ± 5%										
LAS 2705	53	5.0	85	\$47.00	\$44.00	\$36.00	\$32.00	\$32.00	\$28.00	\$26.00
-12V ± 5%										
LAS 2712	53	5.0	85	\$47.00	\$44.00	\$36.00	\$32.00	\$32.00	\$28.00	\$26.00
-15V ± 5%										
LAS 2715	53	5.0	85	\$47.00	\$44.00	\$36.00	\$32.00	\$32.00	\$28.00	\$26.00
-24V ± 5%										
LAS 2724	53	3.0	85	\$47.00	\$44.00	\$36.00	\$32.00	\$32.00	\$28.00	\$26.00

LAMBDA IMPROVED SECOND SOURCE PRODUCTS

MODEL	V _o VOLTS	I _o	PRICE QTY								
			1- 24	25- 49	50- 99	100- 249	250- 499	500- 999	1000- 2499	2500- 4999	
LMC 7905CK	5.0	1.5	\$2.90	\$2.80	\$2.52	\$2.00	\$2.00	\$2.00	\$1.90	\$1.80	
LMC 7906CK	6.0	1.5	2.90	2.80	2.52	2.00	2.00	2.00	1.90	1.80	
LMC 7908CK	8.0	1.5	2.90	2.80	2.52	2.00	2.00	2.00	1.90	1.80	
LMC 7912CK	12.0	1.5	2.90	2.80	2.52	2.00	2.00	2.00	1.90	1.80	
LMC 7915CK	15.0	1.5	2.90	2.80	2.52	2.00	2.00	2.00	1.90	1.80	
LMC 7918CK	18.0	1.5	2.90	2.80	2.52	2.00	2.00	2.00	1.90	1.80	
LMC 7924CK	24.0	1.5	2.90	2.80	2.52	2.00	2.00	2.00	1.90	1.80	
LLM 320K-5	5.0	1.5	3.30	3.00	2.75	2.35	2.35	2.35	2.15	2.00	
LLM 320K-6	6.0	1.5	3.30	3.00	2.75	2.35	2.35	2.35	2.15	2.00	
LLM 320K-8	8.0	1.5	3.30	3.00	2.75	2.35	2.35	2.35	2.15	2.00	
LLM 320K-12	12.0	1.5	3.30	3.00	2.75	2.35	2.35	2.35	2.15	2.00	
LLM 320K-15	15.0	1.5	3.30	3.00	2.75	2.35	2.35	2.35	2.15	2.00	
LLM 320K-18	18.0	1.5	3.30	3.00	2.75	2.35	2.35	2.35	2.15	2.00	
LLM 320K-24	24.0	1.5	3.30	3.00	2.75	2.35	2.35	2.35	2.15	2.00	

MONOLITHIC SWITCHING REGULATORS

DEVICE TYPE	PAGE	PRICE QTY								
		1- 24	25- 49	50- 99	100- 249	250- 499	500- 999	1000- 2499	2500- 4999	
LAS 3800	2	\$ 9.95	\$ 7.95	\$ 7.95	\$6.95	\$6.95	\$6.95	\$6.20	\$5.95	
LAS 6300K	10	13.00	11.00	10.25	9.50	8.75	7.50	7.00	6.60	
LAS 6300L	10	10.00	9.25	8.50	8.00	7.50	6.75	5.75	5.00	

POWER TRANSISTOR DARLINGTONS

MODEL	PAGE	VOLTAGE RATING VOLTS	RATED POWER WATTS	PRICE QTY			
				1- 99	100- 999	1000- 2499	2500- 4999
PMD10K40	69	40	150	\$1.90	\$1.55	\$1.39	\$1.33
PMD10K60	69	60	150	2.10	1.68	1.50	1.44
PMD10K80	69	80	150	2.32	1.86	1.67	1.60
PMD10K100	69	100	150	2.70	2.16	1.94	1.86
PMD11K40	69	40	150	1.90	1.55	1.39	1.33
PMD11K60	69	60	150	2.10	1.68	1.50	1.44
PMD11K80	69	80	150	2.32	1.86	1.67	1.60
PMD11K100	69	100	150	2.70	2.16	1.94	1.86
PMD12K40	69	40	100	1.10	.88	.81	.78
PMD12K60	69	60	100	1.14	.91	.84	.81
PMD12K80	69	80	100	1.20	.95	.87	.85
PMD12K100	69	100	100	1.25	1.00	.92	.89
PMD13K40	69	40	100	1.10	.88	.81	.78
PMD13K60	69	60	100	1.14	.91	.84	.81
PMD13K80	69	80	100	1.20	.95	.87	.85
PMD13K100	69	100	100	1.25	1.00	.92	.89
PMD1601K	69	60	180	2.95	2.35	2.10	2.02
PMD1602K	69	80	180	3.35	2.67	2.40	2.30
PMD1603K	69	100	180	3.90	3.10	2.78	2.66
PMD1701K	69	60	180	2.95	2.35	2.10	2.02
PMD1702K	69	80	180	3.35	2.67	2.40	2.30
PMD1703K	69	100	180	3.90	3.10	2.78	2.66
PMD16K60	69	60	225	4.68	3.78	3.42	3.33
PMD16K80	69	80	225	5.05	4.15	3.50	3.42
PMD16K100	69	100	225	5.60	4.60	3.90	3.80
PMD17K60	69	60	225	5.20	4.20	3.80	3.70
PMD17K80	69	80	225	5.60	4.60	3.90	3.80
PMD17K100	69	100	225	6.16	5.06	4.30	4.18
PMD18K60	69	60	240	5.38	4.35	3.93	3.83
PMD18K80	69	80	240	5.80	4.77	4.03	3.93
PMD18K100	69	100	240	6.44	5.29	4.49	4.37
PMD19K60	69	60	240	5.92	4.78	4.33	4.21
PMD19K80	69	80	240	6.38	5.25	4.43	4.37
PMD19K100	69	100	240	7.08	5.82	4.93	4.81

MODEL	VOLTAGE RATING VOLTS	RATED POWER WATTS	PRICE QTY			
			1-99	100-999	1000-2499	2500-4999
2N6050	60	150	\$3.04	\$2.25	\$2.10	\$1.75
2N6051	80	150	3.38	2.50	2.25	1.94
2N6052	100	150	3.78	2.80	2.50	2.25
2N6053	60	100	1.96	1.45	1.35	1.05
2N6054	80	100	2.30	1.70	1.53	1.13
2N6055	60	100	1.38	1.05	1.05	1.05
2N6056	80	100	1.69	1.25	1.18	1.13
2N6057	60	150	2.50	1.85	1.80	1.75
2N6058	80	150	2.70	2.00	1.97	1.94
2N6059	100	150	3.17	2.35	2.30	2.25
2N6282	60	160	3.38	2.50	2.35	2.20
2N6283	80	160	3.65	2.70	2.53	2.32
2N6284	100	160	4.05	3.00	2.90	2.75
2N6285	60	160	4.05	3.00	2.92	2.80
2N6286	80	160	4.32	3.20	3.10	3.00
2N6287	100	160	4.73	3.50	3.40	3.30
2N6383	40	100	1.51	1.12	1.09	1.05
2N6384	60	100	1.69	1.25	1.19	1.13
2N6385	80	100	1.82	1.35	1.28	1.19
2N6648	40	100	1.51	1.12	1.09	1.05
2N6649	60	100	1.69	1.25	1.19	1.13
2N6650	80	100	1.82	1.35	1.28	1.19

MODEL	VOLTAGE RATING VOLTS	RATED POWER WATTS	PRICE QTY			
			1-99	100-999	1000-2499	2500-4999
LMJ 900	60	90	\$1.69	\$1.25	\$1.18	\$1.05
LMJ 901	80	90	1.96	1.45	1.36	1.12
LMJ 1000	60	90	1.49	1.10	1.08	1.05
LMJ 1001	80	90	1.76	1.30	1.22	1.12
LMJ 2500	60	150	3.30	2.50	2.35	1.70
LMJ 2501	80	150	3.70	2.80	2.63	1.85
LMJ 3000	60	150	2.70	2.05	1.93	1.65
LMJ 3001	80	150	3.20	2.40	2.26	1.85
LMJ 4030	60	150	5.00	3.75	3.53	2.40
LMJ 4031	80	150	5.50	4.15	3.90	3.00
LMJ 4032	100	150	6.50	4.85	4.56	3.12
LMJ 4033	60	150	4.30	3.25	3.06	2.40
LMJ 4034	80	150	4.85	3.60	3.38	3.00
LMJ 4035	100	150	5.55	4.20	3.95	3.12
LMJ 11011	60	200	6.05	4.50	4.41	4.33
LMJ 11012	60	200	5.40	4.00	3.92	3.87
LMJ 11013	90	200	6.61	4.90	3.86	4.81
LMJ 11014	90	200	5.86	4.35	4.28	4.23

SWITCHING POWER DARLINGTONS

MODEL	PAGE	VOLTAGE RATING VOLTS	CURRENT RATING AMPS	PRICE QTY							
				1-24	25-49	50-99	100-249	250-499	500-999	1000-2499	2500-4999
PMD20K120	17	120	14	\$ 8.10	\$7.20	\$6.61	\$6.39	\$5.76	\$4.85	\$4.41	\$3.97
PMD20K150	17	150	14	9.00	8.00	7.35	7.10	6.40	5.39	4.90	4.41
PMD20K200	17	200	14	11.00	9.90	9.00	8.70	7.80	6.60	6.00	5.40
PMD25K120	17	120	9	5.40	4.80	4.45	4.32	3.87	3.27	2.97	2.67
PMD25K150	17	150	9	6.00	5.45	4.95	4.80	4.30	3.63	3.30	2.97
PMD25K200	17	200	9	7.20	6.60	6.00	5.80	5.20	4.40	4.00	3.60

HIGH SPEED SWITCHING TRANSISTORS

MODEL	PAGE	VOLTAGE RATING VOLTS	CURRENT RATING AMPS	PRICE QTY			
				1-99	100-999	1000-2499	2500-4999
PMS3K425	12	425	3	\$3.35	\$2.70	\$2.40	\$2.35
PMS5K425	12	425	5	4.30	3.45	3.10	3.05
PMS8K425	12	425	8	9.30	7.45	6.70	6.55

DUAL POWER DRIVERS

MODEL	PAGE	LOGIC FUNCTION	PRICE QTY			
			1-99	100-999	1000-2499	2500-4999
LPD-4101	80	Dual AND	\$6.00	\$4.00	\$3.50	\$3.25
LPD-4102	80	Dual NAND	6.00	4.00	3.50	3.25
LPD-4103	80	Dual NOR	6.00	4.00	3.50	3.25
LPD-4104	80	Dual OR	6.00	4.00	3.50	3.25
LPD-4105	80	Dual INVERTER	6.00	4.00	3.50	3.25
LPD-4106	80	Dual BUFFER	6.00	4.00	3.50	3.25

OVERVOLTAGE PROTECTORS

MODEL	PAGE	NON SUPPLY VOLTAGE VOLTS	TRIP POINT VOLTAGE VOLTS	PRICE QTY			
				1-99	100-249	250-999	1000
L-2-OV-5	78	5	6.6±0.2	\$ 2.50	\$ 2.00	\$ 1.90	\$ 1.70
L-2-OV-6	78	6	7.3±0.2	2.50	2.00	1.90	1.70
L-2-OV-12	78	12	13.7±0.4	2.50	2.00	1.90	1.70
L-2-OV-15	78	15	17.0±0.5	2.50	2.00	1.90	1.70
L-2-OV-18	78	18	20.5±1.0	2.50	2.00	1.90	1.70
L-2-OV-20	78	20	22.8±0.7	2.50	2.00	1.90	1.70
L-2-OV-24	78	24	27.3±0.8	2.50	2.00	1.90	1.70
L-6-OV-5	78	5	6.6±0.2	5.00	4.00	3.75	3.40
L-6-OV-6	78	6	7.3±0.2	5.00	4.00	3.75	3.40
L-6-OV-12	78	12	13.7±0.4	5.00	4.00	3.75	3.40
L-6-OV-15	78	15	17.0±0.5	5.00	4.00	3.75	3.40
L-6-OV-24	78	24	27.3±0.8	5.00	4.00	3.75	3.40
L-6-OV-28	78	28	31.9±1.0	5.00	4.00	3.75	3.40
L-12-OV-5	78	5	6.6±0.3	15.30	11.10	10.45	9.50
L-12-OV-6	78	6	7.4±0.2	15.30	11.10	10.45	9.50
L-12-OV-9	78	9	10.5±0.5	15.30	11.10	10.45	9.50
L-12-OV-12	78	12	13.8±0.5	15.30	11.10	10.45	9.50
L-12-OV-15	78	15	17.0±0.5	15.30	11.10	10.45	9.50
L-12-OV-20	78	20	22.8±0.7	15.30	11.10	10.45	9.50
L-12-OV-24	78	24	27.3±0.8	15.30	11.10	10.45	9.50
L-12-OV-28	78	28	31.9±1.0	15.30	11.10	10.45	9.50
L-12-OV-30	78	30	33.5±1.0	15.30	11.10	10.45	9.50
L-20-OV-5	78	5	6.6±0.3	22.20	15.60	14.50	13.20
L-20-OV-6	78	6	7.4±0.2	22.20	15.60	14.50	13.20
L-20-OV-12	78	12	13.8±0.5	22.20	15.60	14.50	13.20
L-20-OV-15	78	15	17.0±0.5	22.20	15.60	14.50	13.20
L-20-OV-20	78	20	22.8±0.7	22.20	15.60	14.50	13.20
L-20-OV-24	78	24	27.3±0.8	22.20	15.60	14.50	13.20
L-20-OV-28	78	28	31.0±1.0	22.20	15.60	14.50	13.20
L-20-OV-30	78	30	33.5±1.0	22.20	15.60	14.50	13.20
L-35-OV-5	78	5	6.6±0.3	27.80	20.13	18.90	16.90
L-35-OV-6	78	6	7.4±0.2	27.80	20.13	18.90	16.90
L-35-OV-12	78	12	13.8±0.5	27.80	20.13	18.90	16.90

FULL-WAVE, CENTER-TAP RECTIFIERS

MODEL	PAGE	V _{RRM} VOLTS	I _o AMPS	PRICE QTY				
				1-99	100-249	250-999	1000-2499	2500-4999
COMMON CATHODE								
PMR27K050	84	50	15	\$2.08	\$1.25	\$1.10	\$.95	\$.91
PMR27K100	84	100	15	2.15	1.30	1.15	1.00	.96
PMR27K200	84	200	15	2.25	1.35	1.20	1.05	1.00
PMR35K050	84	50	30	4.75	2.85	2.57	2.18	2.10
PMR35K100	84	100	30	5.00	3.00	2.70	2.30	2.20
PMR35K200	84	200	30	5.25	3.15	2.83	2.42	2.30
COMMON ANODE								
PMR31K050	84	50	15	\$2.08	\$1.25	\$1.10	\$.95	\$.91
PMR31K100	84	100	15	2.15	1.30	1.15	1.00	.91
PMR31K200	84	200	15	2.25	1.35	1.20	1.05	1.00
PMR36K050	84	50	30	4.75	2.85	2.57	2.18	2.10
PMR36K100	84	100	30	5.00	3.00	2.70	2.30	2.20
PMR36K200	84	200	30	5.25	3.15	2.83	2.42	2.30

NOTES

NOTES

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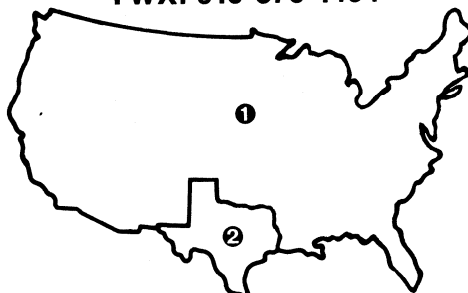
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